

Partial Discharge Characterization of New Environmentally Friendly Insulating Liquids

A study on the Insulating
Behaviour of Transformer Liquids

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Behaviour of Transformer Liquids

by

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Preface

To begin with, I would like to mention my father, Eng. Rafi Fazlalizadeh, who was always by my side. Without his unwavering support and attention, this work would not have been possible. He consistently encouraged and supported me, even when he was seriously ill. I also extend my deepest gratitude to my mother for providing for me during the early stages of my life and for her unreserved help and patience during my absence while completing this work.

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*M. Fazlalizadeh
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Summary

Unwanted breakdown of insulation is one of the primary challenges affecting the reliability of power systems. Liquid insulation is commonly used in power grids and subsea installations due to its preferred qualities, such as heat transfer and safety. Some insulating liquids have been available since the earliest times, and some have been proven to be harmful to humans, animals, and the environment. Consequently, the search for better candidates has always been ongoing. Unfortunately, the knowledge of the quality of the insulating liquids is insufficient since numerous phenomena related to electrical discharge and breakdown in insulating liquids happen unexpectedly, often surprising the personnel involved.

Therefore, gaining a deep understanding of the breakdown physics of these liquids and mechanisms leading to the degradation of their dielectric properties is crucial. This understanding ensures that the new replacements are suitable for specific equipment. Partial discharge (PD), as one of the initial stages before breakdown and a major player in ageing the insulators, is the main focus of this work. One way to monitor a liquid's insulation quality is through the Partial Discharge Inception Voltage (PDIV) level. Based on PDIV measurements, there are defined standards, such as IEC 61294. This work is focused on investigating the behaviour of different types of liquids under the AC voltage stress above and below the PDIV level to determine if the defined standards reflect the dielectric characteristics of each liquid properly.

Three liquids have been examined, including two candidates from commonly used insulating liquids in electrical installations. Tests are performed in a needle-plane geometry under high-voltage AC stress. The same liquids and geometry have been used multiple times in previous studies under lower-frequency AC voltage, yielding unexpected results. Therefore, this study aims to investigate deeper by testing the liquids at higher frequencies and comparing observations with previous results. The experimental setup used for this work includes a 20 kV resonance voltage source, and the needle-plane gap distance is 20 mm. The current flow from the plane to the needle is mostly capacitive in the range up to 100 nA, but space charges are proved to play a major role in the conductive component of the current. PD behaviour has been found to relate to molecular structure and, in some cases, to the previous stresses, likely due to residual ions remaining from previous cycles. The presence and nature of space charges have been investigated carefully by testing liquids at higher frequencies below and above the PDIV level. Further, the test setup was facilitated, and half-cycle voltages were applied to involve only one polarity of space charges, and the differences were observed. Another part of this work involved the optical PD measurement method, which utilized two photomultiplier tubes (PMTs) and a coincidence circuit to filter out ambient dark noise. This method was employed alongside the well-known charge acquisition method to detect the initial PD pulses in the liquid. The results demonstrated a noteworthy difference between the two methods, prompting further investigation.

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Nomenclature

Abbreviations

Abbreviation	Definition
AC	Alternative Current
DC	Direct Current
HV	High voltage
IEC	International Electrotechnical Commission
ISA	International Standard Atmosphere
ISO	International Organization for Standardization
PCB	Polychlorinated Biphenyl
PCT	Polychlorinated Terphenyls
PD	Partial Discharge
PDIV	Partial Discharge Inception Voltage
PMT	Photo Multiplier Tube
RMS	Root Mean Square
SCLC	Space Charge Limited Current
UV	Ultraviolet Radiation
CFD	Constant fraction discriminator

Symbols

Symbol	Definition	Unit
V	Voltage	[V]
r_0	Needle tip radius	[m]
J	Current density	[A/mm ²]
ρ	Charge density	[C/mm ³]
J_0	The current density of charges with a charge density of ρ_0	[A/mm ²]
e	amount of charge of an electron	[C]
m_e	mass of an electron	[g]
q	Charge	[C]
C	Capacitance	[F]
E	Electric field	[V/m]
r_b	Radius of the crosssection	[m]
f	Frequency	[1/s]
g	Gain	
Q	Quality factor	
ρ	Density	[kg/m ³]
μ	Mobility	[m ² V ⁻¹ s ⁻¹]
τ	Period	[s]
ν	Velocity	[m/s]
μ_z, ν_z, ϕ_z	ellipsoid coordinate axis moments	
η, ζ	spherical coordinate moments	

1

Introduction

1.1. On the History of Insulators

The need for a good insulator has a history as old as the commercial use of electric energy. There have been numerous materials tested and employed since the 18th century as insulators. From the pieces of wood within primitive electrical circuits to the most advanced thermal and electrical insulator materials inside spaceships nowadays. They all use different kinds of materials as insulators to prevent the electric current from flowing in unwanted directions.

1.1.1. Transformer Oil as an Insulator

The primary function of the transformer oil is to bring good insulation between conductors inside the transformer. Another important duty is cooling the transformer coils and taking away the heat generated inside the transformer due to losses. It must, therefore, have high dielectric strength, thermal conductivity, and chemical stability and should keep these properties even at high temperatures for extended periods. Transformer oil or insulating oil is a stable oil at high temperatures and has excellent electrical insulating properties. It has been used vastly in oil-filled transformers, switchgear, and cables.

They preferably have flash points greater than 140°C (284°F), pour points less than -40°C (-40°F), and a dielectric breakdown greater than $28\text{kV}_{\text{RMS}}/\text{mm}$ [1].

1.1.2. Diversity of the Liquids Used for This Purpose

A key insulating oil for use inside transformers is mineral oil. petroleum-based oils were the primitive oils that were used in 1892. however, because of the lower price, the industry moved to paraffin-based mineral oil besides its low properties like high pour point or the generation of insoluble sludge. transitioning from 1925, naphtha-based oil was commonly used in transformers, but some drawbacks, like high flammability, put installations at huge risk [2]. In 1930, Polychlorinated biphenyl (PCB) was introduced as an alternative. The flammability concerns 1899 the industry had moved to solved, but PCB's toxicity and non-biodegradable nature posed environmental hazards [3]. By revealing the danger of these materials the scope of interest has been changed to other liquids like silicon oil, introduced in 1970, replicated PCB's properties, but its widespread use was limited due to high prices.

The 1990s saw the advent of synthetic esters, which were mostly synthesised from Alcohols and acids [4], offering enhanced dielectric strength, biodegradability, lower viscosity, and higher fire resistance. A histogram of the development of the transformer oil has been shown in Figure 1.2.

As environmental regulations tightened, vegetable oil and natural esters gained prominence. Subsequently, in 2006, a blend comprising 80% mineral oil and 20% synthetic ester proved to enhance insulating oil performance [5]. Between 2008 and 2012, there was a shift towards vegetable oil-based palm fatty acid esters, alongside research exploring nanofluids' potential in transformer insulation [6, 7]

1.2. Environmental Impact of Transformer Liquids

Due to mineral oil's numerous drawbacks, such as limited biodegradability, potential future scarcity, and the presence of poly-nuclear aromatic hydrocarbons, which can be easily released into the environment in the event of a transformer explosion, ester oils have garnered attention as a promising alternative to mineral oil [4, 8]. contamination risk of the water supply is more severe compared to the contamination risk of soil [9]. Environmental concerns about using the mineral oils have been increased due to the low ratio of biodegradability (less than 20% over a 28-day period [10]) of the mineral oils.

1.2.1. European Commission Directive on the Disposal of PCBs/PCTs

PCBs and PCTs, synthetic compounds manufactured extensively worldwide from the 1930s to the 1980s, found broad applications in lubricants and various electrical and hydraulic equipment. Their widespread use has resulted in their presence being detected in soil, surface water, groundwater, and even food sources. Classified as probable human carcinogens, these chemicals induce various adverse effects in both human and animal populations.

On 16 September 1996, the European Commission identified PCBs as *highly toxic* and *heavily re-*

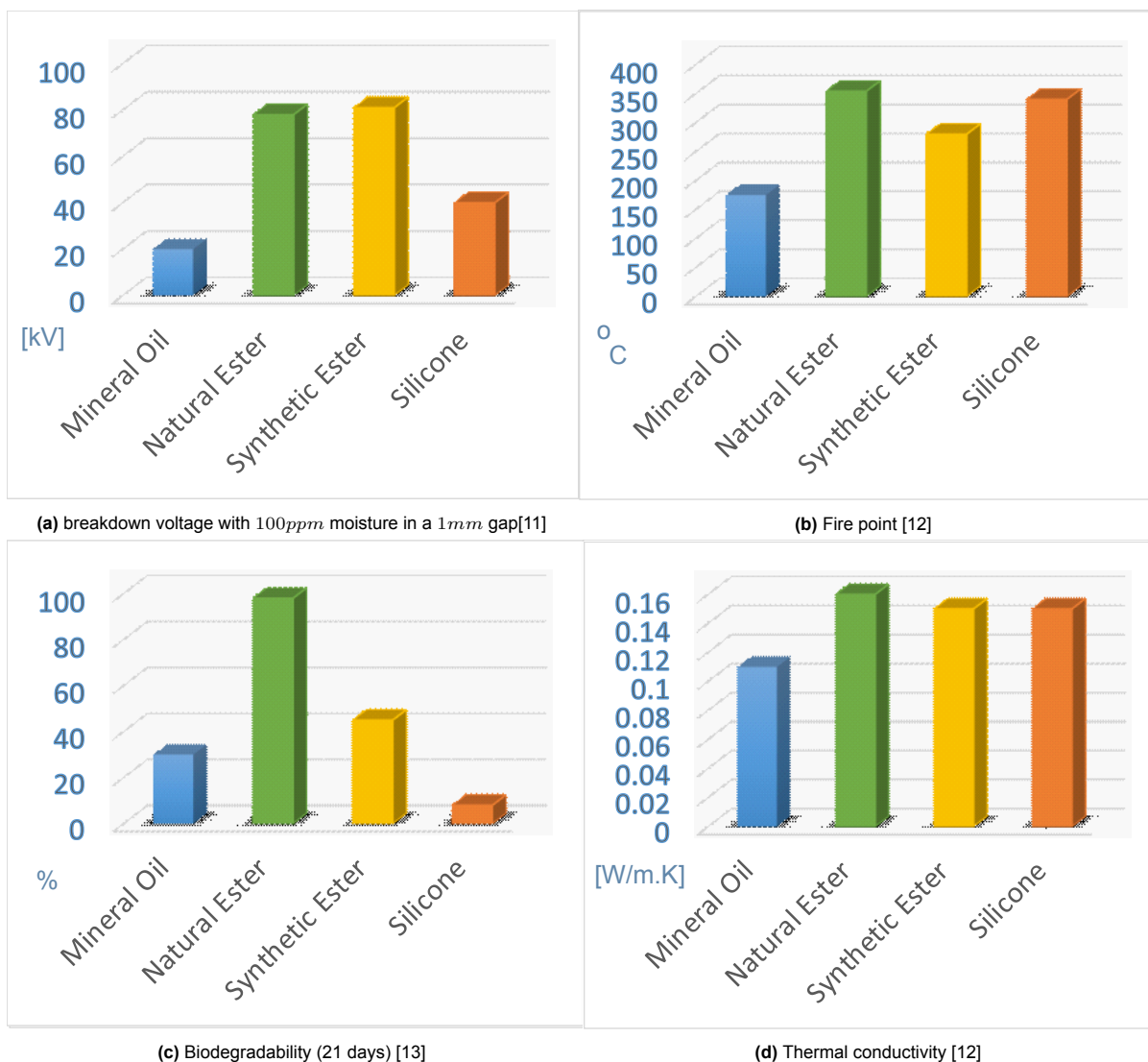


Figure 1.1: Comparison of typical characteristics of transformer oils

stricted. Therefore, a time schedule has been defined for the disposal of these products[14]. Products containing these chemicals are still being discarded, and so this type of waste must be handled with great care [14].

1.2.2. New Environmentally Friendly Liquids

The development of new insulating liquid materials is driven by various factors, including environmental requirements, safety concerns, and economic considerations. Consequently, transformer manufacturers are encountering new specifications aligned with these evolving demands. Vegetable oil-based transformer fluids are increasingly supplanting mineral oil-based products in the market due to their superior performance and notable environmental and safety benefits.

In 1892, the feasibility of using ester oil extracted from seeds was tested; however, no significant efforts were undertaken until the severe oil crisis of 1970. Researchers once again turned their attention to the use of vegetable oils due to their high biodegradability (> 95%), low toxicity, high flash points, and reduced flammability, making them more environmentally friendly. There are many important characteristics that different oils can be compared for use as transformer oil. but the most important characteristics that can be named for this purpose include electrical, physical, and chemical characteristics. Among these characteristics, there are some important ones like biodegradability, flash fire point, and price and production capacity.

Figure 1.1 and Table 1.1 show comparisons of some of the important characteristics between different types of transformer oils.

biodegradability is a two-sided matter. It is good to have a transformer oil that can last for longer, but on the other hand, if for any reason it spills into the environment, it is better to have a high biodegradability factor to dissolve quickly.

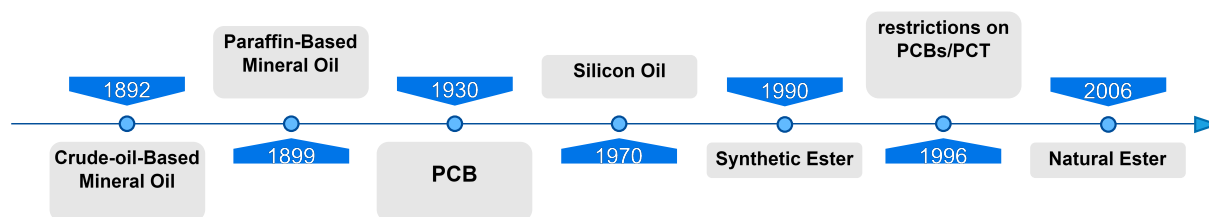


Figure 1.2: Histogram of development of the transformer liquids over time

1.3. Objectives of This Work

While various standards exist to assess the quality and properties of transformer oils [15, 16], there remains a significant question regarding whether these standards truly reflect the performance of these liquids when used as transformer oil. This uncertainty arises because each standard was developed based on the properties of the prevalent liquid at the time of its inception.

Initially, this approach was appropriate as it allowed for the differentiation between high-quality, functional transformer oil and inferior, unsuitable alternatives, thus establishing a means to regulate transformer oil quality. However, with the introduction of new liquids, such as synthetic ester, as relatively green transformer oils and newer candidates, these standards may no longer align with the characteristics of these newer liquids. In this study, our focus lies on examining the dielectric characteristics of different insulating liquids. The aim is specifically to investigate deeper into the partial discharge formation in dielectric liquids.

Experimenting with various liquids to observe their electrical behaviour under high voltage AC voltage with different frequencies is the main approach of this work. Careful investigation of the process of inception of partial discharge across different liquids and the role of space charges in initiating partial discharges within the liquids will be studied.

Partial discharge detection with different methods to evaluate the procedure of PD is one of the goals of this work. Assessing the accuracy and reliability of the evaluation process of standard based on insights garnered from experimental observations across different insulation liquids will be performed. A schematic of the thesis outline can be found in Figure 1.3.

Table 1.1: Comparison transformer oils typical values [17]

Property	Mineral oils	Silicone oils	Synthetic esters	Vegetable oils	Test method
Dielectric breakdown, kV	30–85	35–60	45–70	82–97	IEC 60156
Relative permittivity at 25°C	2.1–2.5	2.6–2.9	3.0–3.5	3.1–3.3	IEC 60247
Viscosity at 0°C, mm ² s ⁻¹	76	81–92	26–50	143–77	ISO 3104
at 40°C	3–16	35–40	14–29	16–37	
at 100°C	2–2.5	15–17	4–6	4–8	
Pour point, °C	–30 to –60	–50 to –60	–40 to –50	–19 to –33	ISO 3016
Flash point, °C	100–170	300–310	250–270	315–328	ISO 2592 (1)
Fire point, °C	110–185	340–350	300–310	350–360	
Density at 20°C, kg dm ⁻³	0.83–0.89	0.96–1.10	0.90–1.00	0.87–0.92	ISO 3675
Specific heat, Jg ⁻¹ K ⁻¹	1.6–2.0	1.5	1.8–2.3	1.5–2.1	ASTM E1269
Thermal conductivity, Wm ⁻¹ K ⁻¹	0.11–0.16	0.15	0.15	0.16–0.17	DCS
Expansion coefficient, 10 ⁻⁴ K ⁻¹	7–9	10–6.5	10–5.5–5.9		ASTM D1903

(1) Cleveland open cup procedure.

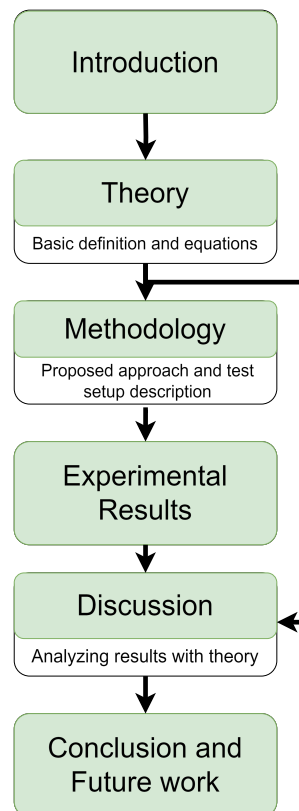


Figure 1.3: Flowchart illustrating the thesis outline

2

Theory

2.1. Electrical Standards Related to Transformer Oils

The International Electrotechnical Commission (IEC) is a worldwide organization that standardizes electrical and electronic equipment in close cooperation with ISO. IEC has defined different standards for evaluating the electrical characteristics of transformer oils.

Some important standards with their release date have been listed in Table 2.1. By examining the years of release of the standards along with the histogram depicted in Figure 1.2, it becomes evident that certain standards were established before introducing newer environmentally friendly transformer oils. A careful focus specifically on the IEC 61294 standard is required, which outlines the inception voltage of partial discharges in transformer liquids.

However, revisiting some basic physical concepts is crucial to prior to a technical discussion of the IEC 61294 standard.

Table 2.1: Standards related to transformer oil

Standard Code	Year of Definition	Description
IEC 60156	1963	Method for the determination of the electric strength of insulating oils
IEC 60270	1981	Partial discharge measurements
IEC 60897	1987	Methods for the determination of the lightning impulse breakdown voltage of insulating liquids
IEC 1072	1991	Methods of test for evaluating the resistance of insulating materials against the initiation of electrical trees
ISO 5725	1986	Precision of the test methods - Determination of repeatability and reproducibility for a standard test method by inter-laboratory tests
IEC 61294	1993	Insulating liquids - Determination of the partial discharge inception voltage (PDIV) - Test procedure
SS-EN 60455-3-1	2003	Resin-based reactive compounds used for electrical insulation - Part 3: Specifications for individual materials - Sheet 1: Unfilled epoxy resinous compounds
SS-EN 61100	1993	Classification of insulating liquids according to fire point and net calorific value
SS-IEC 628	1986	Insulating oils - Gassing of insulating liquids under electrical stress and ionization

2.1.1. IEC 61294 Standard for Determining PDIV

IEC 61294 is a technical report that states the progress of a method to measure PDIV [15]. According to [15], PDIV of an insulating liquid is the lowest voltage at which a PD occurs of an apparent charge equal to or exceeding 100 pC when the sample is tested under the specific conditions. The purpose of the configuration of the test cell is to create a point-sphere gap. It consists of a 12.5 mm to 13 mm diameter steel ball bearing and a point electrode which should be $3 \text{ }\mu\text{m}$ radius tip with a distance of $50 \text{ mm} \pm 1 \text{ mm}$ between two electrodes. The volume is about 300 ml [15]. The high voltage circuit shall provide an AC voltage frequency 48 Hz to 62 Hz fully sinusoidal and PD free [15]. The PD measuring device shall be in range of 10 kHz to 300 kHz [15].

test liquid shall be filtered through the vacuum by a $5 \text{ }\mu\text{m}$ sintered glass [15]. the procedure includes filling the test cell and waiting for some time, then increasing the voltage with the rate of 1 kV/s until a 100 pC partial discharge is observed and rapidly bringing the voltage to zero and repeat it for 10 times with the same test sample and after that do the test once again with a new needle and also renew the liquid and execute the same test. Then calculate the mean value of two inception voltages [15].

the proposed test cell in this standard has been shown in the Figure 2.1.

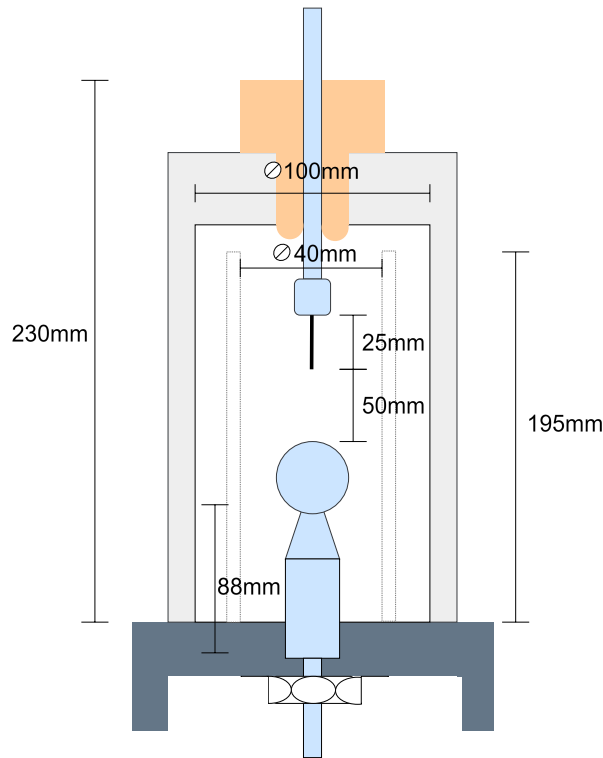


Figure 2.1: Dimensions of an IEC 61294 standard test cell for PDIV test implemented from [15]

2.2. Needle-Plane Geometry in Electrostatics

Point-plane geometry is a configuration in electrostatics that involves analyzing the electric field produced by a point electrode in the vicinity of an infinite conducting plane. In this setup, a singular point electrode is positioned at a certain distance above or below an infinitely large, flat conducting surface. The primary objective of studying point-plane geometry, which is also known as needle-plane geometry (will be mentioned as needle-plane after this), is to understand the distribution of electric field and the resulting electric potential in the space between needle and plane.

The resulting nonlinear electric field in the space between the point and the plane is according to the Laplacian field. This configuration can go up to 1000 kV/mm without breaking down in a reasonable test volume.

Various analytical methods exist for calculating the electric field in a needle-plane configuration. There are two main approaches to approximating the needle tip curve. It can be considered as a hyperboloid or as a cone.

Radu has shown in a comparative study [18] that the hyperboloidal model of the needle is more precise for calculating the field strength at the tip of the needle (the maximum field) and along the symmetrical axis. However, this model leads to significant errors in calculating the field for other points, especially for the points neighbouring the area of the conical surface of the needle.

2.2.1. Hyperboloidal Model of the Needle Tip

By considering the prolonged ellipsoid system of coordinates (μ_z, ν_z, ϕ_z) , Durand [19] and Mason [20] obtained a general relation of the electric field in every point inside the domain D which is axially symmetrical.

$$E_{M,N}(\mu_z, \nu_z) = \frac{2U}{Ac} \frac{1}{\sqrt{(\mu_z^2 - \nu_z^2)(1 - \nu_z^2)}} \quad (2.1)$$

where U is the difference in potentials of the electrodes ($U = V_1 - V_2$, in Figure 2.2), d is the distance between electrodes, $c = \sqrt{d^2 + dr_0}$ the focal distance of the homo focal hyperboloid and ellipsoids,

$A = \ln((c+d)/(c-d))$ is a constant, and the axis domains are $\mu_z \in [1, \infty)$, $\nu_z \in [0, d/c]$, and $\phi_z \in [0, 2\pi]$

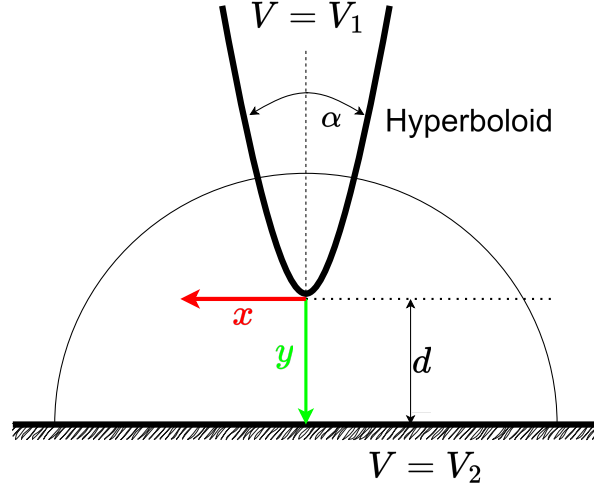


Figure 2.2: Hyperboloidal needle-plane geometry adopted from [18]

Bamji in [21] investigates the use of the boundary element method to calculate electric field profiles at needle tips. The study compares the boundary element method with the finite difference and finite element methods. The paper discusses the importance of accurately determining local electric fields in polymeric insulation to prevent degradation and failure of high-voltage devices. It explains the limitations of the finite difference and finite element methods and introduces the boundary element method as an alternative approach. The relation for the electrical field at the needle tip, which is placed at the coordinate origin, is [21]:

$$E(y) = E_{max} \frac{dr_0}{dr_0 + 2dy - y^2} \quad (2.2)$$

$$E_{max} = \frac{2U}{r_0 \cdot \ln(1 + 4d/r_0)}$$

In [18], Radu has used this method to estimate the electrostatic field in a needle plate arrangement with the characteristics as ($r_0 = 10\mu\text{m}$, $d = 2.5\text{mm}$, $U = 5\text{kV}$). This analytical model has been compared with a numerical model based on the finite element method using a Cartesian system of coordinates, using $n = 14000$ nodes and a $m = 27492$ finite elements network. The errors of the analytical method compared to the numerical method have been concluded in the Table 2.2. The difference reference is:

$$\epsilon_{an} = (E_n - E_a)/E_n \quad (2.3)$$

Radu has considered the deviation of the numerical method to be less than 1% at the needle tip and below 0.1% for the points located at more than $10\mu\text{m}$ from the needle tip. It is obviously a deviation of 0.22% at the closest point that, if added to the possible deviation of the numerical method, could result in a total deviation of 1.22% from the real electric field at the needle tip.

2.2.2. Sensitivity of Electric Field to the Geometry

Meanwhile, the hyperboloidal model might give a good approximation for the electric field at the needle tip. However, the Laplace equation is a useful tool for approximating the electric field in the points with distance to the needle tip. According to Moon and Spencer [22], the voltage over the gap can be defined as:

$$V(\eta) = B \ln \tan(\eta/2) \quad (2.4)$$

therefore, the electric field can be defined as

$$E = \frac{1}{a \sqrt{\sinh^2 \zeta + \sin^2 \eta}} \frac{B}{\sin \eta} \hat{\eta} \quad (2.5)$$

Table 2.2: Comparison of the Laplacian electrical field on the needle axis by analytical (E_a) and numerical(E_n) approaches adopted from [18]

$y(\mu\text{m})$	0.2	1	4	10	30	50	100	400	1500	2400
$E_a(\text{kV}/\text{mm})$	139.8	121.1	80.75	48.42	20.82	13.3	7.04	1.94	0.687	0.579
$E_n(\text{kV}/\text{mm})$	139.5	121.2	80.99	48.71	20.97	13.4	7.09	1.96	0.689	0.579
$\epsilon_{an}(\%)$	-0.22	0.08	0.3	0.58	0.71	0.7	0.69	0.62	0.31	0.05

Where a and B are defined as follows:

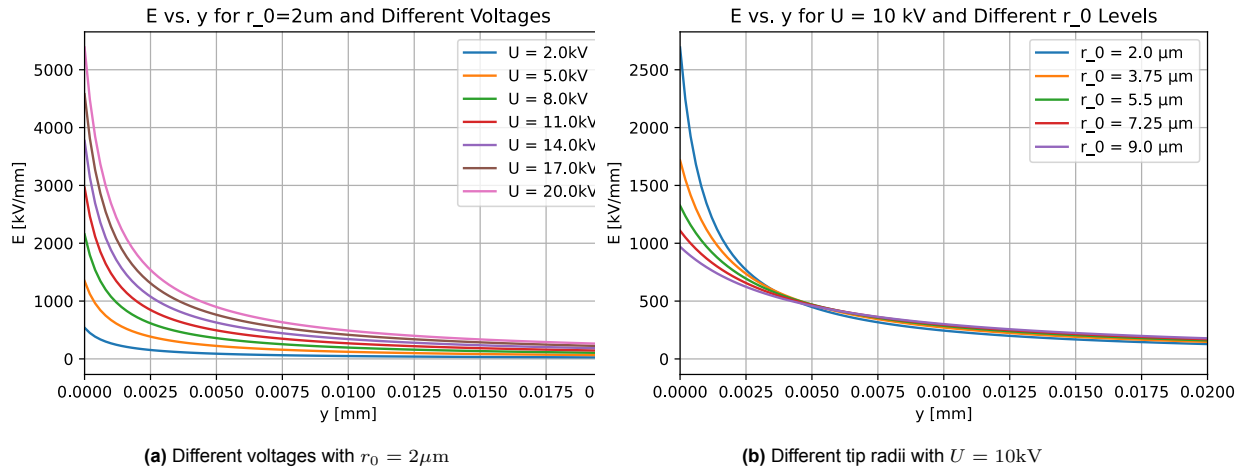
$$B = \frac{V}{\ln(\tan(\frac{\pi}{2} - \sqrt{(1 - \frac{d}{a})/2}))} \quad (2.6)$$

$$a = d + \frac{r_p}{2} \quad (2.7)$$

Based on the above relation, the sensitivity of the electrical field to the parameters can be obtained from equation Equation 2.8 [23].

$$\frac{\Delta E_{tip}}{E} = \frac{\Delta V}{V} + \frac{\Delta r_p}{r_p} \left(1 + \frac{1}{\ln \frac{4d}{r_p}}\right) + \frac{\Delta d}{d} \left(\frac{1}{\ln \frac{4d}{r_p}}\right) \quad (2.8)$$

According to this equation, any change in the voltage will change the field with the same amount. A change in the magnitude of $1 \mu\text{m}$ of the tip radius of the needle around $5 \mu\text{m}$ will change the electrical field by 22%, and a change in the magnitude of 1 mm over $d = 20\text{mm}$ will change the electric field by 0.5% [23]. Therefore, the electric field intensity changes drastically by any changes in the needle tip radius [23]. Figure 2.3 showing the results of Equation 2.2 for different voltage levels and different tip radii in range with the values that will be used in this work. They clearly show the exponential increase in the field strength by decreasing the distance to the needle tip.

**Figure 2.3:** E versus y in Laplacian electric field with different tip radii and voltages

it can be resulted from the studies explained that the electric field varies linearly with distance. This is also in accordance with experiments of Lesaint and Top [24], which showed that different distances do not affect PDs significantly according to Equation 2.8.

2.3. Dielectric Liquids

A *dielectric liquid* is a liquid substance characterized by its poor conductivity of electric current. It serves the purpose of insulating electrical components and systems. The historical adoption of petroleum-based mineral oil as an insulating agent in transformers dates back to the introduction of transformer usage. In 1892, "General Electric" initiated the use of the first petroleum-based oils as insulating liquids in transformers [25]. The same reference has given a comprehensive review of the historical use of petroleum-based mineral oil as an insulating liquid in transformers and the recent focus on renewable and biodegradable alternatives.

The environmental impact of nonrenewable mineral oil is highlighted. This has led researchers to direct their attention to renewable and biodegradable alternatives.

The objective of this project is to evaluate the partial discharge (PD) behaviour of different types of transformer oils. To achieve this, two liquids representing different types of commonly used transformer oils were selected. One candidate, *Nytro 10XN*, is a mineral oil. The second candidate belongs to the synthetic ester group, known as *MIDEL 7131*.

Meanwhile, tests have also been conducted on *Cyclohexane*, a well-known stable and nonpolar liquid commonly used as a model fluid with good characteristics for spectroscopy applications.

2.3.1. Cyclohexane

Cyclohexane is a cyclic hydrocarbon with the molecular formula C_6H_{12} . It is a colourless, flammable liquid with a distinctive odour. The molecule consists of a ring of six carbon atoms, each bonded to two hydrogen atoms. Cyclohexane is a type of *cycloalkane*, which is a class of *aliphatic hydrocarbons* containing only single bonds [26].

One of the key characteristics of Cyclohexane is its stability due to its symmetrical structure. The carbon-carbon bonds in the Cyclohexane ring have a tetrahedral geometry, which minimizes the strain in the molecule and makes it relatively stable compared to other cycloalkanes with smaller or larger ring sizes.

Cyclohexane is a nonpolar liquid that does not dissolve well in water but is soluble in many organic solvents. It is commonly used as a solvent in various industrial processes, including the production of pharmaceuticals, plastics, and other chemicals. Additionally, Cyclohexane is often employed in the laboratory for reactions that require a nonpolar solvent.

While Cyclohexane is relatively stable due to its symmetrical structure (see Figure 2.4), its flammability is a property shared by many hydrocarbons. The presence of carbon and hydrogen, combined with the availability of oxygen in the air, allows for the initiation of combustion reactions. This is why hydrocarbons, including Cyclohexane, are considered flammable.

Cyclohexane is commonly used in spectroscopy for several reasons: **Lack of Absorption:** Cyclohexane has minimal absorption in the ultraviolet (UV) and visible regions of the electromagnetic spectrum. This property makes it suitable as a solvent for UV-visible spectroscopy, allowing accurate measurements of samples dissolved in it. **High Transparency:** Cyclohexane is transparent to light in the UV and visible regions, enabling clear transmission of light through the sample. This transparency is crucial for obtaining reliable spectroscopic data. **Chemical Inertness:** Cyclohexane is relatively inert chemically, meaning it does not readily react with most substances. This inertness reduces the likelihood of interfering chemical reactions with the sample, ensuring the integrity of the spectroscopic measurements. **Low Fluorescence:** Cyclohexane exhibits low fluorescence, which can be advantageous for fluorescence spectroscopy experiments. A low fluorescence background minimizes interference with the emission spectra of fluorescent samples. Table 2.3 shows Characteristics of used liquids. More details and datasheet have been given Appendix B.

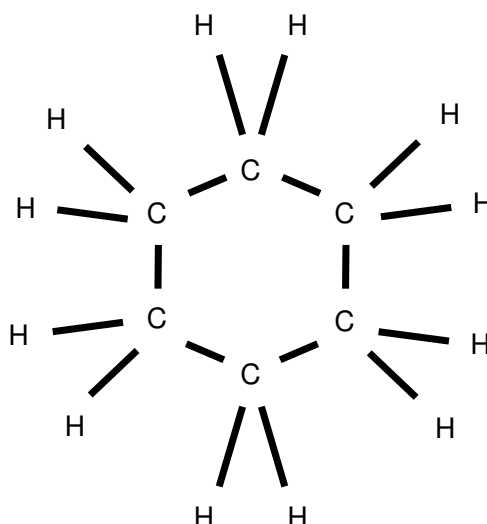


Figure 2.4: Cyclohexane molecule structure implemented from [27]

2.3.2. Nytro 10XN

Nytro 10XN is a naphthenic oil provided by the Swedish company *Nynas*. (see Figure 2.5) A mineral oil commonly employed in transformer insulation due to its transparent and colourless nature, consisting of various hydrocarbons [28]. *Nytro 10XN* is a type of mineral oil specifically designed for use in electrical transformers. Mineral oils like *Nytro 10XN* are commonly used as insulating fluids in transformers to help dissipate heat and prevent arcing between components [29].

Nytro 10XN is an excellent mineral oil that conforms to both *ASTM D3487-16* and *IEC 60296 Edition 5.0* boasts remarkable heat transfer properties owing to its low viscosity and viscosity index [29].

Leveraging naphthenic properties, this oil enables the transformer to initiate operation at exceptionally low temperatures without requiring the use of pour point depressants.

Meeting and surpassing stringent requirements for dielectric strength, this insulating oil delivers optimal performance when stored and handled appropriately. (see datasheet of this liquid in Appendix B)

2.3.3. MIDEL 7131

MIDEL is the world's leading brand of ester transformer fluid. Since the 1970s, *MIDEL* has been used in hundreds of thousands of installations worldwide [31].

MIDEL 7131 is a biodegradable synthetic ester dielectric liquid designed to provide an alternative to mineral oil, silicone liquid and dry-type transformers [32].

It has a relatively high fire point of 316°C , significantly increasing the fire safety of transformers and reducing the need for fire protection equipment [25]. *MIDEL 7131* has an extremely low pour point of -56°C , making it a highly effective solution in colder climates.

MIDEL 7131 has exceptionally high moisture tolerance. This characteristic enables the extension of cellulose insulation life [33]. It is currently used in a wide range of transformer applications and is proven up to 433kV. Perfectly suited for non-free-breathing and free-breathing transformers (due to their excellent oxidation stability) located indoors or outdoors [31].

In Wang's doctoral study in [33], valuable insights were gained regarding the performance of *MIDEL 7131* as an insulating liquid compared to the mineral oil *Gemini X*. A significant portion of the thesis is dedicated to investigating the Partial Discharge Inception Voltage (PDIV) [33].

Employing a Needle-Sphere electrode configuration and incrementally increasing the 50 Hz AC applied voltage by $1\text{kV}_{\text{peak}}/\text{step}$, both the PD repetition rate and the maximum PD amplitude were documented, as depicted in Figure A.1 in Appendix A. The electrode separation distance was 50 mm, with the spherical electrode possessing a diameter of 12.5 mm. Notably, for both *Midel* and *Gemini X*, the

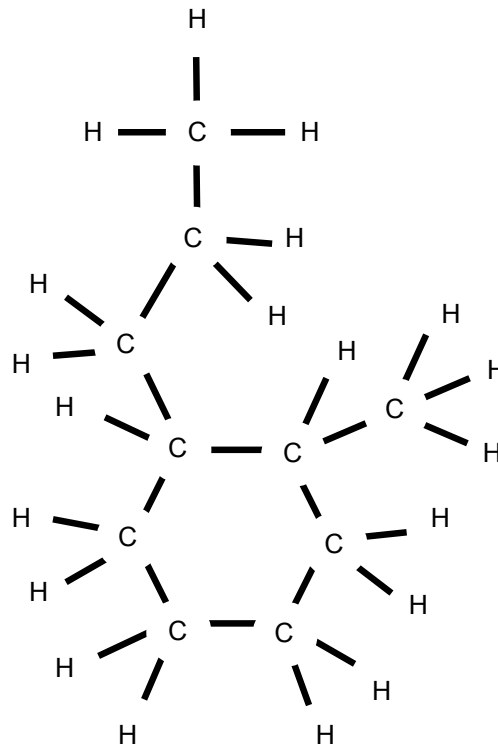


Figure 2.5: Naphthenic oil structure implemented from [30]

Table 2.3: Characteristics of the liquids

Parameter	unit	Cyclohexane	Nyro 10XN	MIDEL 7131
Relative Permittivity	[1]	1.9	2.17	3.17
Conductivity	[pS _{DC} /m]	2.9	0.60	23.61
Density	[kg/m ³]	779	876	970
Viscosity	mm ² /s	1.309@17°C	7.6@40°C	28@40°C

maximum PD amplitude exhibits a similar trend across various voltage levels, demonstrating a linear increase [33, 34]. The datasheet of MIDEL 7131 is given in Appendix B.

2.4. Polarization of the Liquid in the Electric Field

Polarization of the insulating liquids in an electric field is not an instantaneous phenomenon. There are four different mechanisms responsible for polarization: electronic, ionic, molecular orientation, and inter-facial.

Electronic and ionic are relatively faster and instantaneous, but the latter two are slower and will take some time, from seconds to hours [35].

In non-polar liquids, the conduction relaxation time ($\tau = \frac{\epsilon}{\sigma}$) is a measure that shows how slow or fast this process can be [36].

Electronic polarization occurs due to the tendency of electrons orbiting more at one side of the atomic nuclei of the atom. This results in a temporary dipole effect within the atom induced by the electric field. Relative permittivity is a measure of how a material can resist the electric field compared to the vacuum. In other words, relative permittivity defines how much the material can be polarized. The permittivity of

the vacuum is a constant with a value of

$$\epsilon_0 = 8.854 \times 10^{-12} \frac{F}{m} \quad (2.9)$$

There are three phases with real and imaginary parts of the relative permittivity change with frequency that have been shown in Figure 2.6. There are three mechanisms involving these nonlinearities of the real and imaginary parts of the relative permittivity. The three polarization mechanisms involved from high frequency to low include atomic deformation ($\tan \delta_{De}$), then dipole orientation ($\tan \delta_{Or}$) in which the application of an electrical field encourages molecules to align. Dipoles in liquids generally rapidly orient themselves in response to the electrical field. This is while the thermal noises will reduce the dipole orientation [23] and interfacial polarization ($\tan \delta_{Pol}$).

An approximate demonstration of different effects and the change in the real and imaginary parts of the permittivity that is due to electric polarization has been shown in Figure 2.6.

Polarization is not an instantaneous effect, and there is a relaxation time after an AC voltage is applied

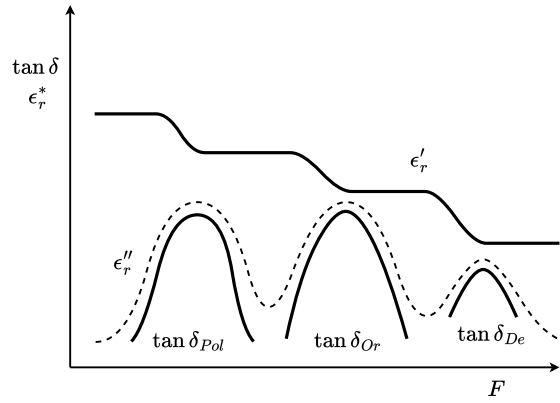


Figure 2.6: Approximate trend of the relative permittivity as a function of frequency

to the insulator for the dipoles to start to form from one direction and disappear from the opposite direction. This is being done with certain inertia due to the relaxation time, and when the change of the AC electric field is faster than the polarization of the material in any of the mechanisms, it results in a nonlinear trend in both components of the relative permittivity.

The connection between the electric field and displacement in an insulating material with a permittivity of ϵ_r^* may be defined as [23]:

$$D^* = D e^{-j\delta} = \epsilon_r^* \epsilon_0 E \quad (2.10)$$

In which the ϵ_r^* is the complex relative permeability, it can be isolated as:

$$\epsilon_r^* = \frac{D \cos \delta}{\epsilon_0 E} - j \cdot \frac{D \sin \delta}{\epsilon_0 E} = \epsilon_r' - j \epsilon_r'' \quad (2.11)$$

where it is common to write

$$\frac{\epsilon_r''}{\epsilon_r'} = \frac{\sin \delta}{\cos \delta} = \tan \delta \quad (2.12)$$

This term has been widely used as the *Loss factor*.

2.5. Electric Discharge Mechanism in Liquids

discharge mechanism in the liquids includes complex interactions between electric fields, charged particles, and liquid molecules.

There are different stages in the discharge mechanism, such as initiation, avalanche ionization, formation of channel, current flow and breakdown, recombination and aftereffect. However, in this study, the main focus is on the initiation stage.

There are pivotal factors influencing electric discharge in liquids, which are:

- Type of the liquid: because different liquids have different ionization energies
- Electrode material: electrode material can affect the initiation and development of the discharge
- Electric field strength: increasing the field strength can increase the chance of ionization and formation of conductive channels
- Temperature: A higher temperature can enhance electron emission and reduce the viscosity of the liquid.
- Impurities and bubbles: The presence of impurities or gas bubbles in the liquid can provide nucleation sites for discharge

In some liquids, which primarily contain dissolved ions or electrolyte solutions like impure water, for example, upon the application of a voltage across the solution, positively charged ions (cations) migrate towards the negative electrode (cathode) and vice versa. This motion of ions causes the flow of electric current.

In metals and other conductive liquids, like mercury, free electrons are the principal carriers of charge. Meanwhile, in non-polar insulator liquids, partial discharges and, subsequently, complete discharges occur primarily due to the presence of charge carriers¹. This region always present in gases, However it might not present in dielectric liquids [37]. Without these charge carriers, no discharge would take place in the liquid. A question might arise: What are these charge carriers, and what is their source?

2.5.1. Ionisation

Ionization mechanism in insulating liquids refers to the process by which molecules within the liquid break apart into atoms, ions, or particles. Ionization typically occurs in polar insulating liquids, where the molecules possess an uneven distribution of electrical charge, resulting in a dipole moment. When subjected to an electric field or other forms of energy, these polar molecules may separate into positively charged ions (cations) and negatively charged ions (anions).

2.5.2. Electron Emission Mechanism

Electron emission mechanism refers to a phenomenon where electrons are emitted from the surface of a material when subjected to certain conditions. Here are the key emission effects. Thermionic emission happens when a material is heated to a high temperature. Electrons gain enough energy to overcome the work function (the energy barrier) and escape from the surface. When the same mechanism happens by gaining energy from the electric field, it is known as *field emission*. In a strong electric field, The electrons can gain enough energy to leave the cathode and push into liquid. The required field strength is often in the order of $10^7 - 10^8$ V/cm [38].

There are also other forms of electron emission, such as photoelectric emission and secondary electron emission. Figure 2.7 shows the field emission and field ionisation happening due to the tunnelling effect.

In [39], a model of the transformer liquid with the capability of showing the electron emission mechanism in an electrical field has been developed by using the tunnelling theory in COMSOL. the test sample used is a needle plain with $1mm$ gap. Electron density has been investigated in the gap in positive and negative voltage.

Tunneling Effect

As explained in the electron emission mechanism, there is an energy barrier for the electrons inside the contacts to release into the liquid, known as the work function.

Surprisingly, some electrons are emitted into the liquid, or the molecules are ionized by the field-emitting electrode to the anode without having enough energy field. This phenomenon is known as *Tunneling effect*.

The tunnelling effect is a phenomenon in quantum mechanics where particles, such as electrons, can pass through potential energy barriers even though classical physics would suggest they don't have enough energy to do so.

Classically, if a particle encounters a barrier that is higher than its energy level, it cannot pass through it and would be reflected back.

¹charge carriers are moving charged ions and free electrons in a liquid

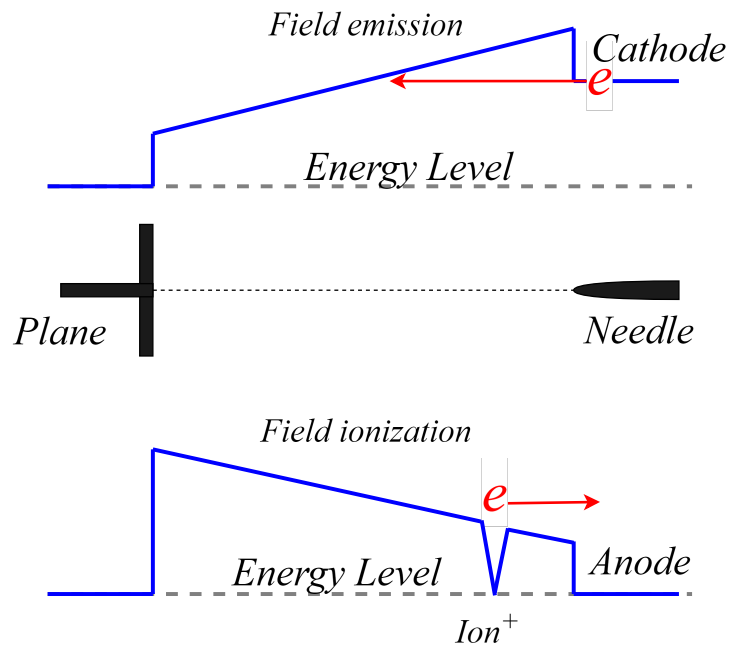


Figure 2.7: field emission and field ionisation happen due to tunnelling effect implemented from [23]

However, according to quantum mechanics, there is a finite probability that the particle can tunnel through the barrier, appearing on the other side without having enough energy to surmount the barrier as classically expected.

Blocking Effect

In the needle-plane geometry, the *blocking effect* refers to the phenomenon in which the strong electric field at the needle tip causes the creation and accumulation of space charges in the region close to the needle.

These charges block the influence of the plane, and field distribution is highly localized around the needle tip. (see Figure 2.8) The blocking effect covers the electrode with a thin charged layer. Findings in [40] show that the blocking effect strongly impacts the current flow.

At higher electric fields, the conductivity is affected by ionization in the liquid. Poole-Frenkel and hopping model [41, 42] and thermionic (Schottky) injection and tunnelling (Fowler-Nordheim) mechanisms of the electrons jumping between different energy levels. These are theories explaining the mechanism of space charge forming in the liquid in the presence of a High electric field [43]. Electronic transport in organic and polymer devices, primarily light-emitting diodes. Thermal processes such as thermionic and field emission are described in [42],

2.6. Space Charges

Charge carriers within a dielectric liquid are named *Space Charges*. This can occur in vacuum tubes, semiconductors, or insulating materials. They play a crucial role in the discharge process. The established mechanisms for generating these charge carriers are named in [44] as:

- Electrons are emitted from the cathode by the electron emission.
- Electrons and positive ions are formed by ionization.

These mechanisms invariably result in residual charged particles remaining within a liquid. The phenomenon is mitigated by additional processes like *recombination*. These two opposing processes es-

establish a balance of charge carriers within the liquid, where particles are in constant motion, either parallel or opposite to each other, ensuring an overall constant charge.

The quantity of these carriers may vary with temperature or electric field strength changes.

The mobility of charge carriers in a dielectric liquid is not only an intrinsic parameter of the conduction process but also very important for explaining the pre-breakdown process. The charge carriers' mobility also affects the electrostatic behaviour of fine and ultra-fine (micrometre and sub-micrometre) floated particles in the fluid [45].

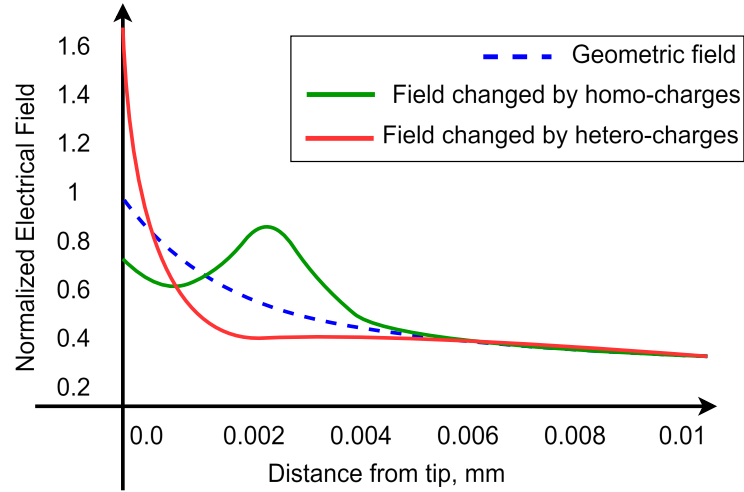


Figure 2.8: Effect of homocharges and heterocharges on the electric field locally. Implemented from [23]

2.6.1. Space Charge Limited Current

Due to the low mobility of the ions in an insulating liquid, the current is limited by the movement of the charge carriers, which are moving slower than charge carriers in gases.

2.6.2. Shokley-Ramo Theory

The Shockley–Ramo theorem is a method for calculating the electric current induced by a charge moving in the vicinity of an electrode [46, 47]. The Shockley-Ramo theorem states that the current flowing out of an electrode because of charge carriers is as follows:

$$i = E_{\nu} \cdot q \cdot \nu \quad (2.13)$$

where q is the amount of charge of the charge carrier, E_{ν} is the electric field in the direction of movement, and ν is the instantaneous velocity.

2.6.3. Child's Law

Child's law was proposed by Clement D. Child in 1911 [48]. This law states that the current density due to space charge limited current (SCLC) between two parallel plane contacts of a vacuum chamber can be obtained by

$$J = \frac{4\epsilon_0}{9} \sqrt{\frac{2e}{m_e}} \frac{v^{\frac{3}{2}}}{d^2} \quad (2.14)$$

In this case, J is the current density, d is the distance between the cathode and anode, e is the amount of the charge of an electron, and m_e is the mass of an electron.

2.6.4. Drift Regime (Mott-Gurney's Law)

A more general form of the Child's Law that gives the current density between two plane contacts with an insulating material as a medium is Mott-Gurney's law [49, 50].

$$J = \frac{9}{8} \epsilon \mu \frac{V^2}{d^3} \quad (2.15)$$

Where J is the current density, L is the thickness between two contacts, $\epsilon = \epsilon_0 \cdot \epsilon_r$ is the permittivity, V is the applied voltage, and d is the distance between contacts. μ is defined as electrical mobility or charge carrier mobility in Mott-Gurney's law. Frenkel's model for fixing the reduced form of Mott-Gurney's law is accessible at [50].

Electrical mobility (μ) is related to the strength of the field (E) and drift velocity (ν_d), which is defined as:

$$\mu = \frac{\nu_d}{E} \quad (2.16)$$

A charged particle in gas or liquid will accelerate in a uniform electric field until it reaches a certain speed. This speed is defined as drift velocity.

Electrical mobility is proportional to the net charge of the particle. It is inversely proportional to the Stokes radius of the ion. Considering a charge density of ρ_0 of the residual charge carriers in the region². From Shockley-Ramo theory, The current density of charges with a charge density of ρ_0 of residual ions is as follows [23]:

$$j_0 = \rho_0 \mu \frac{V}{d} \quad (2.17)$$

By comparing two equations Equation 2.17 and Equation 2.15 a condition can be resulted as:

$$\frac{d^2}{V} > \frac{\epsilon \mu}{\sigma} \quad (2.18)$$

In which σ is the conductivity of the liquid. this inequality shows a condition in which the SCLC is limiting the flow of charges between contacts. Therefore, the relation between the electric field strength and conductive current³ changes from linear to nonlinear when this condition applies.

It is important to note that this inequality is based on the plane-plane geometry. However, it can give a good perspective on the relation of conductivity with space charge limited current.

The current density resulting from the drifting of charged particles can be calculated from the mobility.

$$J = en\mu E \quad (2.19)$$

In which n is the number of charged particles, Ohm's law can be written in the form of

$$J = \sigma E \quad (2.20)$$

where conductivity is

$$\sigma = en\mu \quad (2.21)$$

2.6.5. Influence of Space Charges on the Electric Field

The electric field is not only affecting the movement and presence of space charges in the gap, but space charges also affect the electric field locally [51].

The maximum electrical field at the needle tip is enhanced in the presence of heterocharges and reduced slightly by homocharges.

The following formula has been suggested by Atten [52] for the point-plane geometry in the presence of space charges:

$$E_{th} = \frac{2V}{r_0 \ln\left(\frac{4d}{r_0}\right)} - \left(\frac{\eta I}{8\pi\epsilon^2}\right)^{0.5} \int_0^d \frac{g(x)}{\sqrt{E(x)}} dx \quad (2.22)$$

In this equation, the influence of the charge $Q(x)$ has been defined as $g(x)$, and η expresses the dynamic viscosity. There is no practical way to measure $g(x)$; therefore, the electric field's quantitative evaluation is impossible.

²The active region with extremely high electric field around the needle tip

³Conductive current: The AC component of the current waveform which is in phase with voltage

2.7. Conductive Current in Dielectric Liquid

One of the significant phenomena resulting from the motion of space charges in a dielectric liquid is conductive current. However, the initial step in understanding this is knowing the conductivity in dielectric liquids.

The current flowing between two contacts connected to a high-voltage AC is almost always leading in phase. If it is modelled by a capacitance, which is not far from the nature of an insulator, then based on the polarization theories in section 2.4, the complex capacitance of the test liquid and loss factor relation with capacitive and conductive current values are as follows:

$$\begin{aligned}\bar{C} &= (\epsilon'_r - j\epsilon''_r) \cdot C_0 \\ \tan(\delta) &= \frac{1}{R\omega\epsilon'_r C_0} + \frac{\epsilon''_r}{\epsilon'_r} \\ \tan(\delta) &= \frac{I_r}{I_c}\end{aligned}\quad (2.23)$$

The capacitive component of the current (I_c) is due to different mechanisms of polarization. The resistive or conductive component of the current (I_r) is primarily associated with space charges and SCLC within the liquid. In other words, the conductive part of the current, which is in phase with voltage, is markedly influenced by both the number of space charges and their mobility.

The general form of the conductivity characteristics for gases and dielectric liquids for a plane-plane electrode geometry may be presented as in Figure 2.9 according to the molecular configuration of liquids [37].

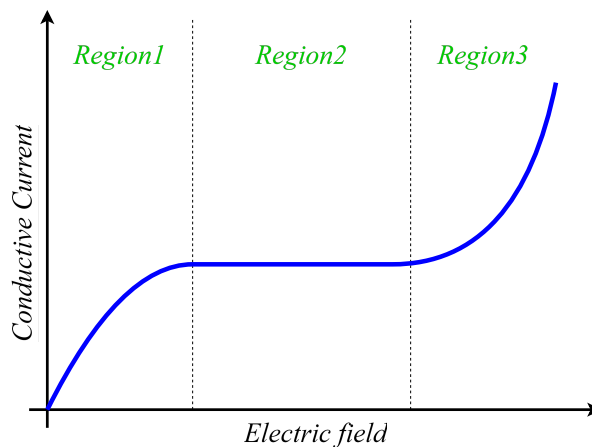


Figure 2.9: General relationship between current and field strength for liquid dielectrics. Implemented from [35]

It comprises three regions. Region 1 in which the liquid primarily exhibits Ohmic behaviour. At low electric fields, the conductive current has a linear relation with the electric field.

Region 2 is around 1 kV/cm to 10 kV/cm. Another name for this region is a subohmic region. It demonstrates that most ions in the insulator have been depleted due to electric field stress. However, this stress isn't sufficient to cause ionization in the insulator. Thus, there are no additional charge carriers formed. As a result, the current is saturated by increasing the voltage stress.

A source of charge carriers must exist for a continuous current to flow through the liquid. In the ohmic and subohmic regions, natural ionization of ionic impurities (uncontrolled chemical impurities) can be presented as the charge carriers [37]. The simplest assumption of these impurities is to consider them as ion pairs moving freely inside the liquid.

In Region 3, a new source of charge carriers emerges in the insulator liquid. This corresponds to ionization, occurring at higher voltage stresses, and is characterized by an exponential relation. At higher field stress, the current increases rapidly and at 40 – 100 kV/cm, a breakdown may occur [37].

There are two standards used for measuring conductivity in liquids for electrotechnical purposes. IEC 61620 recommends applying a low-frequency square voltage, resulting in a field below 0.1 kV/mm. On the other hand, IEC 60247 [16, 53] utilizes an AC voltage, leading to an applied field ranging from 0.05 to 0.25 kV/mm.

In [23], different liquids have been tested under sinusoidal AC high voltage (below PDIV) with low frequency (0.1 Hz) and a pulse waveform with 50 μ s duration. The results for the AC voltage tension show different behaviour between the positive and negative sides of the cycle, which has been explained to be because of the lower barrier for injection of electrons other than an injection of holes [23]. It also has been explained that this level of conductivity, in theory, leads to a non-physical electric field to obtain from $j = \sigma \cdot E$; therefore, there must be homocharges in the vicinity of the tip to lower down the field to a lower value.

Distinguishing between bulk ionization and electrode emission phenomena can be challenging, especially in higher fields. Nonetheless, experimental observations indicate that at low fields, bulk liquid ionization due to Thomson-Onsager phenomena is dominating. On the other hand, for higher fields, the main source of current carriers is due to ion injection from the interfaces [54].

Low Field

By increasing the voltage, charge carriers move relatively to the applied field. In a liquid at thermodynamic equilibrium, conductivity depends on the number of charge carriers [23]. The ohmic behaviour is up until the ion transport time exceeds the relaxation time. (ionic relaxation time which is $\tau = \frac{\epsilon}{\sigma}$).

After the whole charge carriers have been swept out of the bulk volume, the current will no longer increase proportionally to the voltage [23], and there will be a nonlinear saturation trend between conductive current and applied electric field (Region 2 in Figure 2.9. Accumulation of these charges around the one contact additionally, affects the electric field, lowering it down than the Laplacian field. (see subsection 2.6.5)

The current density in the plane-plane geometry can be obtained by the following relation

$$\epsilon_r \epsilon_0 \frac{dE}{dx} = \rho(x) \quad (2.24)$$

In which ρ is the charge density in 1 dimension, E is the electric field. Considering Mott-Gurney's law (Equation 2.15), conductive current can be described as [54] :

$$I^{0.5} = K(V - V_0) \quad (2.25)$$

$$K \approx 1.06 \sqrt{\frac{\epsilon \mu}{d}} \quad (2.26)$$

$$V_0 = 0.5 E_p r_p \ln\left(\frac{4d}{r_p}\right) \quad (2.27)$$

Where V is the applied voltage and V_0 is the threshold voltage, While ϵ and μ are permittivity and mobility, and E is the electric field of the needle. d is the needle-plane distance. There is no better model available for the needle-plane geometry currently to predict the electric current in high electric fields ($> 5 \cdot 10^4$ V/mm)

High Field

The conductivity at the higher fields has been observed to increase exponentially with the electric field. This is due to the formation of new space charges in the region. Results from the impulse voltage show a linear increase of charges during the voltage pulse, suggesting the high rate on initial charge carriers followed by drift; however, it's not the case for all of the liquids in all polarities.

A threshold has also been obtained for the space charge limited current. The threshold is likely to be

linked with charge injection [23]. In another way, for the mobility of the charge carriers, the calculated mobility speed for the sinusoidal voltage has been lower than the calculated one for the pulse voltage by an order of magnitude.

It has been explained in [23] that under AC, there is a slow formation of local charges during a long sequence, whereas for the pulse voltage, this charge formation is counter-balanced with the charge drift that is established momentarily. This has been the main cause of the higher conductivity in a specific electric field over time.

Even though the microscopic electric field is below what is expected for the electron process, the local field at the molecular level can be higher than the Laplacian field [23]. The exact field is also dependent on the molecular orientation in the electric field [55, 56].

2.8. Partial Discharge

As defined by IEC 60270, a *Partial Discharge* (PD) is a localized dielectric breakdown of a portion of a solid, liquid or gaseous electrical insulation system under high voltage (HV) stress, which may or may not occur adjacent to a conductor [57].

Major insulation breakdowns in gaseous, liquid or solid insulation are usually initiated by partial discharges. Thus, knowing the partial discharge phenomena gives a lot of information about assessing the dielectric strength of the liquid [44]. One theory about PD formation is the concentration of charges, which increases at one point, causing an increase in temperature. This high temperature may cause *micro-bubbles* or electron avalanches to form within the insulating, which is a mechanism suggested by J. Lehr. It is more likely that PD starts within these micro-bubbles [58].

However, new findings also strengthen the second theory that the electron avalanche formation is prior to micro-bubble creation, as the previous theory can not explain the large difference between positive and negative polarity.

PDs are inherently a stochastic phenomenon. Thus, it is important to consider the PD probability distribution. Given their random nature, individual PD events lack straightforward fixed parameters for prediction. It is notable that numerous small negative PDs exist alongside a limited number of large positive PDs. This observation indicates an inherent asymmetry in the PD phenomenon [23].

2.9. Electrical Measurement of the PDs

The electrical circuit for detecting PD functions is based on a fundamental principle known as the ABC-equivalent model [34] or PD measurements based on charge acquisition. In this model, a void within the insulation is represented by a small capacitance, C_c , connected in series with a larger capacitance, C_b , symbolizing the insulation between the void walls and the electrodes. The remaining insulation is represented by C_c , which exhibits the highest capacitance. This ABC-equivalent circuit is illustrated in Figure 2.10.

Through voltage division, the voltage across the void (V_c) follows the applied voltage but is reduced by a factor determined by $\frac{C_b}{(C_b+C_c)}$. If the applied voltage surpasses the withstand capability of the void, a local transient current flows between its walls, marking the onset of partial discharge, known as the *partial discharge inception voltage* (PDIV). The voltage corresponding to this event over the void is termed the ignition voltage. Upon discharge, V_c rapidly drops to nearly zero. To counteract this voltage drop, the external circuit delivers a transient current to the test object, denoted as I in Figure 2.10. Typically, this charge injection is primarily supplied by a relatively large *coupling capacitor* (C_x) connected in parallel with the test object. The charge carried by I , known as the apparent charge, can be detected by placing a suitable *measuring impedance* (Z_M) in series with the coupling capacitor. It is an apparent charge because it doesn't directly represent the actual discharge current within the void but is proportionate to it and can be calculated through proper calibration [59, 60].

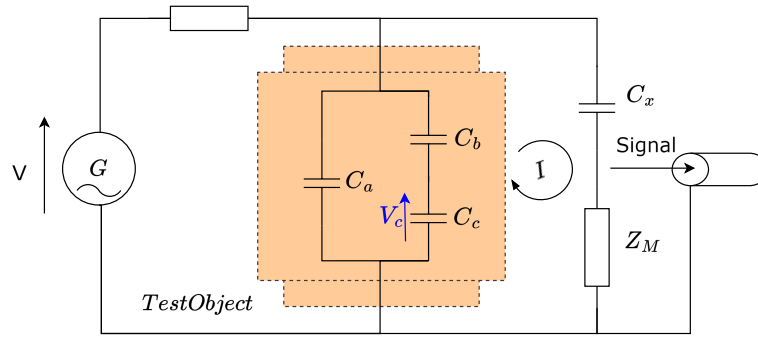


Figure 2.10: the ABC-equivalent circuit

2.10. Definition of PDIV

The PDIV concept, or Partial Discharge Inception Voltage, isn't a straightforward threshold because it also depends on time, with several definitions existing. These definitions also vary depending on whether the measurement is conducted under DC or AC voltage regimes. Experiments in [61] demonstrate that PDIV voltages differ based on the criteria used, making it challenging to establish a single meaningful threshold for assessing the quality of insulating liquids. To address this, a specific definition for PDIV testing is necessary.

A simple definition of partial discharge inception voltage can be the voltage at which partial discharge starts. However, since partial discharge is a random phenomenon, it is related to the chances of happening at some voltage level. In other words, the chance of a partial discharge incident increases by an increment of the voltage, but there is no exact point for starting the partial discharges. Giving more time to the insulator liquid under test will also increase the chance of a PD incident happening. Therefore, a clear definition of PDIV voltage should be taken. It should be accurate and repeatable.

In the previous works in the master thesis by Skirbekk [62]. Her definition of the PDIV was the voltage step in which the first PD happens. The voltage has been increased stepwise by $V_{peak} = 1\text{kV}$ per step, and the length of each step was 1 minute.

In another study by Fagerli [34], PDIV has been defined as a chance of a PD incident by considering the PD repetition rate versus voltage.

However, the IEC has defined the PDIV differently, which is explained in subsection 2.1.1.

2.11. Needle Tip Erosion Over Time

The tip radius of the needle is not constant when it is being tested under HV tension. Previous studies show that the needle becomes dull over time, and, therefore, a time-dependent electrical field is created between the needle-plain area [23].

T.G.Aakre has quantified the rate of change of the needle tip radius over time for a tungsten needle at $d = 20\text{mm}$ in MIDEL 7131 under 18kV voltage as a function of time by fitting a cubic root function in the form of Equation 2.28 to the test results.

$$r_p(t) = \sqrt[3]{\frac{3A}{2\pi}t} \quad (2.28)$$

And the calculated value of the constant in [23] is:

$$A \approx 400 \frac{\mu\text{m}^3}{\text{h}} \quad (2.29)$$

2.12. Light Emission From PD

Energy is released in electron avalanches during the partial discharge. This energy can transform into the form of heat, light, and sound. The heat release during the PD has been shown by Lesaint et al.

[63]. Lundgaard et al. showed that streamers also emit light [64]. The source of light emission from PD is the recombination process. However, the light spectrum of different types of discharge is not the same [65]. In addition to that, The spectrum of surface discharges along a solid dielectric is more complex and influenced by many factors such as solid material, surface condition, and the composition of material [66]. Transformer oil exhibits emission in the range from 350 nm to 700 nm, depending on oil composition. (Figure 2.11) The emitted light's frequency is predominantly formed by hydrogen and hydrocarbon mixtures such as Methane, Ethane, and Ethyl [65].

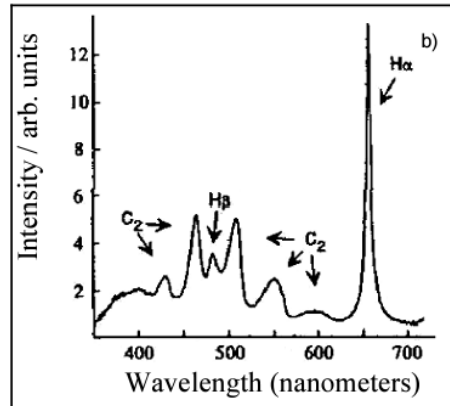


Figure 2.11: Typical emission spectra, discharge in liquid. Adopted from [65]

A small fraction of recombination energy ($\approx 1\%$) is emitted as light in gasses; for liquids and solids, the fraction is even smaller [67]. The amount of the emitted light and its frequency depends on the type of the material (gas, liquid) and other factors like temperature or pressure. Therefore, the spectrum of the emitted light from the PD extends from ultraviolet over the visible range to infrared [65]. Different optical PD detectors can be used depending on the wavelength and the strength of the light emission. Common optical systems for this purpose are UV corona scopes, night vision cameras, low-light enhancers, photo-diodes, and photo-tubes.

The categories of detection are also important: imaging, quantitative, or non-imaging detection sensors. In [65], M. Muhr used two approaches for measuring the light emission from PD of a needle-plate configuration in transformer oil. The voltage range is from 0 kV to 100 kV, and the distance between the needle and plane is adjustable in the range of 1 – 20 cm. The oil is at the normal pressure, and the experimental setup is in a darkened and shielded HV chamber. Experiments have been done by two arrangements. The first arrangement is called Fluorescent optical fibre, in which the needle is positioned perpendicular to a fluorescent optical fibre, which is attached to a conventional optical fibre and then to the sensor. In the other arrangement, which is called a lens, a lens is attached to the end of a conventional optical fibre, and the needle is placed in front of the lens at a distance called the observation area. This PD detection method is tested using air as a medium and compared with the conventional method, which has been proven to be correct. After that, it was used to measure the PD in oil. The detected discharges in oil have been reported to be scattered in amplitude and shape, and positive streamers have shown a superposition of fast pulses of growing intensity. The obtained results are shown in Figure A.2 in Appendix A.

The results show that the electrical PD measuring system is not fast enough to show the fast PD impulses happening in oil due to limited bandwidth.

Schwartz et al. also show that different spectral regions of the radiated light are absorbed from different magnitudes of PDs. (Figure A.3 in Appendix A)

3

Methodology

3.1. Measurement Setup

The nonuniform electric field can be found in electrical installation. The electric field can increase intensely at the sharp edges of the busbar and sharp conductors at the HV side in transformers. The localized strong electric field facilitates the localization of PD events. Therefore, The needle-plane configuration has been selected to produce a nonuniform electric field.

The needle-plane setup offers a known electrical field distribution, making it preferable for studying PD mechanisms. However, in the IEC 61294 standard, a needle-sphere geometry has been suggested for PD inspection, which is explained in subsection 2.1.1.

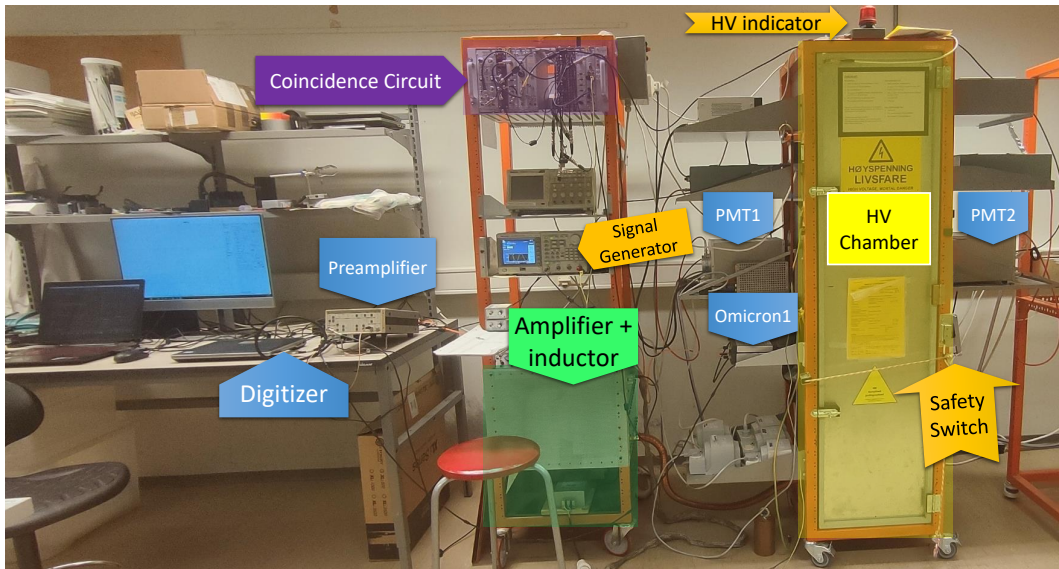


Figure 3.1: The photo of the test setup with placement of different parts

Furthermore, The measurement instruments were susceptible to electrical interference by the electric wiring system, but previous studies helped for proper shielding and filtering of the PD signal [23]. Therefore, the test cell is positioned within a Faraday cage to adequately shield it from external influences. A sinusoidal AC voltage of up to 22 kV is applied to the plate, and the needle has been grounded. Figure 3.1 demonstrates the full setup arrangement and Figure 3.2 shows the inside of the HV chamber. The equivalent electrical circuit of the test setup has been shown in Figure 3.4.

3.1.1. Setup Description and Safety Precautions

The test setup comprises an enclosed chamber housing a step-up transformer at its lower section. (High voltage source in Figure 3.2) Using a chamber to isolate the complete HV parts minimized electromagnetic noise and light interference.

All corners of the chamber have been sealed with conductive tape, and all incoming cables used grounding shields, but making current loops has been avoided.

Inside the HV chamber in Figure 3.2, the test cell (test sample) can be conveniently mounted and dismounted on the upper side of the chamber. The test cell (test sample) has been placed on an insulated plastic mount. The lower side of the test cell has been connected to the capacitors C_1 and C_2 , which are connected to the electrical PD measurement device using an ABC circuit (see section 2.9 and section 3.2). It has been shown in Figure 2.10 as C_x .

There is also a manual grounding device for safe visible grounding while working in the HV chamber. There are also two micro switches and a semi-automatic grounding system attached to different parts of the door of the HV chamber to automatically disconnect the high voltage feed and connect the high voltage arm to the ground when opening the door.

The entire high-voltage connection inside the HV chamber has been made by using round brace rods with $r_b = 1\text{cm}$ that are screwed together for the least possible PD. A shielded coaxial cable connects the

needle to the preamplifier through a safety gas fuse (EMP protector) and two voltage-sensitive diodes (protection diode) to limit the voltage on the measuring side in case of a breakdown inside the test cell. (see Figure 3.3)

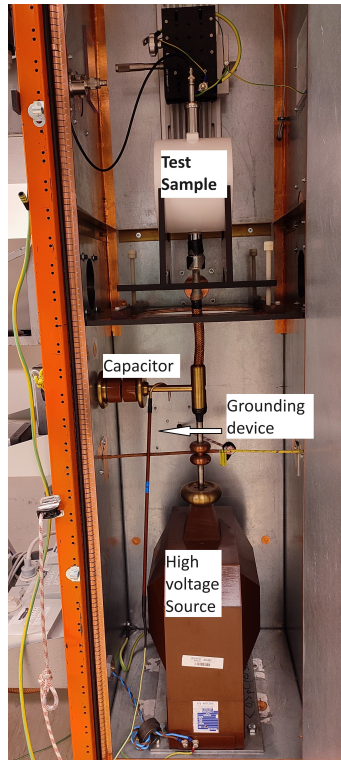


Figure 3.2: Inside HV chamber of the test setup

3.1.2. AC Voltage Circuit

The *Tektronix AFG3052C* signal generator is utilized to generate a pure sine wave with the desired frequency and amplitude for the test. This pure sine wave is directed to a voltage amplifier, and the output feeds an AC transformer.

This transformer has been labelled in Figure 3.3 as an Isolation transformer because it galvanically isolates between the two parts of the circuit. Subsequently, it is supplied to a single-phase step-up HV transformer. (see Figure 3.1 and Figure 3.3)

This section of the circuit operates primarily based on resonance between the capacitance of the high-voltage segment and the inductors added to the primary winding of the final step-up transformer. (see Figure 3.3)

3.2. Discharge Acquisition System

The discharge acquisition system is a system based on an ABC circuit, which has been explained in section 2.9. This system comprises two optically connected parts. The used device has different modules such as *MPD600*, *MPP600* and *MCU502*, *CPL542* along with two known capacitors $C_1 = 1700\text{pF}$ and $C_2 = 1700\text{pF}$ which are part of a capacitive voltage divider with the capacitance inside *CPL542*. Characteristics of the *CPL542* are given in the Table 3.1. This system quantifies current and calculates charge through numerical integration. It is specifically optimized to provide precise measurements of the number and magnitude of partial discharges.

The device enables data storage and the generation of histograms depicting the magnitude and phase of PDs over extended time periods. It is known for its $q(v)$ measuring for the PDIV. The Omicron PD measuring system diagram from its user manual has been illustrated in Figure 3.5. Two discharge

acquisition systems have been used during experiments, which are named *Omicron1* and *Omicron2*; therefore, these names have been used for the discharge acquisition system in the next parts.

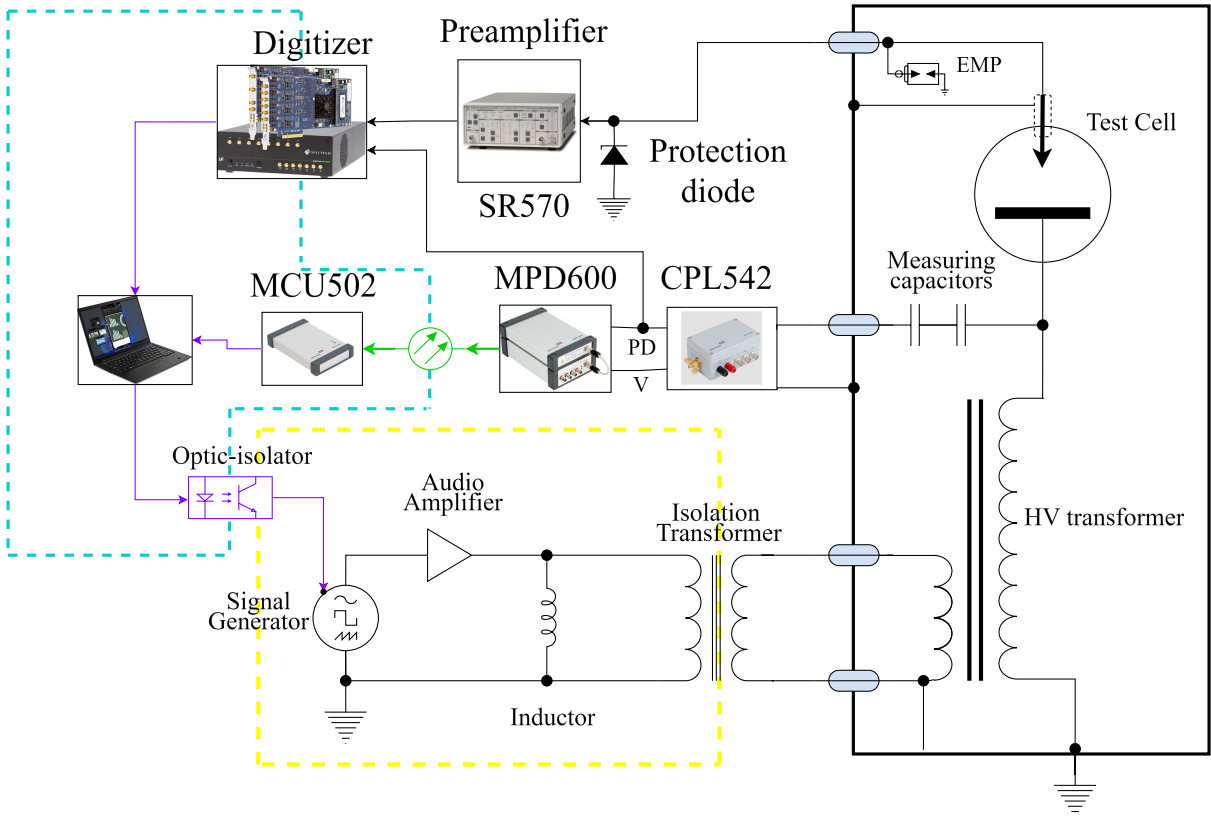


Figure 3.3: Test setup diagram for AC test setup with charge acquisition and current measurement

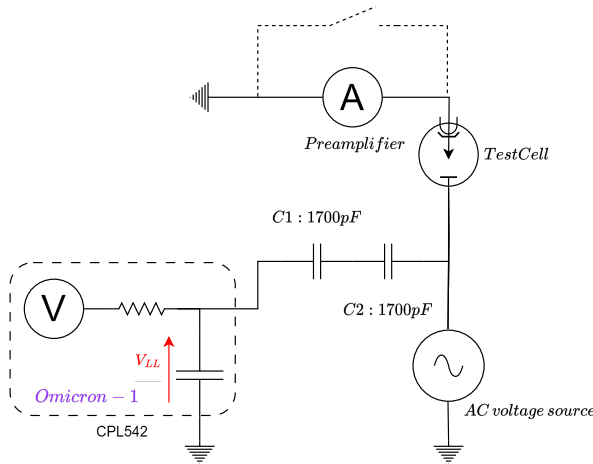


Figure 3.4: Equivalent circuit of the test setup

The expected low-arm voltage has been calculated based on the given information in Table 3.1. Since the highest possible voltage on the low arm is important, The lowest capacitance has been taken into account.

$$V_L = V_{AC} \cdot \frac{C_{measure}}{C_{CPL542} + C_{measure}} \quad (3.1)$$

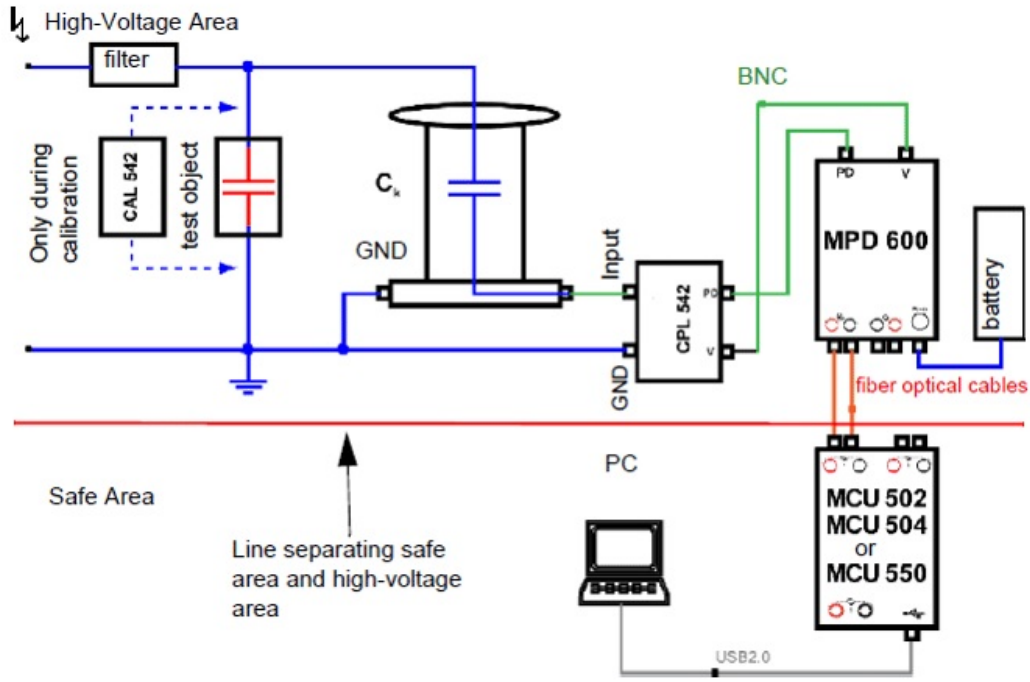


Figure 3.5: Electrical measurement diagram of omicron PD measurement system adopted from [68]

Considering the Highest peak voltage for the test and smallest capacitance available for CPL542, the low voltage side of the capacitive voltage divider is calculated in Equation 3.2

$$V_L = 25000 \cdot \frac{850 \times 10^{-12}}{850 \times 10^{-12} + 30 \times 10^{-6}} = 708\text{mV} \quad (3.2)$$

This voltage is in the range for the CPL542.

The diagram of the rectified AC test setup with two omicron measuring devices and current measurements has been shown in Figure 3.8.

Table 3.1: CPL542 Types and characteristics

MAX Current	0.5A	2A	5A
Low-arm capacitance	30 μF	120 μF	272 μF

3.2.1. Calibration of the Omicron

For correct PD measurement with Omicron, calibration is necessary. a calibration charge injector shown in Figure 3.6 has been used for this purpose. The trigger source has been set to $f = 50\text{Hz}$. then the amount of charge has been set to $q = 20\text{p}$ and the adjacent frame in the Omicron software under Q-tab>calibration settings>QIEC(target) has been set to the same value. The calibrator contacts have been connected between the HV part of the circuit and the ground of the HV. During calibration, the safety grounding connection has been temporarily disconnected from the HV part.

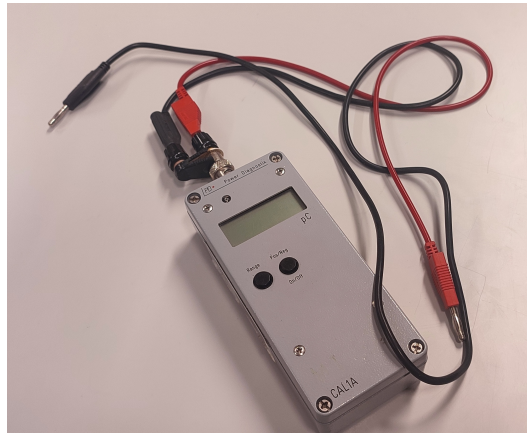


Figure 3.6: Calibrating charge injector

3.2.2. Rectified Half-cycle Voltage Source

To have a deeper understanding of the nature of space charges and test the polarity of the space charges, some tests have also been taken with half-cycle high-voltage AC. Some modifications have been made to the test setup to make the new waveform. Waveform has been mainly rectified by an HV diode. A resistor has also been added to it to control the output voltage of the diode. After the diode, another set of Omicron measurements has been added to the upper side of the high voltage. The equivalent circuit of the modified test setup has been shown in Figure 3.7.

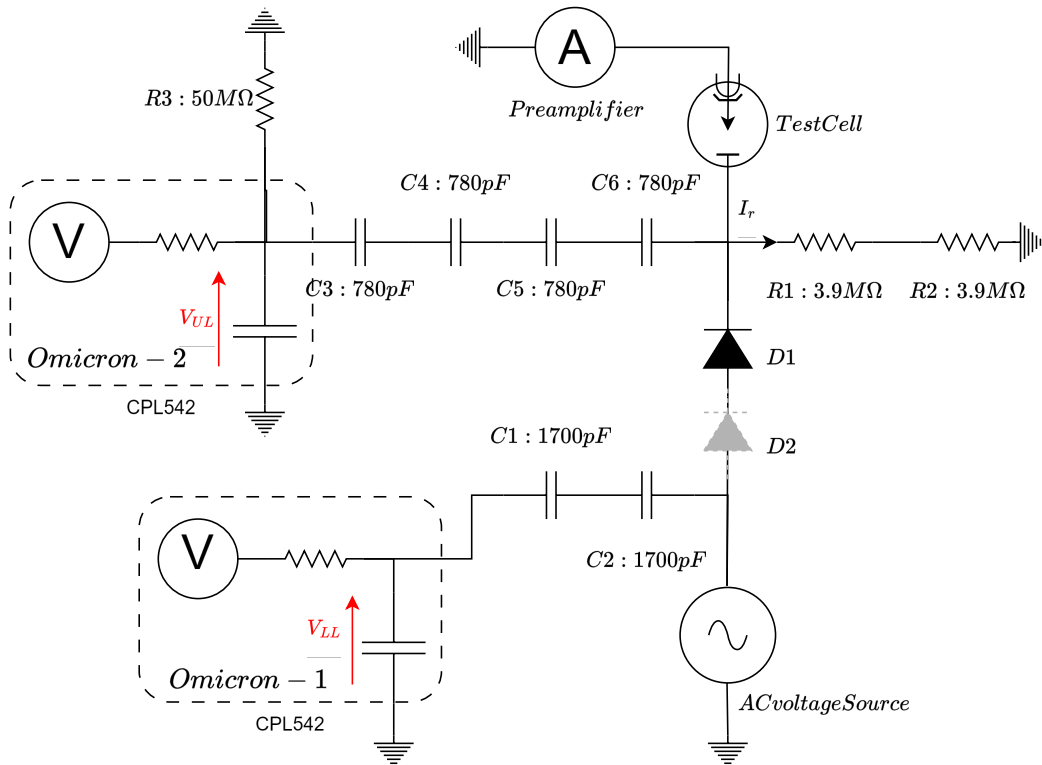


Figure 3.7: Equivalent circuit of the test setup for half-cycle test

The characteristics of the components shown in Figure 3.7 have been given in Table 3.2. The voltage at the Omicron side of the capacitive divider in the upper Omicron device has been calcu-

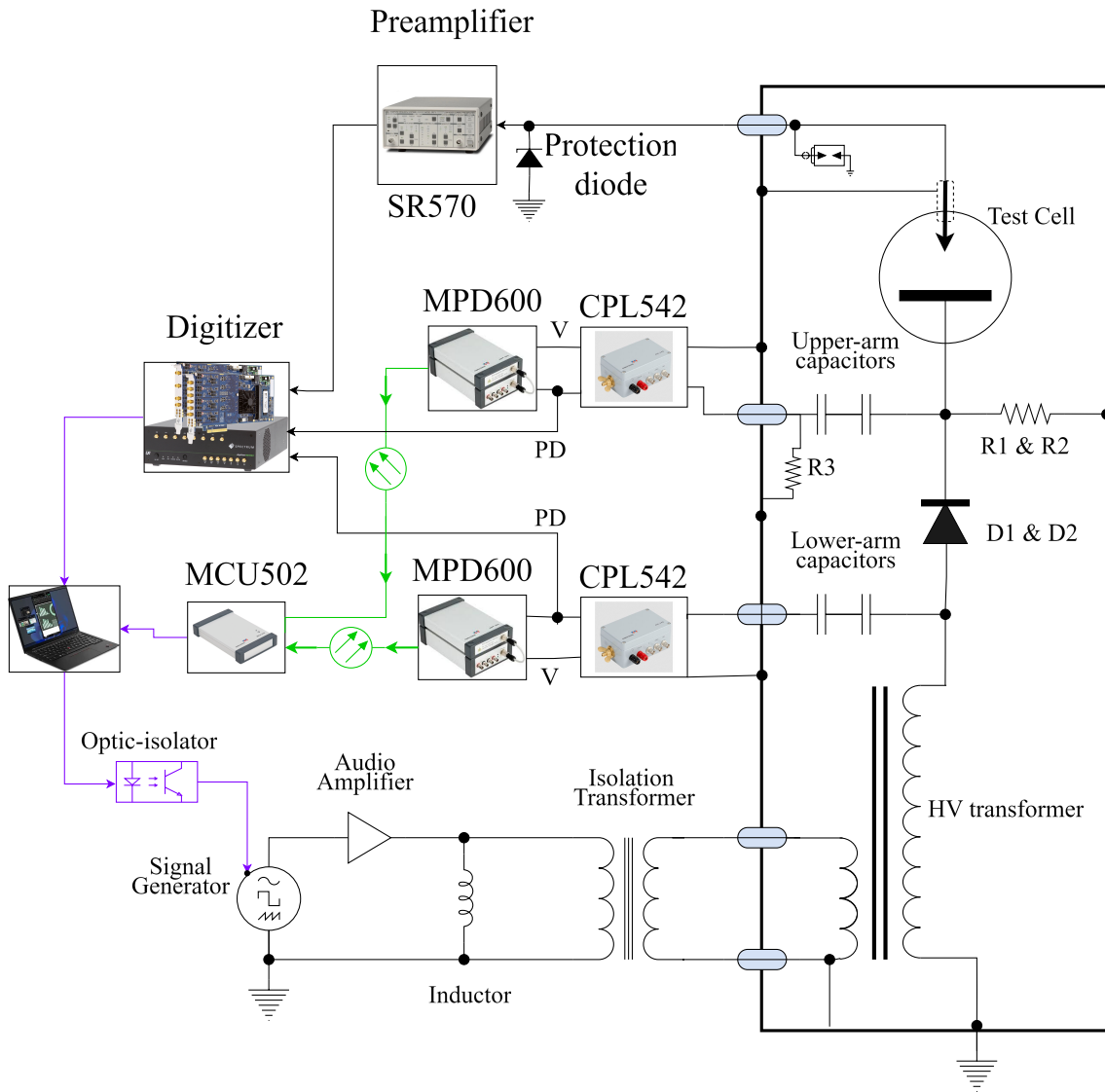


Figure 3.8: Test setup diagram for rectified AC test setup with two omicron devices and current measurement

lated for the worse case in Equation 3.3

$$V_L = 25000 \cdot \frac{780 \times 10^{-12} \times \frac{1}{4}}{780 \times 10^{-12} \times \frac{1}{4} + 30 \times 10^{-6}} = 162.5\text{mV} \quad (3.3)$$

Therefore, the voltage is in the range for CPL542 in the upper arm. because the voltage of the lower side of the voltage divider is relatively small; therefore, it might be vulnerable to noise [38]. The next important quantity to be calculated is the max current through $R1$ and $R2$ and dissipated power. Therefore, the half-cycle's root mean square voltage needs to be calculated to obtain power.

Table 3.2: Details of the elements in Figure 3.7

#	ID	Component	Name	Characteristics
1	D1,D2	High Voltage Diode	UGE3126AY4	$V_{RRM} = 24kV$, $I_F = 0.8A$ (air self cooling), $V_F = 15V$
2	R1,R2	High Voltage Resistor	MG815 – 3.90MEG – 1%	$R = 3.9M\Omega$, $V_{MAX} = 30kV$, $P_{MAX} = 15W$
3	R3	High Voltage Resistor	HTE152 – 50M Ω	$R = 50M\Omega$, $V_{isol} = 45kV$, $P_{MAX} = 15W$
4	C3,...,C6	High Voltage Capacitor	780C40DKT78	$C = 780pF$, $V_{DC} = 40kV$, $\tan\delta = 20 \times 10^{-3}$
5	C1,C2	High Voltage Capacitor	UHV – 12A – 172K	$C = 1700pF$, $V_{DC} = 50kV$, $\tan\delta = 0.1\%$ (max)

$$\begin{aligned}
V_{RMS} &= \sqrt{\frac{1}{\tau} \int_0^{\tau} V^2(t) dt}, \quad V = V_P \sin(\omega t) \\
&= \frac{V_P}{\sqrt{\tau}} \sqrt{\int_0^{\frac{\tau}{2}} \sin^2(\omega t) dt} \\
&= \frac{V_P}{\sqrt{2\tau}} \sqrt{\int_0^{\frac{\tau}{2}} 1 dt - \int_0^{\frac{\tau}{2}} \cos(2\omega t) dt} \\
&= \frac{V_P}{\sqrt{2\tau}} \sqrt{\frac{\tau}{2} - \frac{1}{2\omega} (\sin(\omega\tau) - 0)} \\
&= \frac{V_P}{\sqrt{2\tau}} \sqrt{\tau - \frac{1}{2\pi f} (\sin(\omega\tau))} \\
&= \frac{V_P \sqrt{f}}{2} \sqrt{\frac{1}{f} - \frac{1}{2\pi f} \sin(2\pi)} \\
&= \frac{V_P}{2}
\end{aligned} \tag{3.4}$$

Based on Equation 3.4 if the maximum possible voltage is considered $V_p = 25kV$, then the continuous current flowing through resistor R1 and R2 is as follows

$$I_{R1,R2} = \frac{V}{R1 + R2} = \frac{12.5kV_{RMS}}{7.8M\Omega} = 1.60mA \tag{3.5}$$

The maximum power dissipation capability of R1 and R2 in continuous mode has been given in Table 3.2 by calculating the maximum dissipated power and comparing it with Table 3.2

$$W_{R1,R2} = RI^2 = 3.9M\Omega \cdot (1.60mA)^2 = 9.95W \tag{3.6}$$

It is in the range of allowed continuous power dissipation in resistors. The next parameter to consider is the time constant of the resistor R1 and R2 with C3, ..., C6 and the capacitance of the test cell. Later from The next quantity is the power quality or the power factor of the connected load to the diode. Considering Figure 3.7, the load connected to the diode is a resistive and capacitive load. The power factor of this load is an important quantity that can change the output waveform. The output will be pinned to the maximum in the large capacitive loads; Therefore, the load should be kept resistive with a low time constant to produce the desired waveform. Considering the voltage to be at maximum value of $V_{rms} = 12.5kV$

$$I_R = \frac{V_{RMS}}{R} = \frac{12500}{3.9M\Omega \times 2} = 1.6mA \tag{3.7}$$

It should be explained here that the range for the capacitive current flowing in the test cell can be calculated by simulating the test cell and obtaining the capacitance value in 50 Hz. But since, at this stage of the work, the setup was made and the test cell was already tested in full cycle mode therefore, the range of the capacitive current was known.

$$I_C \approx 20\mu\text{A} \implies Q = \frac{I_R}{I_C} = 80 \quad (3.8)$$

The electrical behaviour of the test cell will be explained in the next section. based on the explanation in section 2.4 and Figure 2.9, it is easy to understand that the test cell can be considered as a capacitance in parallel with a nonlinear resistance and the quality factor shows that the capacitance is larger. therefore, the test cell has been approximated as only a capacitance, and the value is obtained from the results of the practical experiments over the full-cycle AC setup.

$$C \approx 195\text{pF} \implies \tau = R \times C = 2 * 3.9 * 10^6 * 195 * 10^{-12} = 1.52\text{ms} \quad (3.9)$$

and since the maximum frequency of tests is 60 Hz

$$\tau_{Mf} = 1/60 = 16.6\text{ms} \implies \tau_{Mf} = 91 \times \tau \quad (3.10)$$

Therefore, the time constant is small enough not to distort the waveform too much.

3.3. Test Cell

The test cell has a multilayered complex needle design. A screw mechanism with O-ring sealants is incorporated to connect the needle to the test cell to allow removal and measurement of the needle tip without draining the liquid.

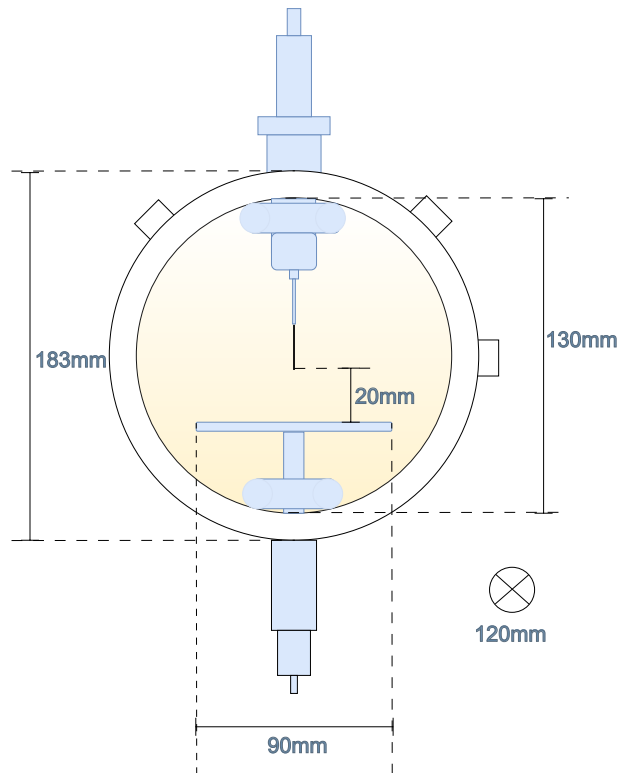


Figure 3.9: The test cell (test sample) [69]

The test cell comprises a cylindrical casing made of PTFE (polytetrafluoroethylene), which securely houses two glass walls on opposing sides. These glass walls are sealed together using O-rings and

serve to enclose the test cell. It has been shown in Figure 3.9. Within the casing, the plate and needle contacts are held in place. An adjustable mechanism located on the plate side regulates the gap between the two contacts.

Capable of containing approximately 800[ml] of liquid, the test cell features a mechanism designed to equalize internal pressure with the surrounding ambient pressure after filling. Multiple holes have been drilled into the PTFE casing to facilitate access, drainage, and filling of the test cell. These holes are sealed using caps fitted with O-rings and threads to ensure proper sealing and secure attachment. The test cell is supported by a stand crafted from plastic.

3.3.1. Needle

Most metals have a boiling point which is relatively low. This can set a limit to the permissible temperature. Tungsten, however, can operate at a temperature of over $3000\text{ }^\circ\text{C}$ [70]. Therefore, tungsten wire with a $100\text{ }\mu\text{m}$ diameter has been used for making the needle. There are multiple shielding and insulation layers added to it. It was designed and manufactured by L. Lundgaard, and G. Berg, in 2002.

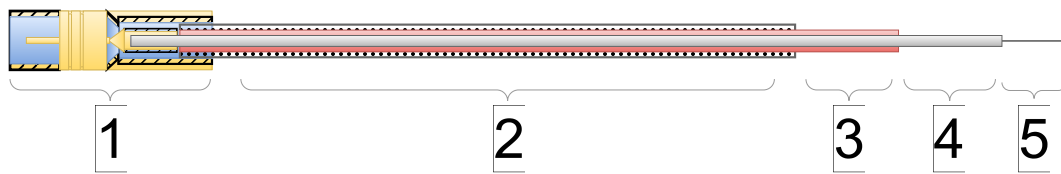


Figure 3.10: Configuration of the used needle (designed by L. Lundgaard, G. Berg)

According to the diagram in Figure 3.10, there is shielding for the wire against the surrounding ground, thereby minimizing noise interference.

A metal tube is holding the wire and taking the signal to the BNC connector at the end of the needle. The tip of the tungsten wire has been electrochemically etched using a 99% NAOH to the desired radius. Different parts of the needle shown in the Figure 3.10 are:

1. (1) Huber Suhner 11 SMB-50-2-41/111NE
2. (2) threaded steel tube
3. (3) insulating plastic tube
4. (4) Teknolab chromatograph needle
5. (5) etched tungsten wire

Measuring the needle tip was performed using a microscope connected to a computer system. Specifically, *Carl Zeiss MicroImaging GmbH* equipment, including *AxioVision* software and an *AxioCam MRc5* camera connected to a *Discovery V12 SteREO* microscope with $100\text{-}1000\times$, was employed for this purpose. The etched needle's tip can be approximated to a rotational hyperboloid at the tip, as illustrated in Figure 3.11. Given the critical significance of the tip radius in determining the electrical field, frequent measurements of the radius were conducted.

3.4. Conductive Current Measurement

The partial discharge causes small but rapid pulses in the current flowing through the needle, typically in the order of microamperes. These current pulses result from the presence and movement of the space charges and occur due to phenomena like avalanches or simultaneous ionization in the electric field.

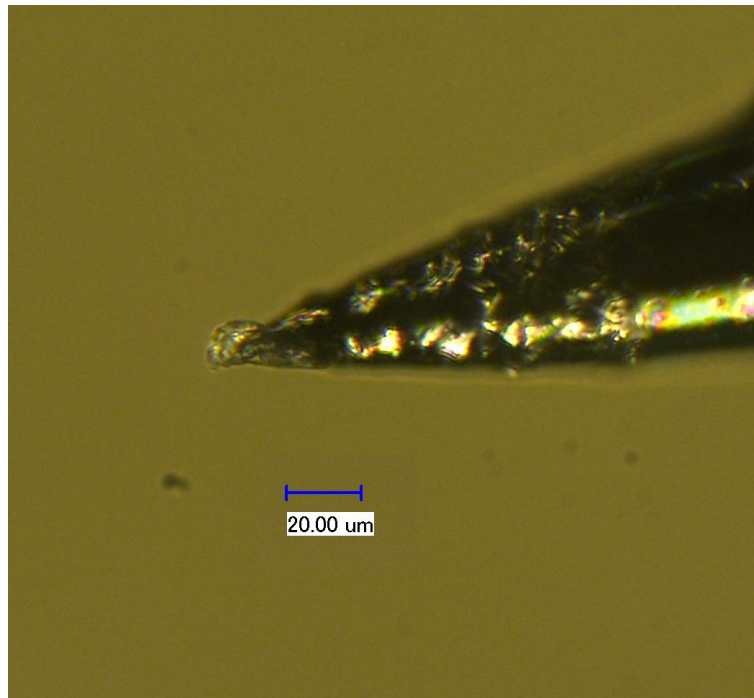


Figure 3.11: An example of etched needle tip under 1000 times magnification

according to the explanations in section 2.4 and explanations about the loss factor and Equation 2.12, the majority of the current flowing through the test cell is capacitive when an AC voltage is applied. There is also a limited nonlinear conductive current. This makes the test cell behave similarly to a parallel RC circuit, as demonstrated in Figure 3.12. Thus, a filtering system needs to be designed to separate out the capacitive current, ensuring that only the resistive component of the current, caused solely by space charges, is retained for accurate measurement. (see Figure 3.12 and Figure 3.20) The current measurement is through an SR570 current preamplifier by Stanford (SR570) connected to the digitizer system. (see Figure 3.3)

This configuration enables accurate monitoring and recording of current and voltage data for analysis and assessment purposes. The preamplifier connects the needle to the ground. The voltage difference between the two ports of SR570 is negligible.

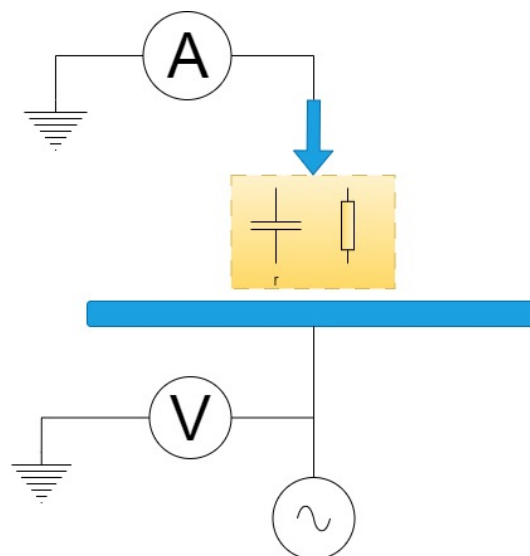


Figure 3.12: Equivalent circuit of the test cell

3.5. Optical Measurement of PDs

Previous studies show that PDs emit photons [57]. According to the theory explained in section 2.12. In order to monitor the radiant energy from the PD incidents at the needle tip, *Photomultiplier Tubes* have been used.

3.5.1. Photomultiplier Tube

A photomultiplier tube (PMT) is a vacuum tube that detects light. It has several applications such as scientific research, medical imaging, and industrial instrumentation. The structure and working mechanism of a photomultiplier tube are as follows: First, light (photons) enters through an incident window and strikes a photocathode (cathode), generating photoelectrons.

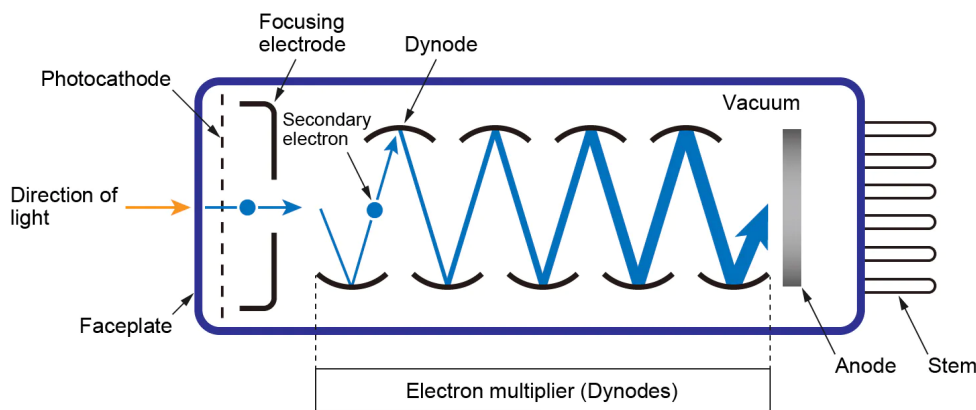


Figure 3.13: Photomultiplier tube structure implemented from [71]

reflected electrons are attracted by the high voltage applied to the cathode, electron multiplier (dynode) plates reflect and add to the electron flux multiple times inside the tube, and the anode is placed in the way of the last flux of electrons, which is the strongest. From the collision of the resulting electron flux to the anode plate a pulse of negative current in PMT's output. Most PMTs have 9 to 12 dynodes and amplify the photon impact by 10 to the power of 5th to 8th, depending on the applied voltage. Finally, the output is a detectable and measurable electric pulse. (Figure 3.13) A PMT is characterized by its ability to achieve an extremely high amplification rate, but this makes the output prone to *electrical noise* caused by *dark current*. Dark currents are distortions happening in PMT's answer due to unwanted photons reaching PMT from external sources or tunnelling effects that can occur in cathode or dynodes. The occurrence of dark current, whereby electrons are emitted from the photocathode due to thermal energy and subsequently strike the dynodes, results in electrical noise. Therefore, detecting the correct signal from the noise can be difficult.

There is a system designed for cooling down the contacts in PMT to minimize the electron emission due to the tunnelling effect.

Most photo multiplier tubes operate at a voltage range of 500 – 2000V. Therefore, the slightest fluctuation in the applied voltage to dynodes changes the magnifying ratio drastically. In order to accurately detect weak light, a low-noise and extremely stable voltage source is essential.

On the other hand, it poses a significant challenge. To address this issue, cooling systems are frequently employed to lower the temperature of the photocathode to sub-zero levels. This serves to reduce the likelihood of electron emission caused by thermal agitation, thereby minimizing the associated noise levels [71]. PMTs are from *Hamamatsu R943-02 EMI model* The optic PD detection system consists of two PMTs facing the test cell from the two opposite sides. Two windows of the test cell are made from tempered glass, which makes it hard to catch any ultraviolet light. However, PMTs are able to detect them. The cooling system is *Hamamatsu C4877* water-enhanced cooling system. A gain of approximately $g = 10^6$ has been set by applying $V_{DC} = 1700V$ according to the gain versus voltage curve for PMTs in Figure 3.14.

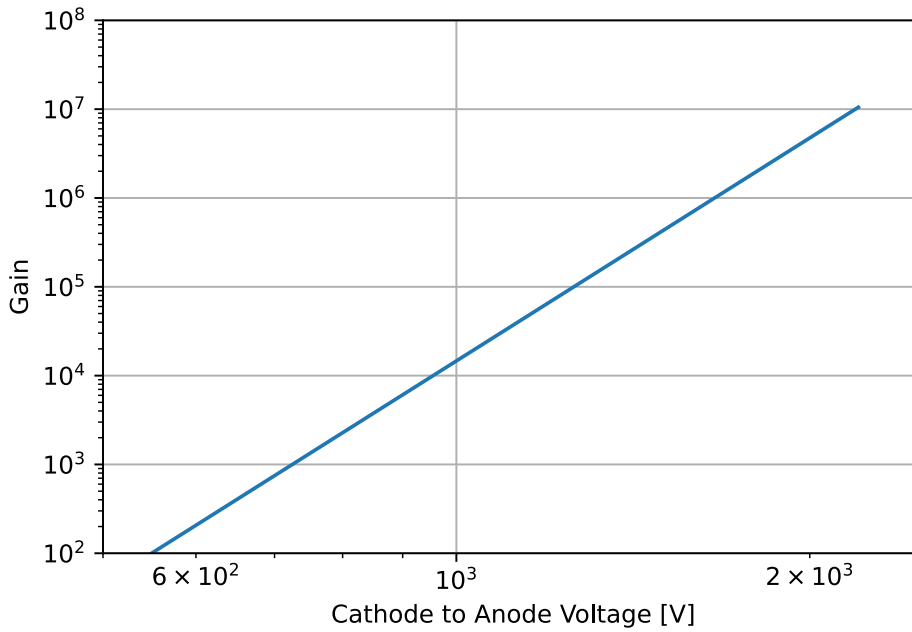


Figure 3.14: Gain versus voltage of PMT tube model R943-02 adopted from [72]

3.5.2. Coincidence Technique

There is always a level of light and thermal noise in the test. The goal of the experiment is to detect relatively small photon emissions of PDs. Therefore, the gain of the PMTs has been set around the maximum gain. The coincidence technique is an effective way to eliminate thermal and ambient light noise. Coincidence circuits have been widely used in particle detection experiments. The coincidence circuit used in this work includes two PMTs and standard coincidence instrumentation. The components of the coincidence circuit are

- Tennelec TC241 shaping amplifier
- Tennelec TC244 Shaping amplifier
- Ortec Quad 935 constant fraction discriminator
- Ortec CO4020 Quad 4-input logic unit
- Tennelec TC308 dual linear gate

Arrangements of the components have been shown in Figure 3.15. The output of PMTs is connected via a 50Ω coaxial cables to amplifiers (TC241 and TC244) and then to a quad constant fraction discriminator (CFD) module. CFD is a module that sets a constant fraction timing for amplified pulses of PMTs [73].

The first channel has been checked, and the threshold has been adjusted by turning the multiturn potentiometer T to 0.4 V . After getting a reasonable signal from the output, they have been fed individually to an AND gate with CO4020. This gate detects the coincidence of the pulses coming from the CFDs. Since there is no internal AND gate with this device, the logic has been made with available logic gates

$$A \cap B = \overline{(\overline{A} \cup \overline{B})} \quad (3.11)$$

For a proper answer, The coarse gain of the amplifier TC241 has been set to 50 and the input signal selector has been set to negative. The BLR switch is in the down position, and the signal from the unipolar output has been taken for the output. The other amplifier is TC244, and the course gain is set to 50 The multiplier is set to unimode, and the delay of the pulse is the same as TC241.

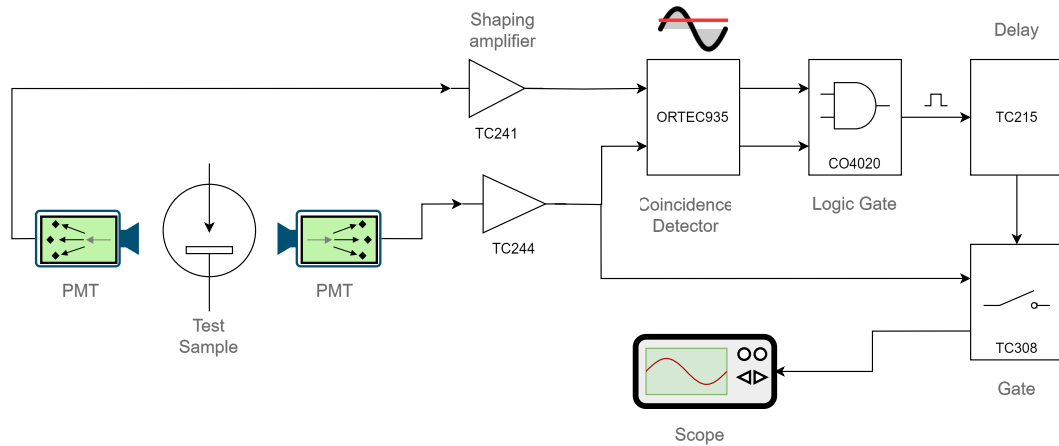
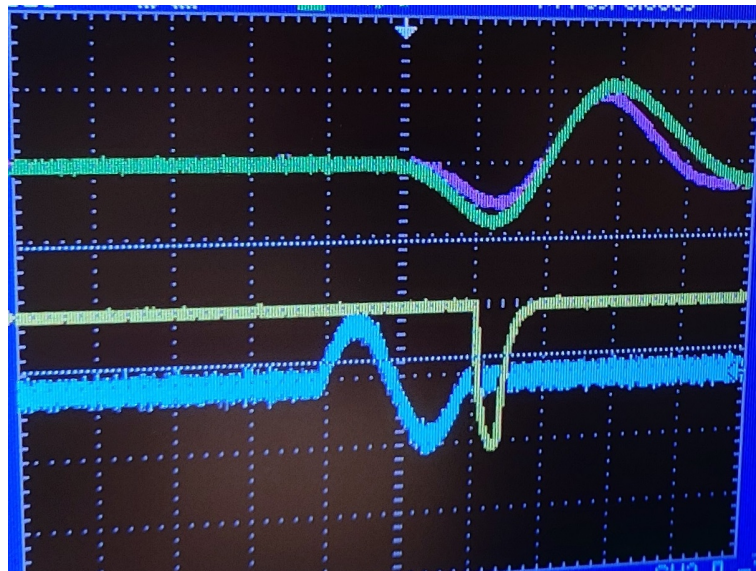


Figure 3.15: Arrangement of components of the coincidence circuit

The logic controller CO4020 DIP switch is critical. By choosing L instead of S for the last switch, longer outputs will be sent after detecting a signal. The gate has been switched by adjusting the DIP switch G to the middle position. The output signal is taken from one of the Y outputs on the device with a downside shape. The logic has been set to react to the increasing Edge instead of the decreasing edge of input signals to catch the moment of a new signal from CFD. After getting the output, the output pulse width is adjusted to $3\ \mu\text{s}$ by turning the W potentiometer.

At this point, the coincidence circuit has been tested with a sample signal formed by the signal generator to both amplifier inputs; the output of CO4020 has been shown in yellow in Figure 3.16. The blue curve shows the input sample signal generated, and the violet and green curves show the output signals of amplifiers TC241 and TC244, which are being fed to CFD, and the yellow curve is the output logic pulse of CO4020, which is a standard NIM logic pulse.



Channel-1: Yellow, 2V , Channel-2: Blue, 10mV
 Channel-3: Violet, 100mV , Channel-4: Green, 100mV
 X-s/div : 500ns

Figure 3.16: CO4020 output signal with amplifier output signals to a sample waveform

After testing the coincidence circuit, PMTs were connected, and the output of the CO4020 was given to the gate module TC308, and the output of one of the PMTs was given as input to the gate.

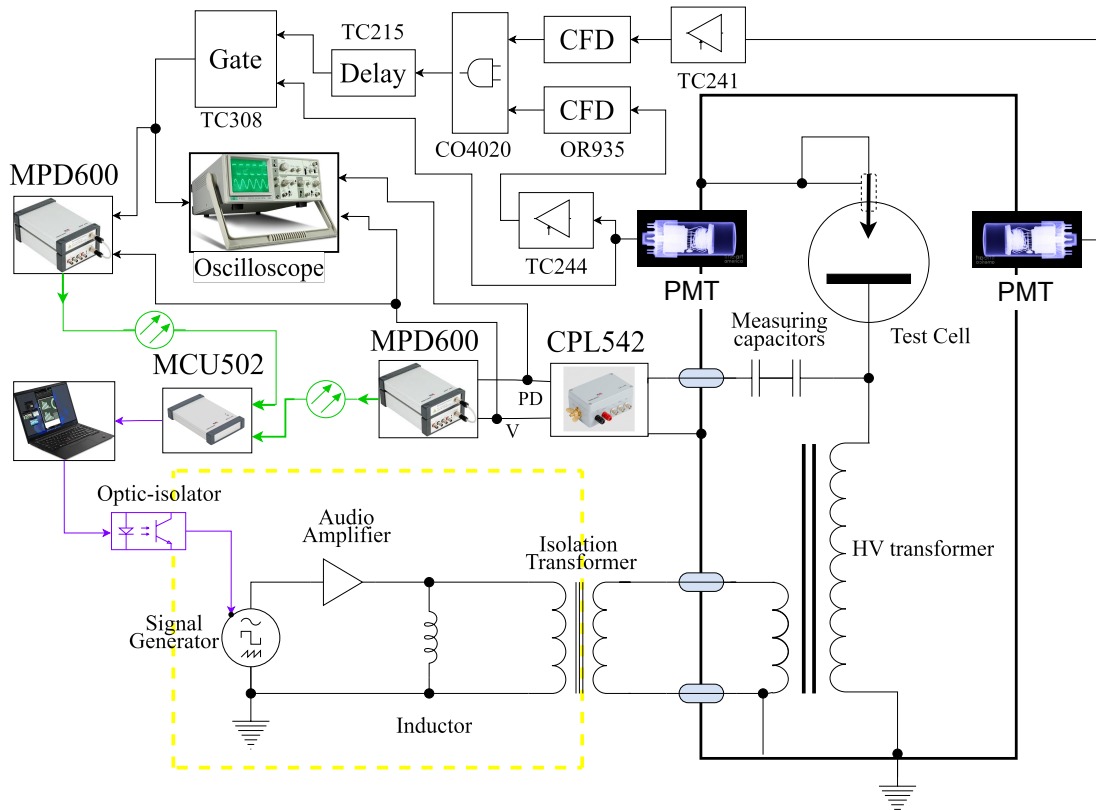


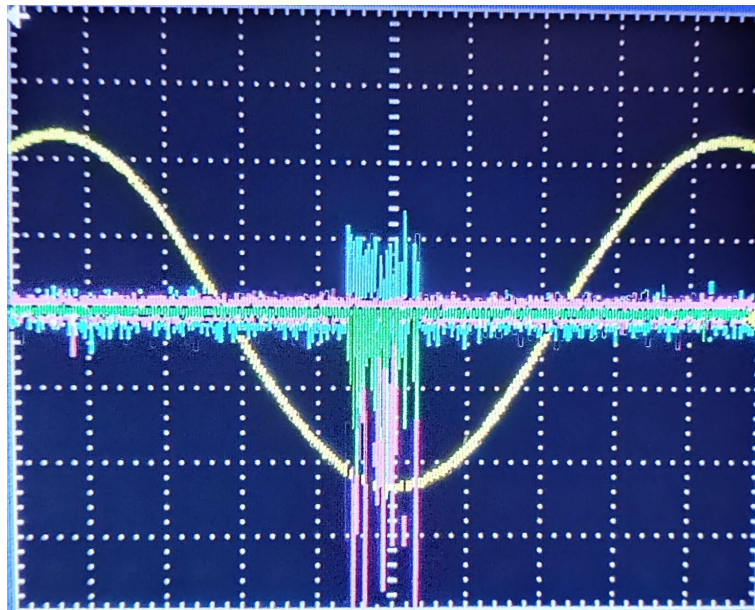
Figure 3.17: Diagram of the test setup for optical PD measurement

Thus, the output of the gate is the PMT signal only in small time intervals where a coincidence signal has been detected. Since it takes about $2\ \mu\text{s}$ for the whole process of coincidence circuit to detect until opening the gate, therefore a delay module TC215 with a $2\ \mu\text{s}$ delay has been connected in the way of signal to output. The circuit is checked for being able to detect PDs in air from a sample wire connected to the high voltage. A diagram of the test setup has been shown in Figure 3.17.

PD in the air from a wire installed to the high voltage side has been detected electrically with omicron and optically using the described system. The output of the optical system has been shown in Figure 3.18. The yellow curve in this figure represents the voltage waveform; the violet curve shows the output of CFD of one of the PMTs, the green curve shows the output of the gate and the blue curve shows the PD output from omicron. (CPL542)

Tektronix TDS2024C Oscilloscope, which has been used for the experiments, is capable of triggering PD signals from omicron. based on this the capability of the optical circuit for detecting any coincidence at PD signals from Omicron has been investigated. Figure 3.19 shows the output of the gate module (green) with PD signal from omicron (Blue).

Obviously, the coincidence circuit detects the PD, which is also detected by omicron, and the signal has been used to trigger the oscilloscope. there is about $2\ \mu\text{s}$ delay from the head of the PD signal to the moment of opening of the gate, which is happening with a relatively fast impulse and then the output signal of the PMT, which has a negative polarity is visible for around $3\ \mu\text{s}$ and then the gate closes with another fast impulse.



**Channel-1: Yellow, 100mV , Channel-2: Blue, 5mV
Channel-3: Violet, 5mV , Channel-4: Green, 10mV
X-s/div : 2.5 ms**

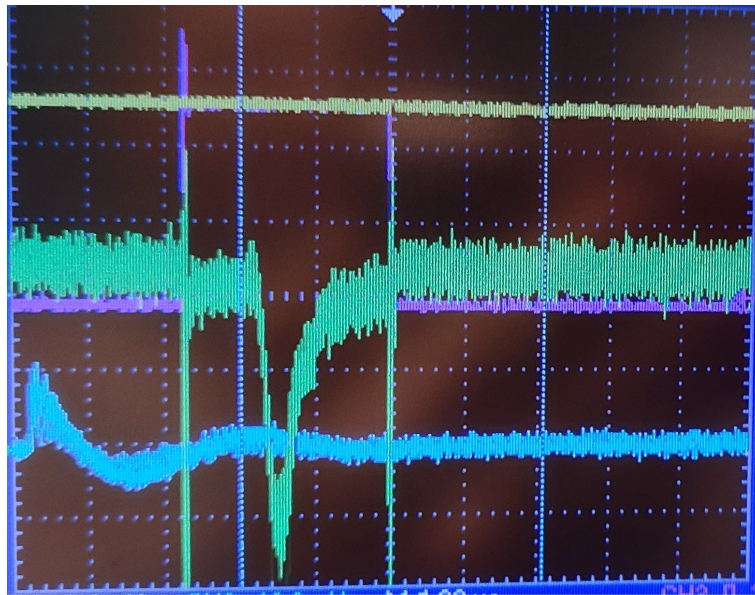
Figure 3.18: Coincidence circuit test results for PD in air

3.6. Selection of Liquids and Characteristics

Pure Cyclohexane sourced from *Uvasol* laboratory was standard for spectroscopy purposes. Prior to testing, the test sample underwent multiple washings with *Isopropanol* and was allowed sufficient time to completely vaporize under appropriate ventilation conditions. Following this preparation, the needle was sharpened to a tip radius of $1.77\ \mu\text{m}$ using the specified etching process in subsection 3.3.1. Subsequently, the test sample was filled with Cyclohexane and subjected to PDIV testing according to the instructions outlined in chapter 2. Additionally, the current through the needle was measured during the experiment. characteristics of Nytro 10XN has given in product datasheet in Appendix B MIDELE 7131 has passed through a $2\ \mu\text{m}$ filter with the help of a vacuum pump, and a new set of equipment has been used to prevent impurities from entering the liquid. A special method has been used for washing the test cell, including:

1. dismantling the test cell
2. washing the parts with isopropanol
3. assembling parts
4. washing with MIDELE 7131

However, due to the high surface tension of MIDELE 7131, impurities remained inside the test cell. The test cell was inspected visually after filling up with $1.2\ \mu\text{m}$ filtered MIDELE 7131, but there were impurities bigger than $1.2\ \mu\text{m}$ visible within the liquid. The test cell has been relaxed for some time for the impurities to settle down away from the needle plane region inside the test cell before tests. Properties of MIDELE 7131 are in Appendix B.



Channel-2: Blue, 10mV , Channel-4: Green, 5mV
X-s/div : 1us

Figure 3.19: Coincidence circuit response to PD pulse in PD trigged scope

3.7. PDIV Definition

There are several definitions exist for the PDIV. It is important to know that there is not a straightforward threshold for it because it also depends on the measuring time. Definitions vary depending on whether the measurement is conducted under DC or AC voltage regimes. Experiments in [61] demonstrate that PDIV voltages differ based on the criteria used, making it challenging to establish a single meaningful threshold for assessing the quality of insulating liquids. To address this, a specific definition for PDIV testing is necessary.

However, the present IEC standards may not sufficiently reflect all the relevant functional dielectric properties of the insulating liquids [51]. After reviewing studies such as [61], a definition has been formulated for this study.

The PDIV considered in this study is obtained by increasing the voltage by 250 volts, with each step lasting one minute. The voltage at which the first PD larger than 2 pC occurs during this process is considered the PDIV.

3.8. Data Analysis

Due to the dominance of capacitive current in the needle [23] appropriate filtering techniques are necessary to extract meaningful information from the raw data acquired by the digitizer. It should be emphasized that the measured current corresponds to the current flow from the needle to the ground, as the setup has been described in detail in subsection 3.1.1 and illustrated in Figure 3.3. the current flowing through the needle to the ground always leads to the voltage waveform. Figure 3.20 is a demonstration of an example of the raw data obtained from the digitizer.

However, obtaining more data from the current waveform without separating the capacitive current is not possible. The voltage and current waveform in practice might look sinusoidal, but they include some higher-order harmonics as well. no matter how much effort has been given to make the voltage waveform close to sinusoidal. According to

$$i_C = C \frac{dV}{dt} \quad (3.12)$$

capacitive current can be obtained by calculating the derivative of the voltage and having the capacitance value. There are always some defects in the waveform of the voltage, which makes the derivative of the voltage a fluctuating wave. Also, considering that measured voltage is no more than a series of data points with a value in time where the gap between each two points is defined by the quality of measurement, it can not be zero. Therefore, filtering in higher-order frequencies is necessary. Understanding the significance of filtering and processing experimental data is important, particularly

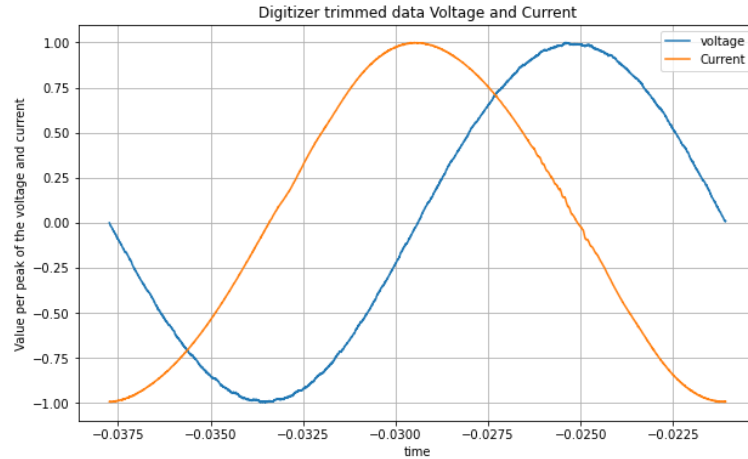


Figure 3.20: Measured voltage and current from Cyclohexane 60 Hz at 17.83 kV voltage peak-peak

when utilizing AC voltage above 10Hz, as in this study. In this setup, the plane is connected to the high voltage, while the omicron device measures voltage from the high voltage (plane) side (see Section subsection 3.1.1). Consequently, to accurately depict the polarity of needle potential and current, the current waveform needs to be inverted relative to the voltage waveform.

3.8.1. Python Script for Analysing the Data

All measured data has been saved in ".asc" format. They have been imported into Python for multiple purposes, like visualising the saved data, filtering the voltage, calculating the I_C , subtracting it from the measured current to obtain conductive current, and further studying data such as finding patterns and curve fitting functions. A flowchart of the code has been shown in Figure 3.21. Different parts of the code include:

- loading data from files with a ".asc" extension in a specified folder path into a dictionary.
- plotting data from the dictionary into separate graphs and saving them.
- Extracting voltage and current data from the loaded data dictionary.
- Performing Fourier Transform on the data to analyze frequency components.
- Filtering out frequency components above a certain limit. (it has been set to 300 Hz for voltage before derivation)
- Calculating derivative of voltage data in the frequency domain to estimate capacitance
- Plotting voltage, capacitance, and resistive current data.
- Saving the plots and resistive current data to files.

The Python script of the code has been given in Appendix C.

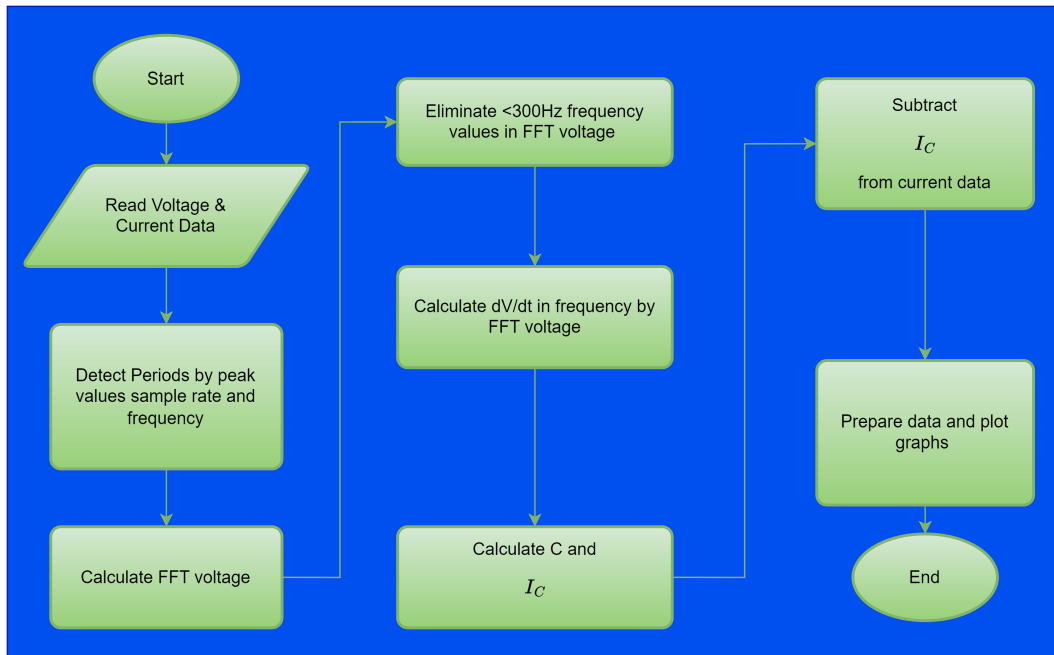


Figure 3.21: Flowchart of the Python script for analysing data

3.9. Test Procedure

The devised test procedure for investigating various liquids and discerning their distinctions is outlined below:

- Fill the test cell with a specified volume of liquid and seal the cell.
- Agitate the test cell to facilitate cleaning.
- Empty the test cell, ensuring that no remains of the "old" liquid are in the test cell.
- Refill the test cell with the new liquid, taking care to prevent the formation of air bubbles.
- Position the test cell centrally within the test rack and align it with the optics.
- Initiate measurements using different waveforms (half-cycle or full-cycle):
 - First, conduct current measurements.
 - Second, perform PD measurements along with needle current measurements in different voltage steps.
 - third, apply a PDIV test from voltages below 2kV until reaching the PDIV.
- Analyze the obtained results with the data analysing tool.

4

Experimental Results

The voltage term in all parts points to the peak difference in electric potential between needle and plane contacts in the test cell. The needle is positively charged at the moment when the plane's charge is negative. The voltage has been measured by CPL542 (see section 3.1 and Figure 3.3) connected to the high voltage arm and plate. Therefore, a 180° phase shift has been added manually to full-cycle Histograms in order to represent the needle's voltage. Applied to all voltage measurements to show the correct polarity of the voltage on the needle.

The gap between the needle tip and the plane has been measured to be $20 \pm 0.5\text{mm}$ in all tests.

Needle tip radii have been measured by using a digital microscope with $1000\times$ capability. They have been measured visually by using some measuring aids from the software of the digital microscope. The accuracy of the needle tip measurement is $\pm 0.5\mu\text{m}$

4.1. PD Patterns in Tested Liquids

The voltage curve drawn in green in all histograms has different values; Therefore, the vertical axis for voltage amplitude has not been shown. Figure 4.1(a) shows the histogram of PD pulses in Cyclohexane at 19.81 kV and 30 Hz. Figure 4.1(b) shows result for Cyclohexane at 19.35 kV and 45 Hz and PD pattern for Cyclohexane at 19.76 kV and 60 Hz has shown in Figure 4.1(c).

All PD histograms have been recorded in 1 minute duration. The same needle has been installed for all three tests in Figure 4.1, in which the tip radius was $1.7\mu\text{m}$ before tests and $6.03\mu\text{m}$ after tests. It should be noted that each dot on the histogram shows the charge magnitude of the PD pulse on the vertical axis, and the phase angle is shown on the horizontal axis in time sequences.

The colour spectrum at the right side of each result shows the intensity of pulses per second. Red colour means more intense, and blue colour means less intense.

All PD histograms have been filtered for Pulses smaller than 1pC because of the noise.

An intense accumulation of PD pulses between 2pC to 11pC in the negative peak of the voltage has been observed in Figure 4.1. There is also a symmetric concentration for the incidents happening at a negative peak. There are some rare incidents happening at the positive peak, but they have a relatively large magnitude. from 40pC to 70pC , and evidence suggests that larger PD incidents in magnitude are more effective regarding needle erosion. (see section 4.4)

PD histogram results for Nytro 10XN tests have been presented in Figure 4.2. In Figure 4.2, (a) shows the result of test under 22.89 kV tension and with 30 Hz frequency while (b) and (c) are showing results under 23.91 kV and 23.26 kV with frequencies of 45 Hz and 60 Hz, relatively. In contrast with Cyclohexane, the PD incidents at the negative peak are more concentrated towards the lower edge and peak voltage line. At the positive peak, there are a few incidents happening above 100pC magnitude. This is different in PD patterns of Nytro 10XN than in other liquids. The tip radius was $5.24\mu\text{m}$ before tests and $11.6\mu\text{m}$ after tests. It has been believed that this pattern may have some parts below 1pC , but PD pulses under this magnitude have not been considered due to noise.

Figure 4.3 shows the PD histogram of MIDEL 7131 tests under 20.86 kV with 30 Hz, 22.41 kV with 45 Hz and 22.75 kV with 60 Hz. These patterns show less number of incidents in the same amount of time because the electrical field of these tests was closer to PDIV. Since they were tested by a needle with the tip round of $5.26\mu\text{m}$ and after tests tip radius was $10.40\mu\text{m}$. But the pattern for PD pulses in the MIDEL 7131 around the negative peak of the voltage is concentrated from 1pC to 30pC but the pattern around the positive peak includes two parts, some rare but large incidents happened from 100pC to 1nC and another part from 1pC to 2pC but more incidents in number in the test time duration.

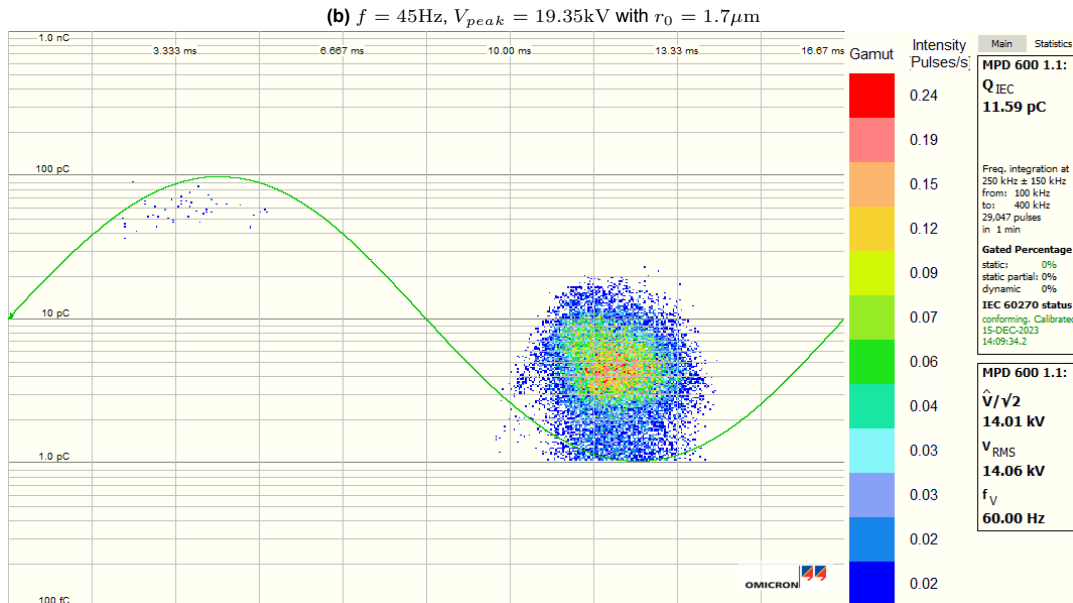
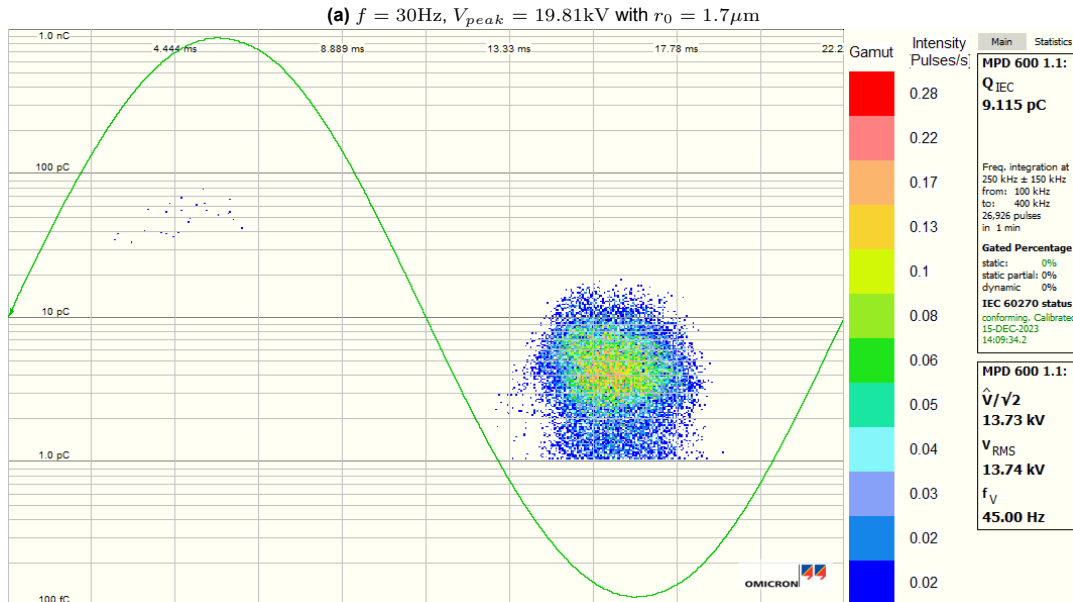
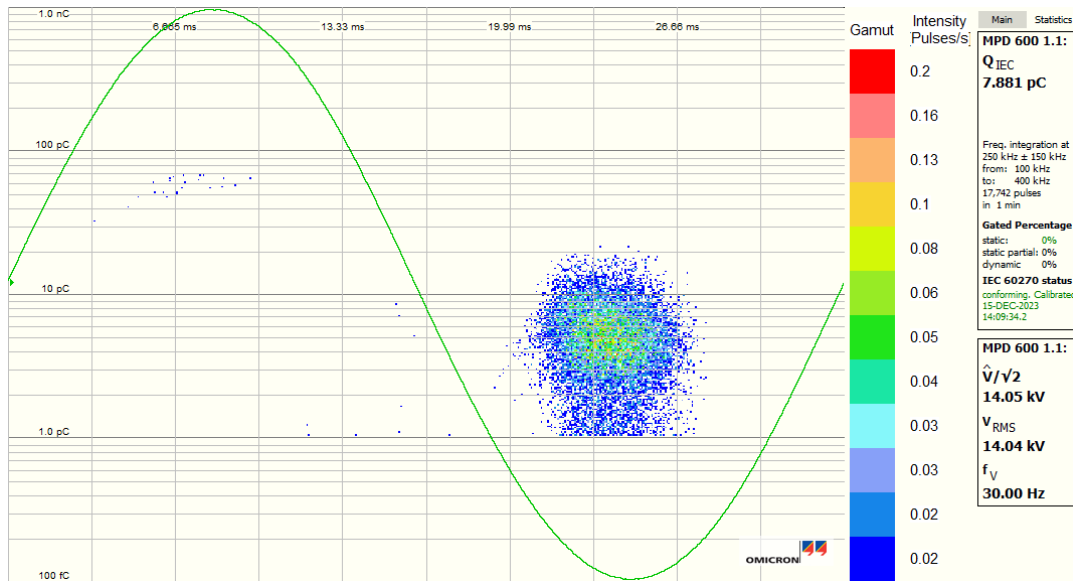


Figure 4.1: Histogram of the PD pulses larger than 1 pC of the tests with Cyclohexane

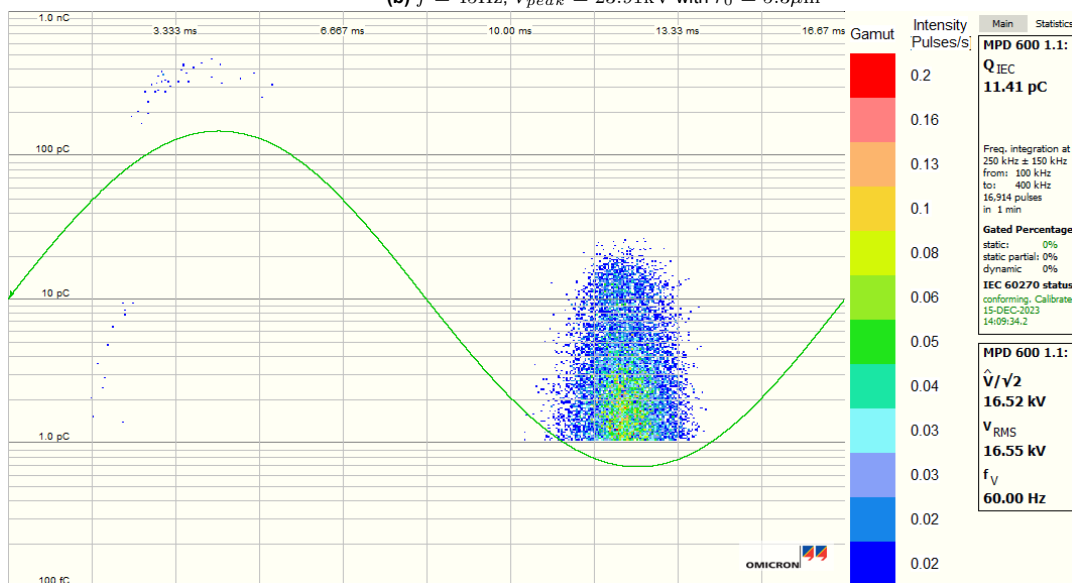
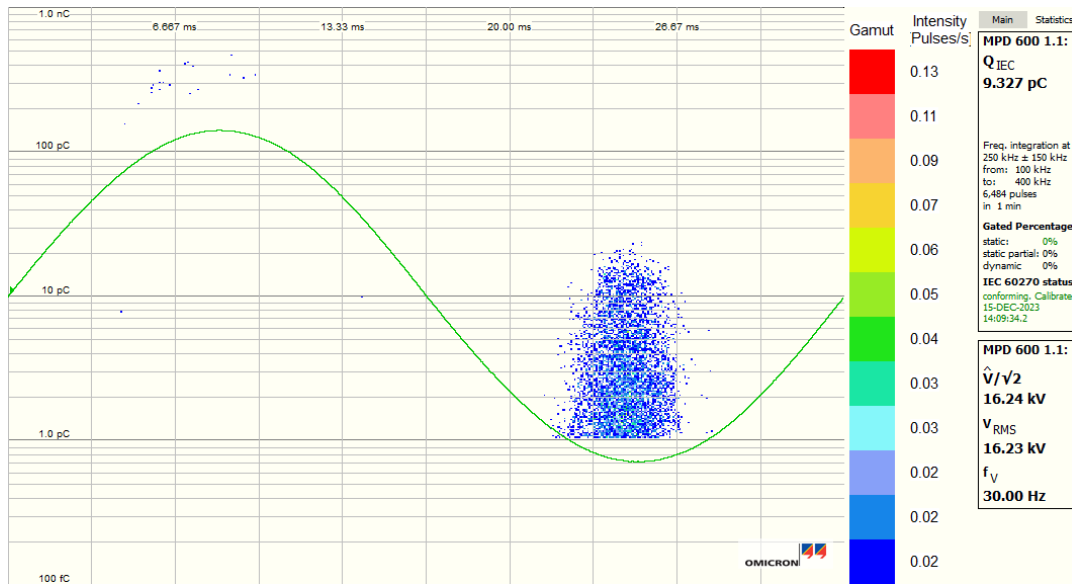
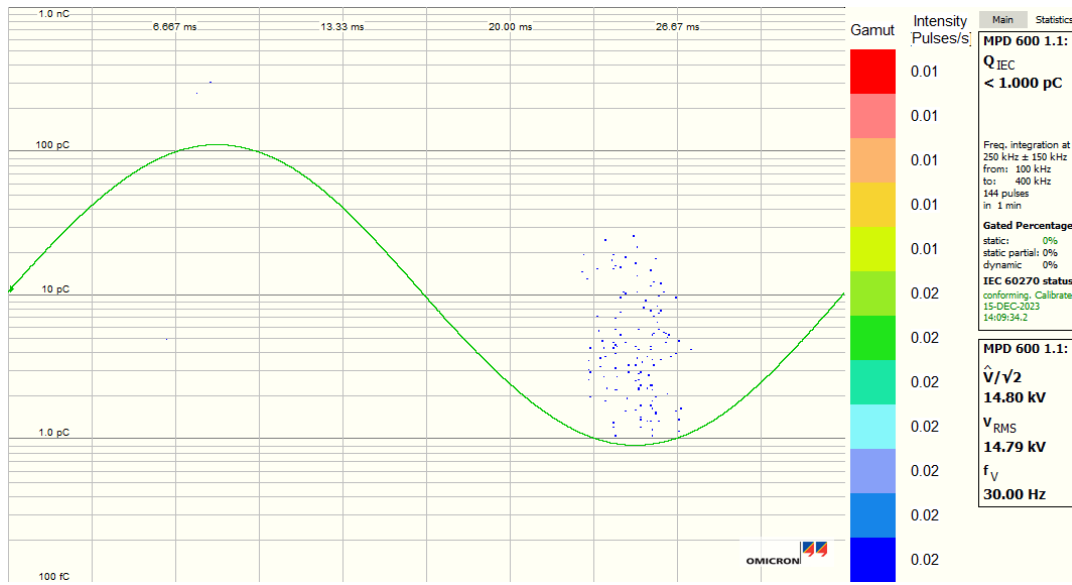
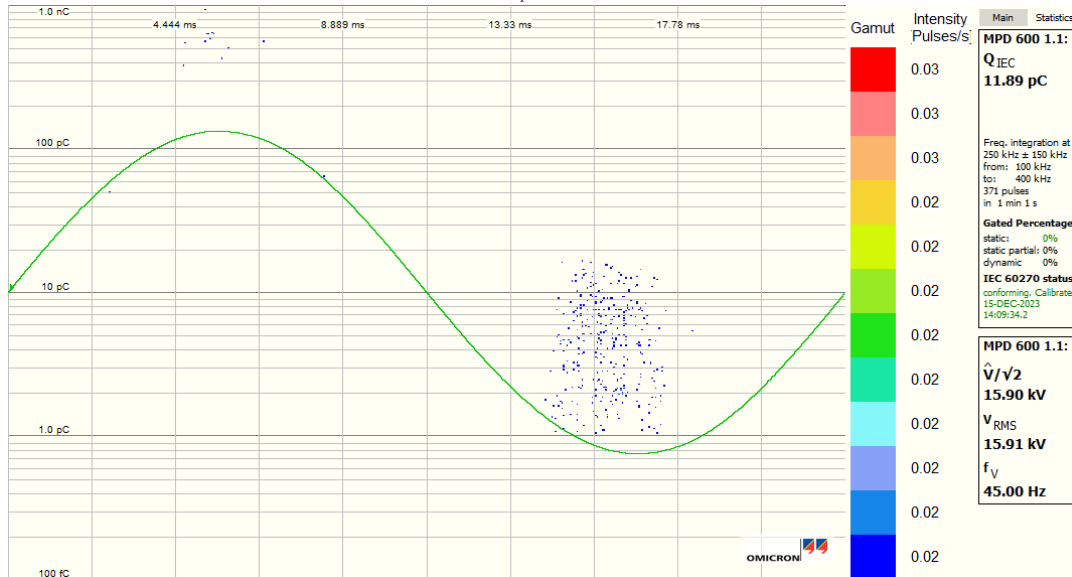


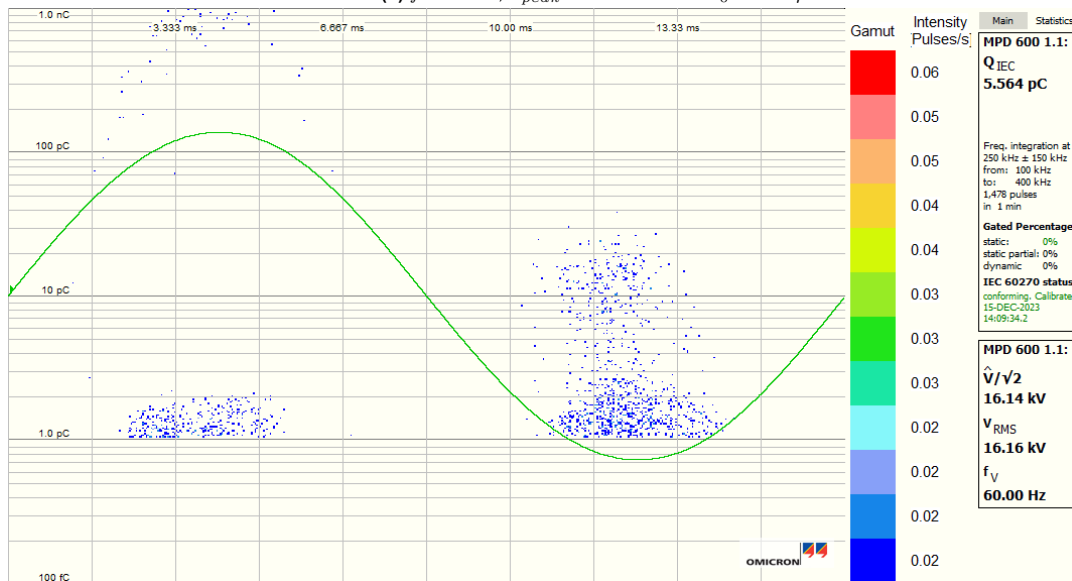
Figure 4.2: Histogram of the PD pulses larger than 1 pC of the tests with Nytro 10XN



(a) $f = 30\text{Hz}$, $V_{peak} = 20.86\text{kV}$ with $r_0 = 5.26\mu\text{m}$



(b) $f = 45\text{Hz}$, $V_{peak} = 22.41\text{kV}$ with $r_0 = 5.26\mu\text{m}$



(c) $f = 60\text{Hz}$, $V_{peak} = 22.75\text{kV}$ with $r_0 = 5.26\mu\text{m}$

Figure 4.3: Histogram of the PD pulses larger than 1 pC of the tests with MIDEL 7131

4.2. Conductive Current Measurements

Figure 4.4 shows a group of graphs of the conductive current waveforms in time for each test for Cyclohexane in four voltage levels and in three frequencies.

Applied voltage waveform in these tests has been shown in these results with blue colour. The peak value of the applied voltage has also been shown on each graph.

The vertical axis of each graph consists of two axes the left is related to the conductive current in nA, and the vertical axis at the right of each graph is related to voltage in kV. Difference in horizontal axis values is due to data recording time and also due to the selection of the cycle in the whole data range. voltage waveform has been recorded from the output of the CPL542 in the upper omicron device.

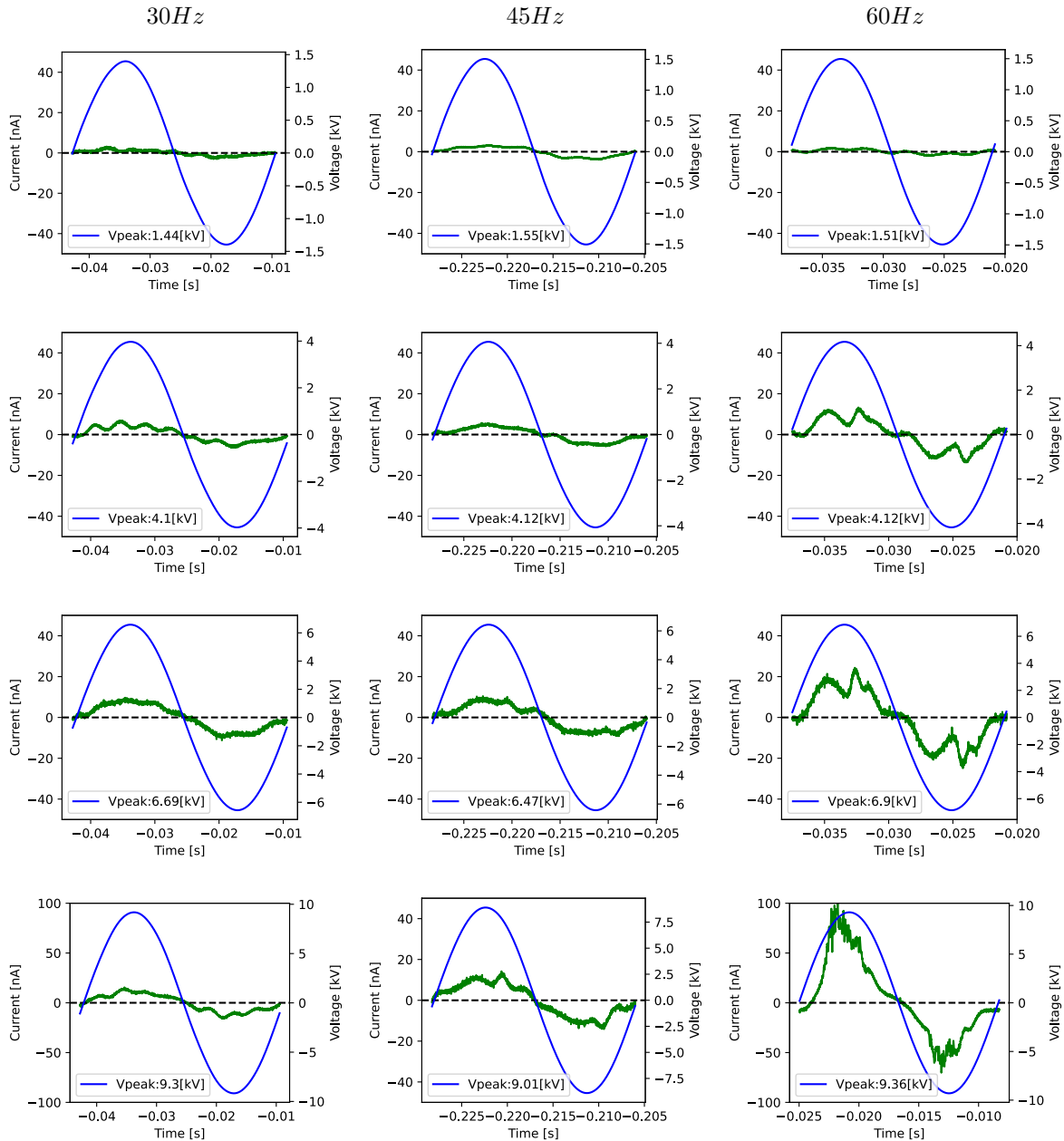


Figure 4.4: Conductive current and voltage versus time in different voltage steps and frequencies for Cyclohexane

Therefore, the data is related to the plane voltage. Voltage data has been inverted to show the voltage on the needle.

It is obvious that the amplitude of the conducting current increases with the increase of the voltage. It has a linear behaviour under PDIV voltage, which is visible in all graphs. As it is evident from the Figure 4.4.

The same needle has been used for each liquid's test group. It should not be neglected that needle erosion happened during the tests, which has been discussed in section 4.4.

The conductive current of Nytro 10XN and MIDEL 7131 has been tested with the same test procedure. The test cell has been dismantled and washed properly according to the explanation in chapter 3 with Isopropanol after emptying Nytro 10XN.

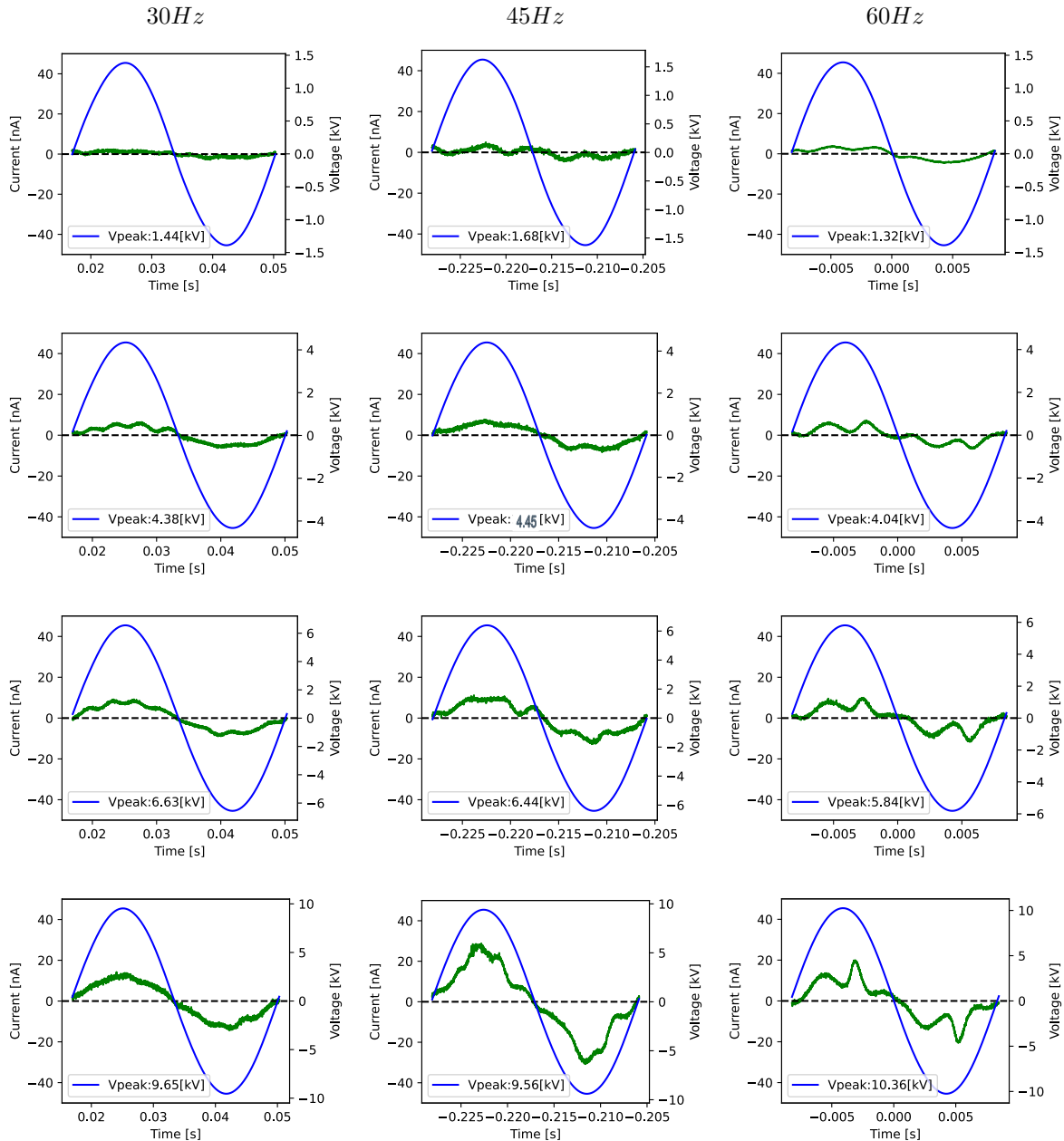


Figure 4.5: Conductive current and voltage versus time in different voltage steps and frequencies for Nytro 10XN

It has been washed again with MIDEI 7131 a couple of times, and the used MIDEI 7131 has been thrown out to make sure that it has the minimum effect on the purity of the liquid in the next test. It is notable that the Nytro 10XN underwent a filtration process utilizing a 1.2 μm filter in conjunction with a vacuum pump for purification.

There is a visible asymmetry between two sides in each half-cycle in the higher frequency (60 Hz). They can be least possibly due to harmonics in the measured current or voltage and more likely due to explanations in chapter 6.

The same filtering settings have been used for all tests, even for the lower frequencies at high voltage levels, but they do not show asymmetry in the current waveform. Another reason this asymmetry is related to the movement of space charges is the same shape between tests with different frequencies. In Figure 4.4, comparing graphs for Cyclohexane at 9.01 kV with 45 Hz with Cyclohexane at 6.9 kV with

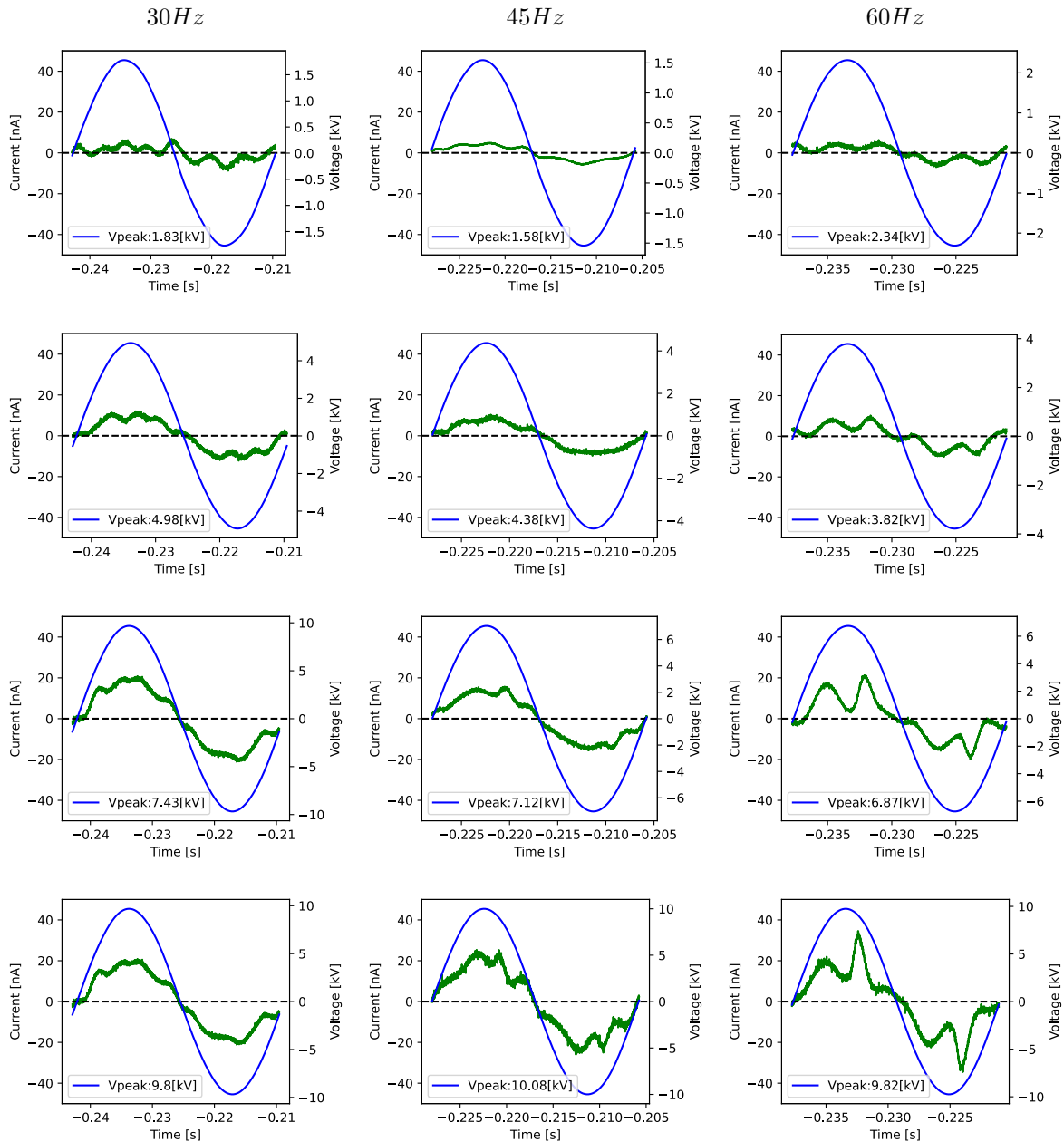


Figure 4.6: Conductive current and voltage versus time in different voltage steps and frequencies for MIDEI 7131

60 Hz, The same trend happens in different frequencies in the second quarter of the cycle. In addition to that, this trend is not related to the molecule since the same asymmetry is detectable in Nytro 10XN and MIDEL 7131. In Figure 4.5 Nytro 10XN at 6.44 kV with 45 Hz and at 5.84 kV with 60 Hz shows the same trend. MIDEL 7131 test results in Figure 4.6 at 7.12 kV with 45 Hz and at 6.87 kV with 60 Hz also showing the same trend. In Figure 4.4 in the graph for $V_{peak} = 9.36\text{kV}$ and 60 Hz the waveform shape has been changed because of PD incidents happening at the voltage peak at both positive and negative peaks of the voltage.

It is notable that PD pulses are happening in positive and negative half-cycles at the same time. This is in accordance with the results from the optical PD measurement which is explained later in section 4.6. The FFT components for frequencies lower than 300 Hz have been used for calculating and resembling the voltage derivative and, consequently, the capacitive current. This frequency has been chosen after many tests because it gives accurate results with lower fluctuations.

4.2.1. Peak Values of the Conductive Current

The peak current graphs for each liquid have been shown in Figure 4.7 for testing Cyclohexane, Figure 4.8 for testing Nytro 10XN and Figure 4.9 for testing MIDEL 7131.

The highest value of the current passing through liquid between needle and plane has been considered in the whole recorded span of each data.

The radius of the needle tip is mentioned for each test; however, it changed during the tests due to needle erosion. It had a considerable influence on the electric field strength. The relativity between conductive current and the field strength is according to

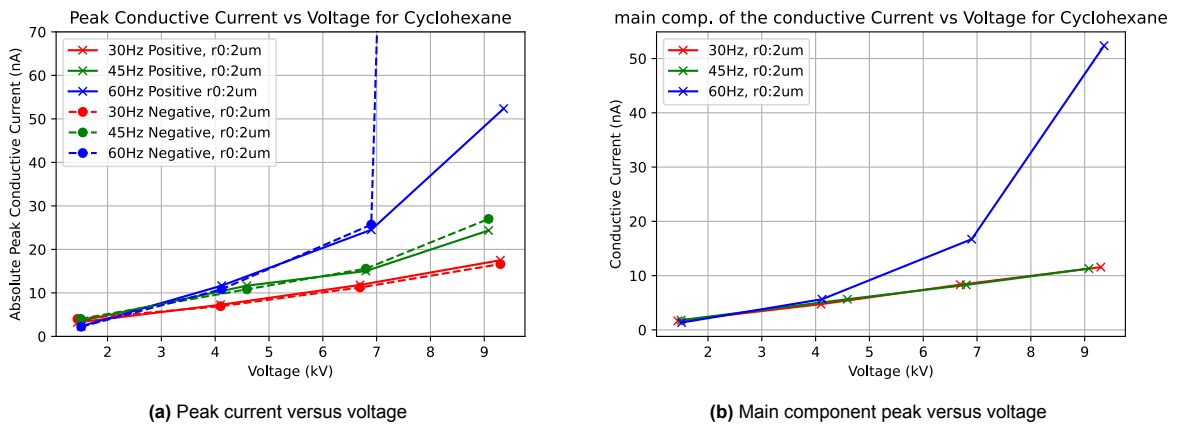


Figure 4.7: Conductive Current peak versus voltage for Cyclohexane

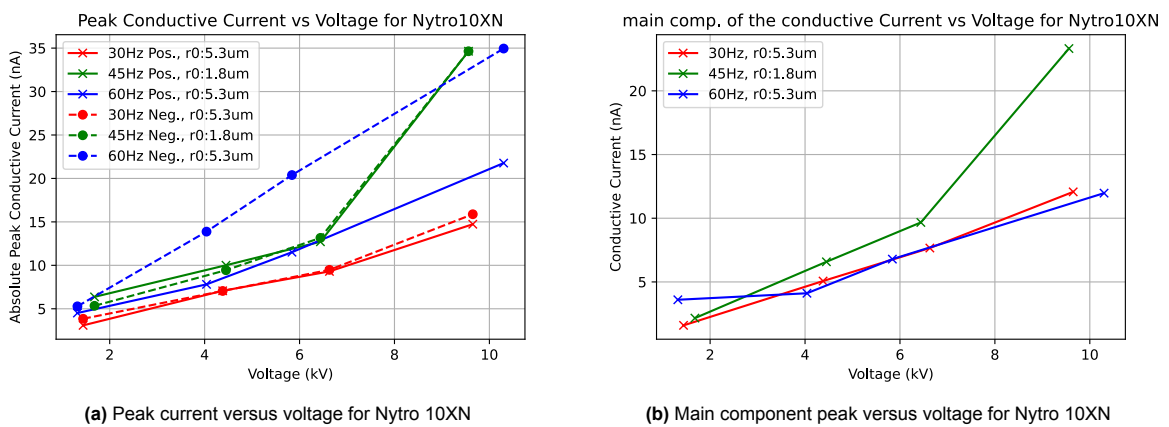


Figure 4.8: Conductive Current peak versus voltage for Nytro 10XN

Figure 2.9. Figure 4.7(a) shows the positive and negative peaks of the conductive current for Cyclohexane tests. This graph shows a direct relation between the frequency and conductive current peak. Figure 4.8 shows the peak current and peak of the main component of the conductive current in Nytro 10XN. These graphs show a slope change at a transitioning point from nonlinear to exponential regions. The same trend has also been observed for MIDE L 7131 at the lower electric field at Figure 4.9. At higher voltages, the peak current values indicate several pulsed peaks resulting from PD incidents, which affect the maximum value.

The maximum value of the data sets with PD incidents usually indicates the maximum range of the preamplifier, which doesn't give any insight into the phenomena.

Therefore, the peak value of the main frequency of the conductive current has been obtained.

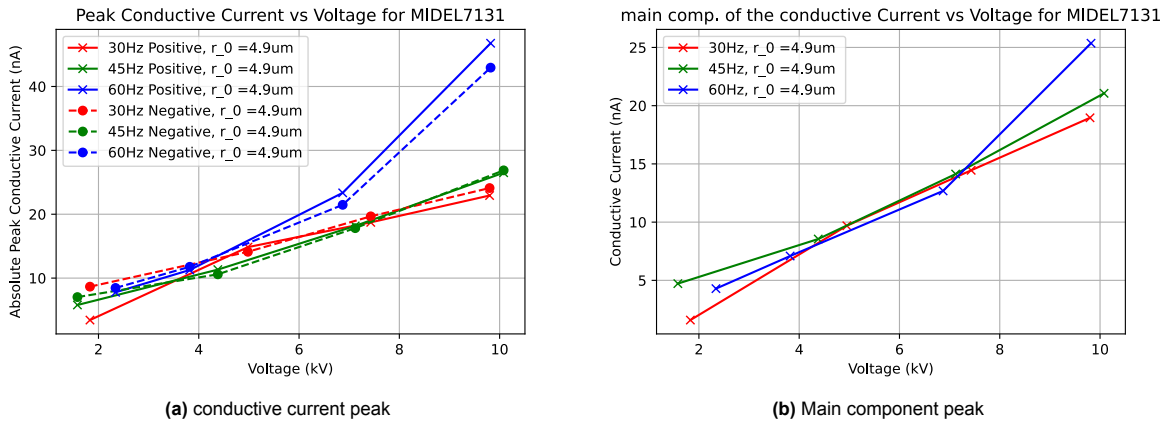


Figure 4.9: Conductive Current peak versus voltage for MIDE L 7131

Consequently, with the change of the needle tip radius due to erosion, the conductive current also changes. This change is relatively linear in Region 1, but it is nonlinear in Region 2 and close to exponential in Region 3.

Another test has been prepared to measure the conductive current of different needle tip radii on Nytro 10XN in different voltages. Figure 4.10 shows results of Nytro 10XN tested with 45 Hz and two different needles tip radii of $1.8\mu\text{m}$ and $2.3\mu\text{m}$ which clearly shows the difference in conductive current in the same voltages due to different field strength and active region volume of the needle tip.

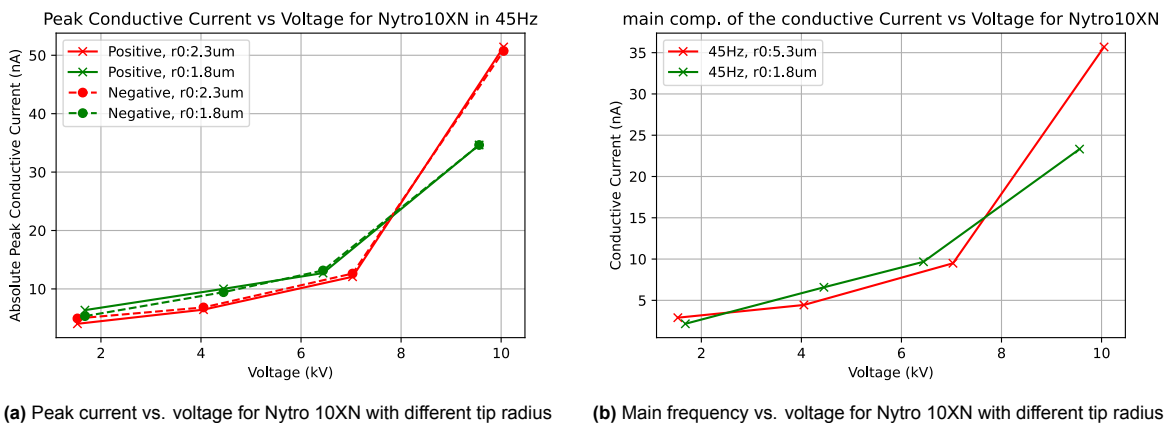


Figure 4.10: Conductive Current peak versus voltage with different needle tip radii for Nytro 10XN

4.2.2. Influence of Remaining Space Charges in the Region After a PD Incident

The influence of space charges on the conductive current is detectable in the periods following a large PD incident in all three liquids. Investigating this phenomenon requires a special situation where a

couple of periods without PDs would follow a large PD incident. Therefore all the tests have been taken at a voltage around PDIV level. Figure 4.11 shows the conductive current (resistive current) in 7 cycles after a PD incident in Cyclohexane under 16.03 kV peak voltage and 30 Hz voltage and frequency. The exponential fit is a function fitted to the negative peaks of the conductive current from the first peak to a few later peaks. The same procedure has been used for Nytro 10XN under 19.3 kV and 30 Hz, which has been shown in Figure 4.12.

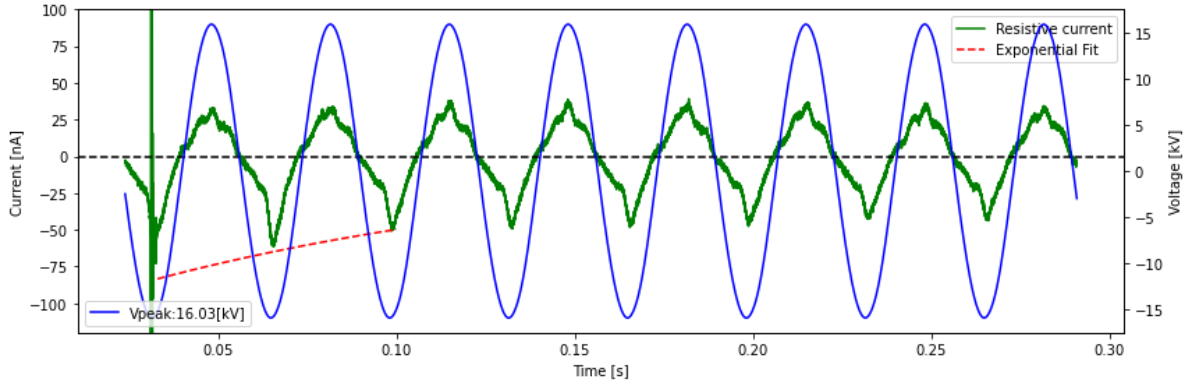


Figure 4.11: Conductive current and voltage waveforms after a large PD incident in Cyclohexane at 30Hz

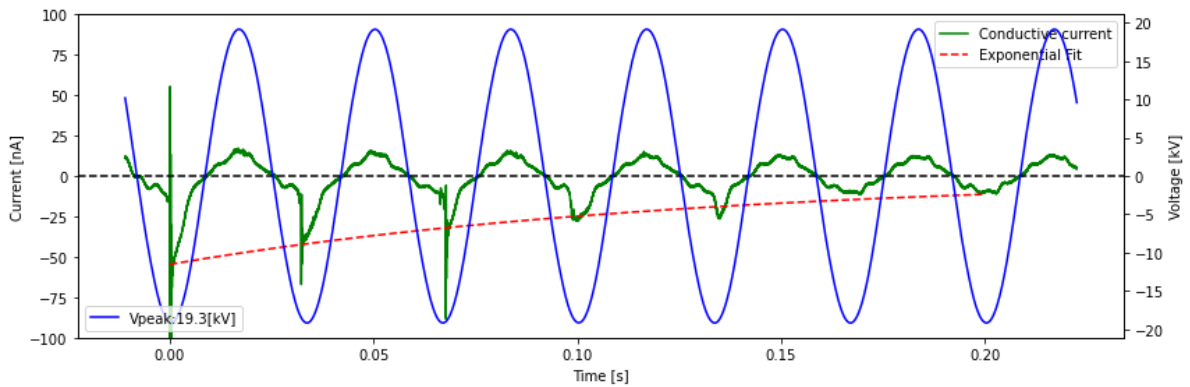


Figure 4.12: Conductive current and voltage waveforms after a large PD incident in Nytro 10XN at 30Hz

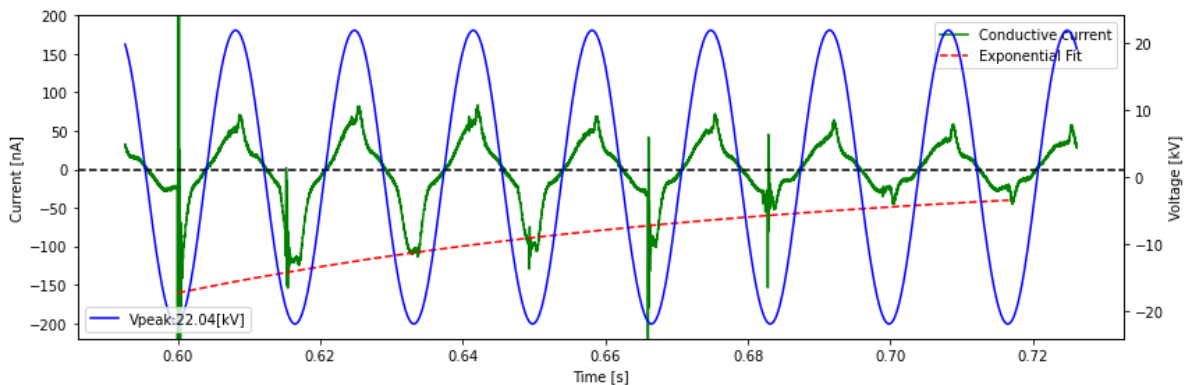


Figure 4.13: Conductive current and voltage waveforms after a PD incident in Nytro 10XN at 60Hz

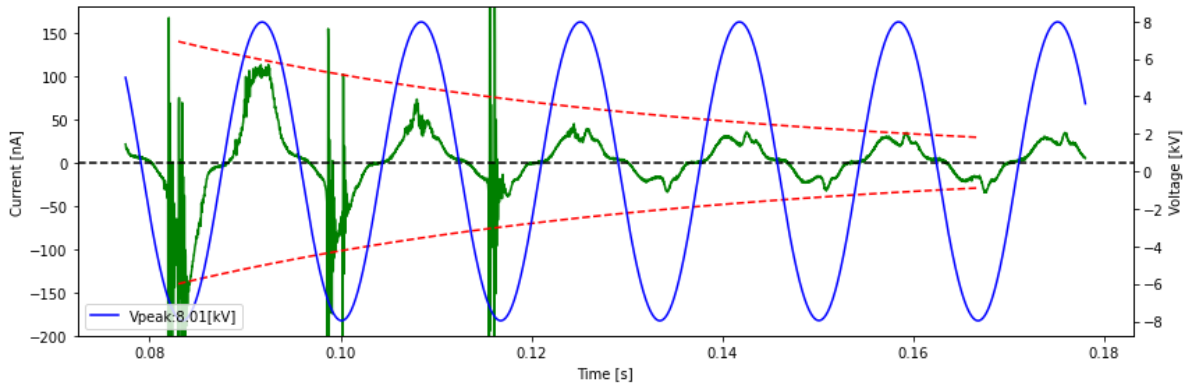


Figure 4.14: Conductive current and voltage wave forms after a PD incident in Cyclohexane at 60Hz

In Figure 4.13, 7 cycles of the conductive current have been shown in Nytro 10XN under 22.04 kV peak voltage and 60 Hz frequency and the Figure 4.15 shows the same trend in results for MIDEL under 19.8 kV. The memory effect for the conductivity current of the liquids after a large PD incident is because of the space charges present in the region for a short time. It later decreases with the same trend in all the tests. Finding an exponential decrease in the conductive current peaks is tricky in higher frequency since, in most parts of the results, it was not possible to find the proper situation where a couple of cycles would follow a large PD incident.

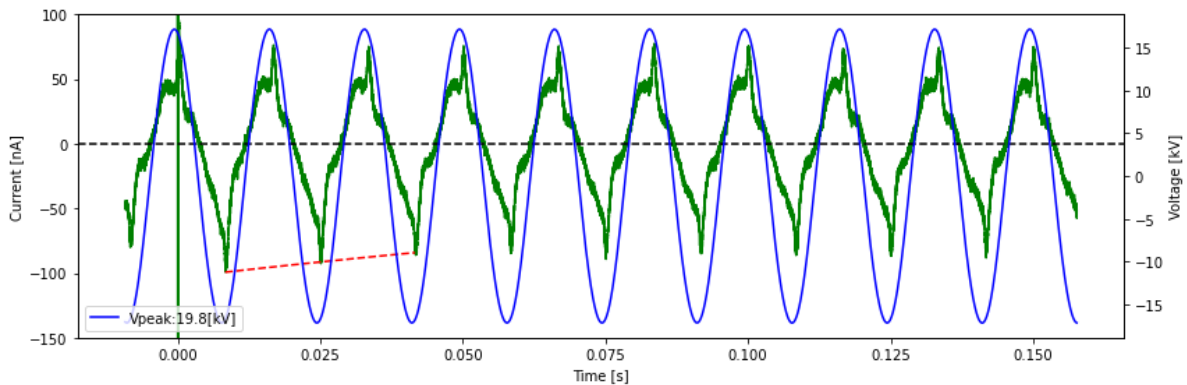


Figure 4.15: Conductive current and voltage waveforms after a PD incident in MIDEL 7131 at 60Hz

4.3. Effect of the Half-cycle Voltage on the Conductive Current

The test setup has been modified to be able to generate half-cycle voltages according to subsection 3.2.2. The modified test setup has two omicron measurement devices to measure the HV circuit from two points. (Omicron 1&2) The equivalent circuit of the test setup has been shown in Figure 3.7. The measured and filtered waveform of the voltage has been shown in Figure 4.16. It is obvious that the first quarter of the voltage waveform is sinusoidal. The second and third quarters decrease with the time constant of the load side in the half-cycle setup. It has been calculated in subsection 3.2.2. The half-cycle test is performed on Nytro 10XN at 45Hz. the peak voltage at the positive and negative half-cycle voltages are ± 19.5 kV. It is obvious that despite the filtered voltage which follows the measured voltage well, there are some differences at the discontinuity points. Figure 4.17 shows the measured voltage and current waveforms; voltage data has been divided by 50 to fit in the graph range, and current data in this graph has been recorded in volts since the output ratio of the preamplifier is 1 V per μ A of current. (see Figure 3.3)

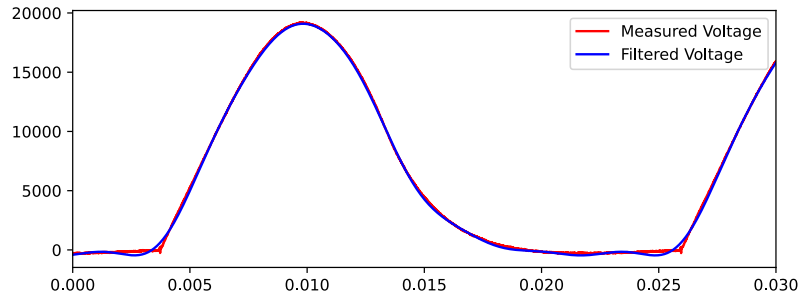


Figure 4.16: Measured and filtered voltage waveforms in positive half-cycle test $V_{peak} = 19.5\text{kV}$

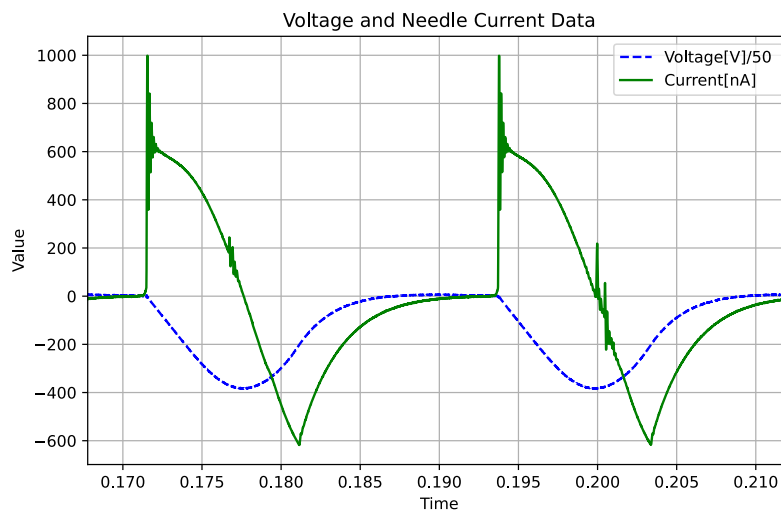


Figure 4.17: Voltage and Current Data of negative half-cycle test together

After adding the rectifier part to the test setup as the configuration in subsection 3.2.2, flipping the diode $D1$ in Figure 3.7 changes the polarity of the voltage, and only the negative half-cycle of the voltage will be applied to the test cell.

In Figure 4.18, the blue curve shows the applied voltage of the needle, the red curve shows the capacitive current calculated with FFT filter for eliminating over 300Hz components, the orange curve showing the measured current of the needle and the green curve shows the conductive current waveform resulted from subtracting capacitive current from measured current.

The conductive current has many fluctuations and has a higher range relative to the conductive current signal in nA range. Therefore, FFT filtering of the higher frequency is not effective in filtering out distortion from the signal.

The histogram of Figure 4.19 shows the noise patterns recorded by Omicron-1 and Omicron-2 during the half voltage test. The voltage waveform was measured at higher and lower HV arms individually. The green curve on all histograms shows the voltage waveform at the Omicron-1 connection point, and the blue curve on all histograms shows the voltage waveform at the Omicron-2 point of connection. The voltage waveform should be inverted since this shows the HV arm and the plane voltage, and the needle voltage is inverted. (see Figure 3.3)

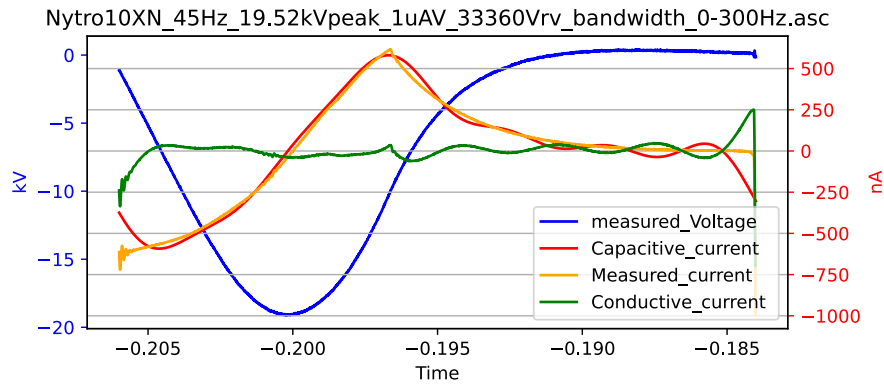


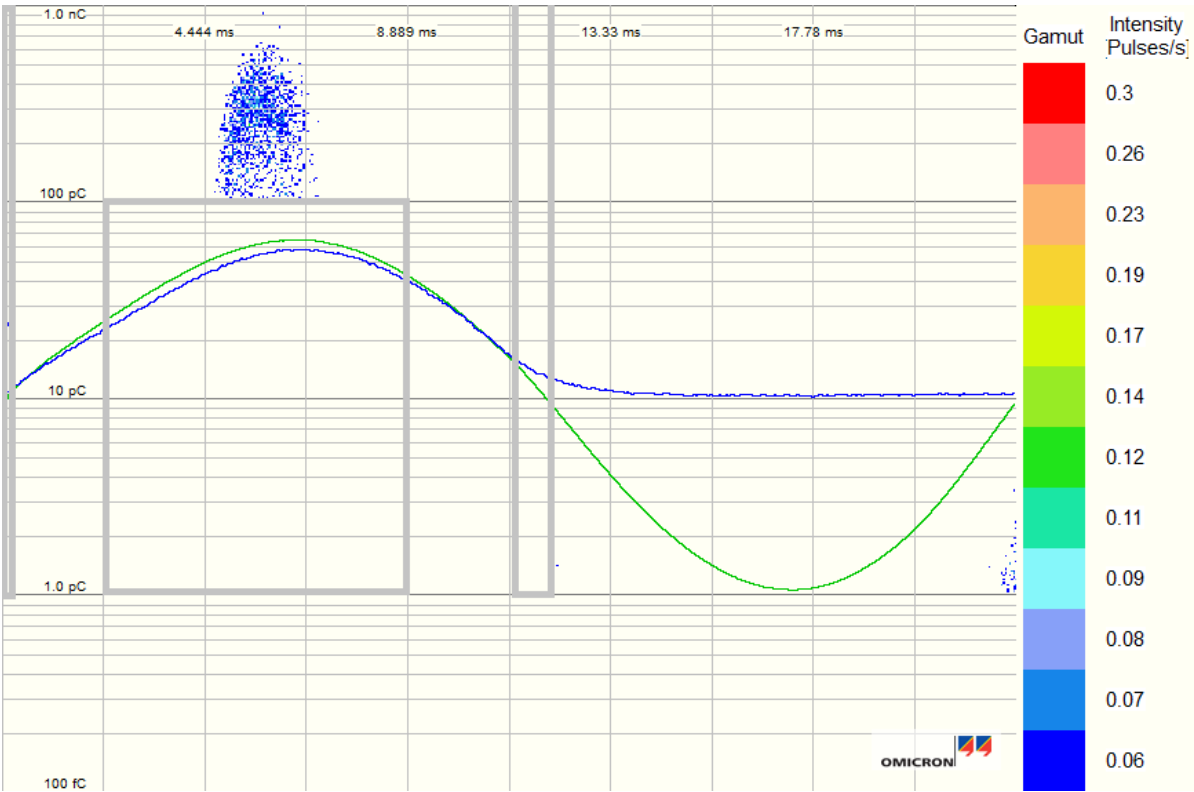
Figure 4.18: Measured voltage and current together with capacitive and conductive current calculated for negative half-cycle test



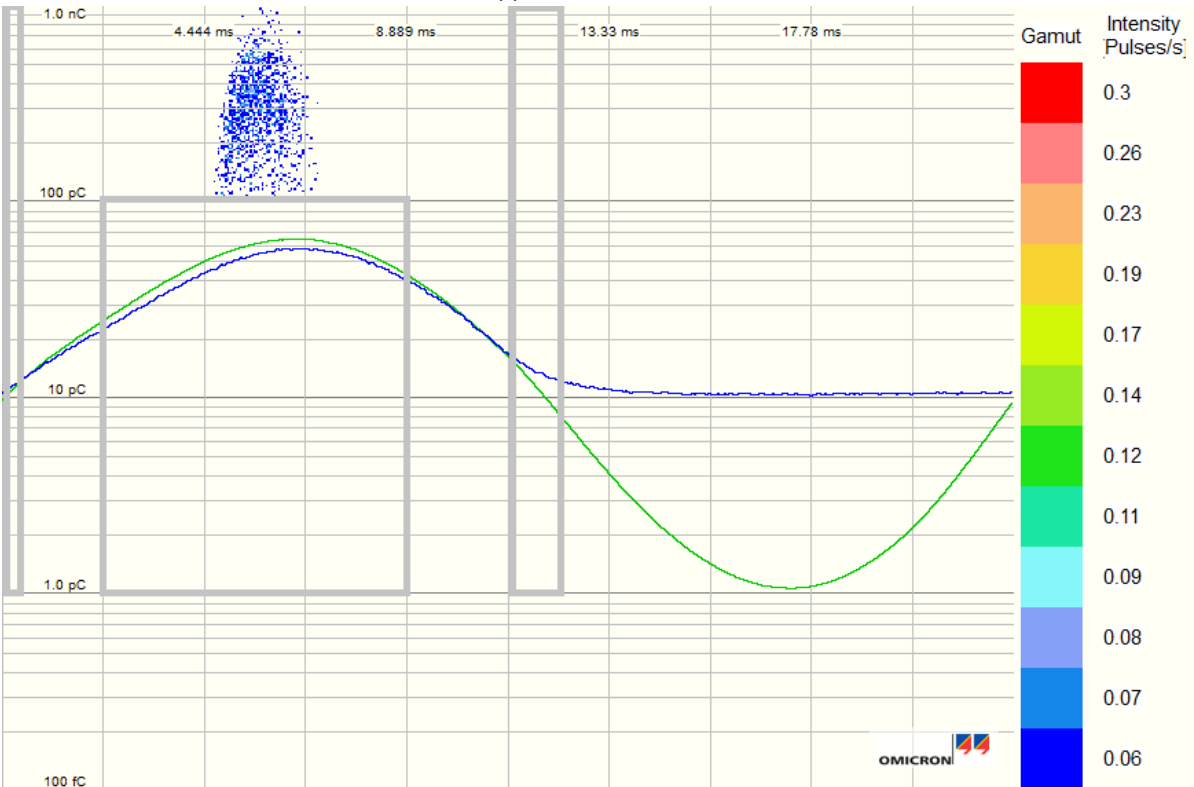
Figure 4.19: Histogram of the noise in negative half-cycle voltage test with Omicron 1&2

The histogram in Figure 4.19 shows the pattern of PD pulses happening inside and around the diode. Pattern 5 and pattern 6 in this figure are mostly related to switching high-frequency fluctuations. Pattern 3 and pattern 4 are due to the internal arrangement of uneven surfaces of the HV diode. Patterns 1, 2, 5, and 6 are easily detectable since they happen close to the zero crossing of the previous sinusoidal HV. Patterns 3 and 4 are in a tricky situation since they are happening around the voltage peak. Histogram of Measured PDs in the negative half-cycle test at $V_{peak} = -12.59\text{kV}$ for $t = 17\text{s}$ seconds has been shown in Figure 4.20.

The green curve in Figure 4.21 shows the voltage of the Omicron 1 at the lower arm. Obviously, the slope of the voltage waveform changes slightly at the moment of switching; therefore, there is a change happening in the resonance circuit of the HV source during the switching process. The blue curve shows the rectified voltage measured by Omicron 2 at the higher arm. It has successfully produced an acceptable waveform. There is a PD pattern excessive to the noise we expected for this experience. At the positive voltage peak (negative peak in the Figure 4.21) between 100pC and 1nC . The needle tip radius for this experiment was $3.33\mu\text{m}$, and after the test, the tip radius was $5.95\mu\text{m}$. It appears in a higher voltage than PDIV level with considering needle tip radius. The range of the green and blue curves is adjusted automatically; therefore, the vertical axis is only related to PD incidents and not to the voltage curves. Histogram of the measured PDs in the positive half-cycle test at $V_{peak} = 15.3\text{kV}$ for $t = 17\text{s}$ is in Figure 4.21.

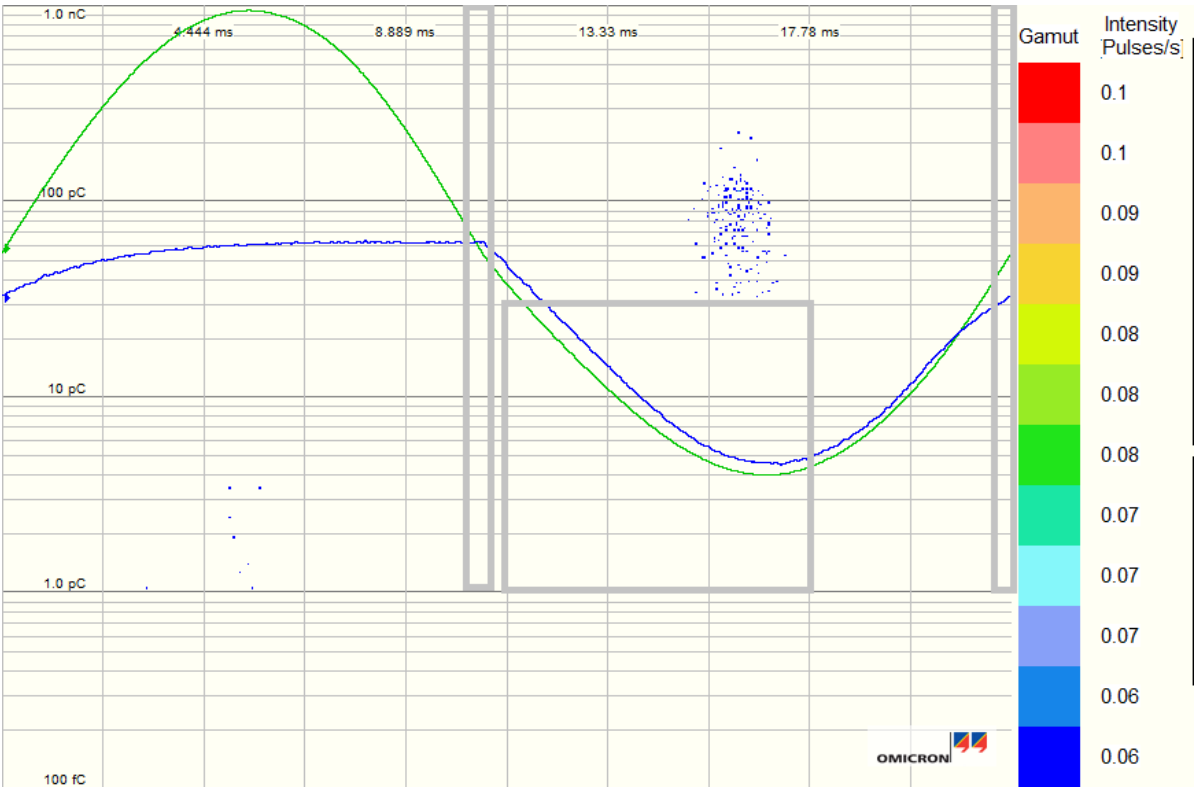


(a) Omicron-1 lower arm

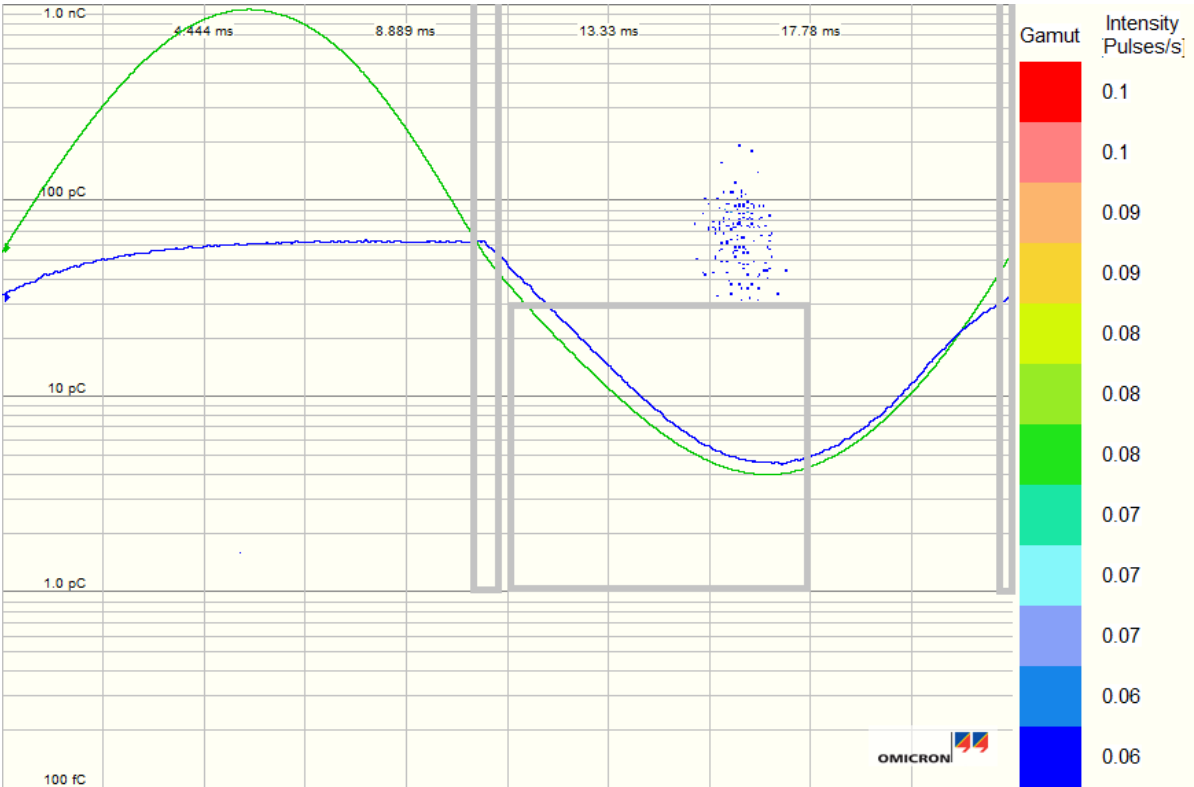


(b) Omicron-2 higher arm

Figure 4.20: Histogram of the negative half-cycle at $V_{peak} = -12.59kV$ (gray rectangle shows the filtered out area)



(a) Omicron-1 lower arm



(b) Omicron-2 high arm

Figure 4.21: Histogram of the positive half-cycle at $V_{peak} = 16.8kV$ (gray rectangle shows the filtered out area)

4.4. Needle Tip Erosion During the Tests

Previous studies show that the needle tip erosion changes when it is under HV tension. (see subsection 2.2.2) Referring to section 2.2, any changes in the needle tip radius affect the maximum electric field drastically in the needle-plane geometry. The rate and the magnitude of this change have been studied in previous research. The importance of those results lies in the geometry and dimensions of the test cell. The same test cell has been used for this study.

Different liquids are tested for needle tip radius change over time and the results have compared with previous studies. There are two hypotheses explaining the main cause of needle tip erosion. First, needle erosion can happen due to reactions with ions being created and dissolved in the region. High electric field effects that are explained in section 2.5, can happen slowly during conductive current flow. Second, PD incidents, including large current pulses happening in a very short time, can be the main cause of the erosion.

In this study, the needle erosion caused by the conductive current will be measured at different voltage levels, both above and below the exponential part of the conductive current (Region 3) and below the PDIV level. The test will be further taken above the PDIV level to the first PD burst. The needle tip radius is measured after the test.

The first observation related to the needle tip shows the change of the needle in time.

For this experiment, the tip radius of the needle is measured as the first action. Then after placement into the test cell, the total time that the needle is being under HV tension is being measured. then, the needle will be measured again, and the changes in the tip radius will be compared with the previous study.

The second experiment has been designed for differing between the conductive current and PD effects on the needle tip.

Tip Radius Change Before and After PD

In this experiment, the needle has been sharpened and adequately measured at the first step. Then, it was loaded into the test cell and was put under different voltage steps.

The voltage rose gradually, and the needle was investigated for any changes after being under that voltage step for several minutes. Step voltages go higher and higher until they surpass the PDIV voltage, and the last step is the step where the first PD has been observed.

Nytro 10XN has been chosen for this experiment since it shows relatively good insulation characteristics. 45 Hz has been chosen since it was closer to grid frequency and exhibited good conductive results in previous tests. The testing procedure was as follows:

1. Needle has been sharpened and adequately measured
2. The test has been taken in several voltage steps lower than PDIV
3. Tip radius has been measured
4. Higher voltage steps applied
5. Tip radius has been measured again

The needle tip radius has been sharpened again for the next test. Measuring the tip radius is being done with the help of a digital microscope, which has been shown in different examples such as in Figure 4.22.

Observations of the needle tip radius in the previous tests begin by considering tests on Cyclohexane with the tip radius of $1.7\ \mu\text{m}$ before tests. After tests, the needle was taken out and measured once again. which has been shown in Figure 4.22. The tip radius of the needle was about $6\ \mu\text{m}$.

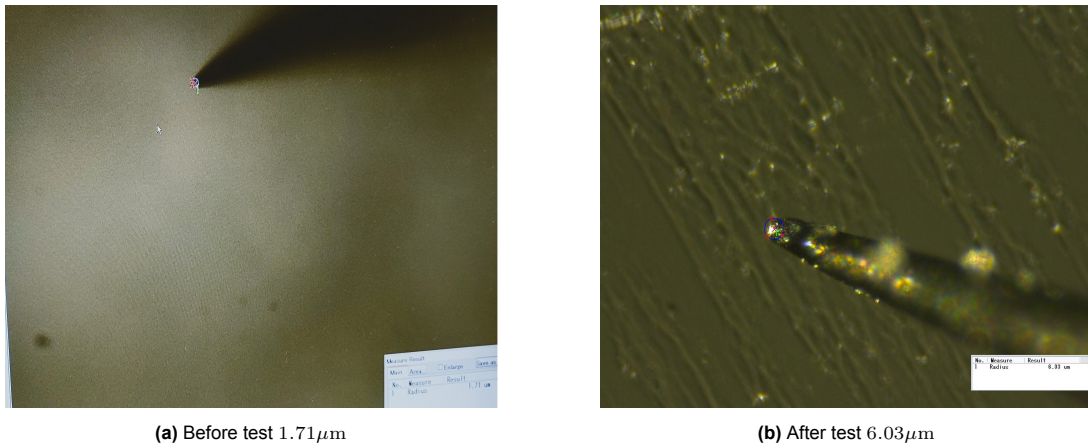


Figure 4.22: Needle tip radii before and after testing Cyclohexane with 30, 45 and 60Hz

Needle tip radii of Nytro 10XN tests is presented in Figure 4.23. The tip radius of the needle in these tests was $5.24 \mu\text{m}$ at the start and $11.6 \mu\text{m}$ at the end.

Needle tip photos for tests with Nytro 10XN have been shown in Figure 4.23.

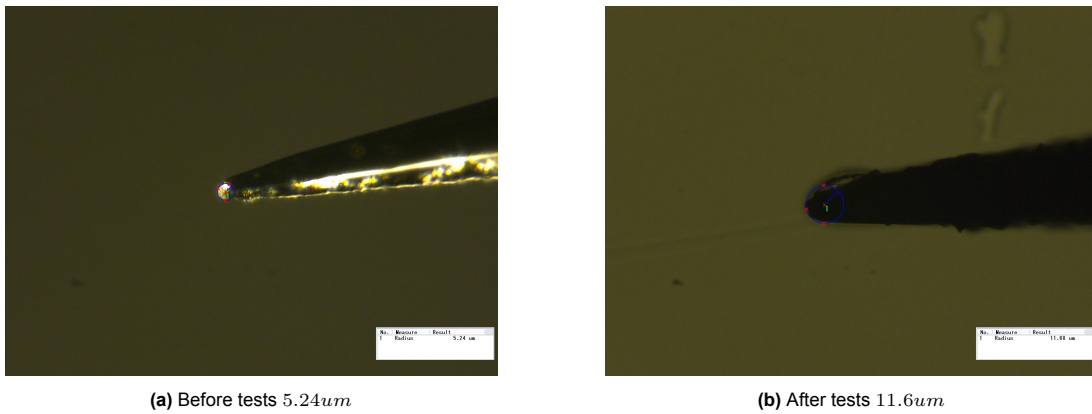


Figure 4.23: Needle tip radii before and after testing Nytro 10XN with 30, 45 and 60Hz

The needle tip has also been measured in testing with MIDE L 7131. The tip radius of the needle before tests was $5.26 \mu\text{m}$, and after tests, it was $10.4 \mu\text{m}$.

Needle tip photos before and after tests with MIDE L 7131 have been shown in Figure 4.24.

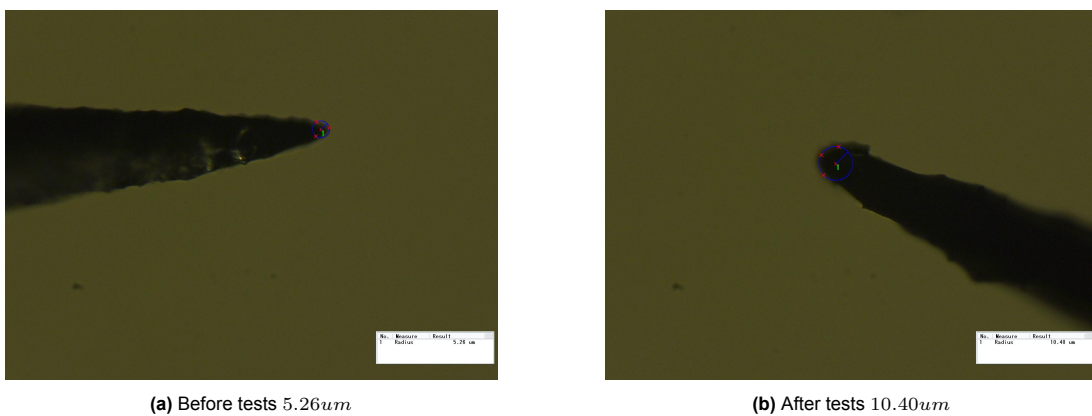


Figure 4.24: Measurement of the Needle tip radii before and after testing MIDE L 7131

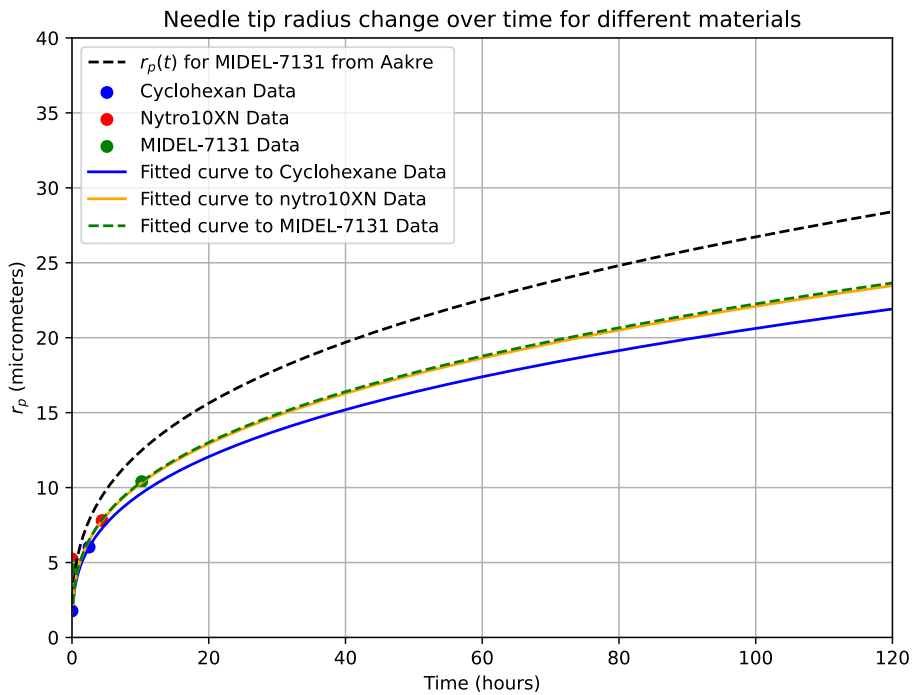


Figure 4.25: Needle change over time in different liquids and the fitted curves

It has been observed that needle erosion occurs while putting it under the HV tension, as presented in the previous study. Recording time in Omicron has been used to calculate the accumulative time span of each setup. However, for this part of the work, the tip radius has been measured before and after each liquid test group.

Resulted data points have been fitted to several cubic root functions and have been compared with previous study in Figure 4.25. Data points in this graph are pushed to the left side since the testing time was limited due to practical considerations; however, to show the difference between the curves properly, it has been extended to 120 hours.

A test has been planned to differ between the erosion caused by PD incidents and by conductive current.

The needle tip radius has been measured before and after each voltage level this time. The needle tip before test has been shown in Figure 4.26a.

It has been installed in the test cell with Nytro 10XN and with a gap distance of 20mm. The test cell has been connected to the voltage source with 45Hz.

The voltage has been increased gradually in the first test. After increasing the voltage to 2.15 kV, it has been applied for 10 mins. The needle tip was inspected after that. Photo has shown in Figure 4.26d.

In the second test, the voltage has been increased to 9 kV applied for 14 mins. The needle tip after the test has been shown in Figure 4.26b.

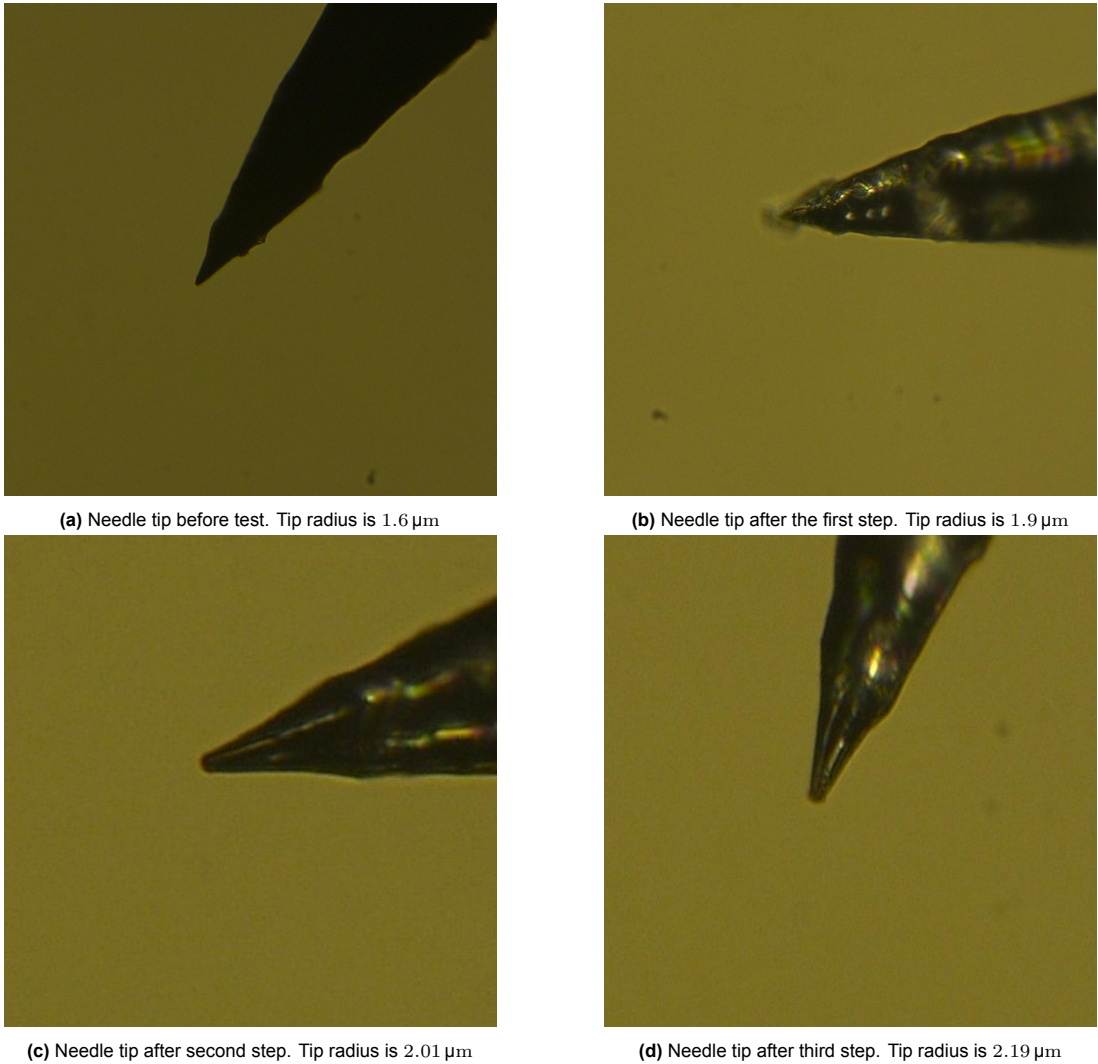


Figure 4.26: Needle tip measuring after each voltage step for Nytro 10XN at 45 Hz

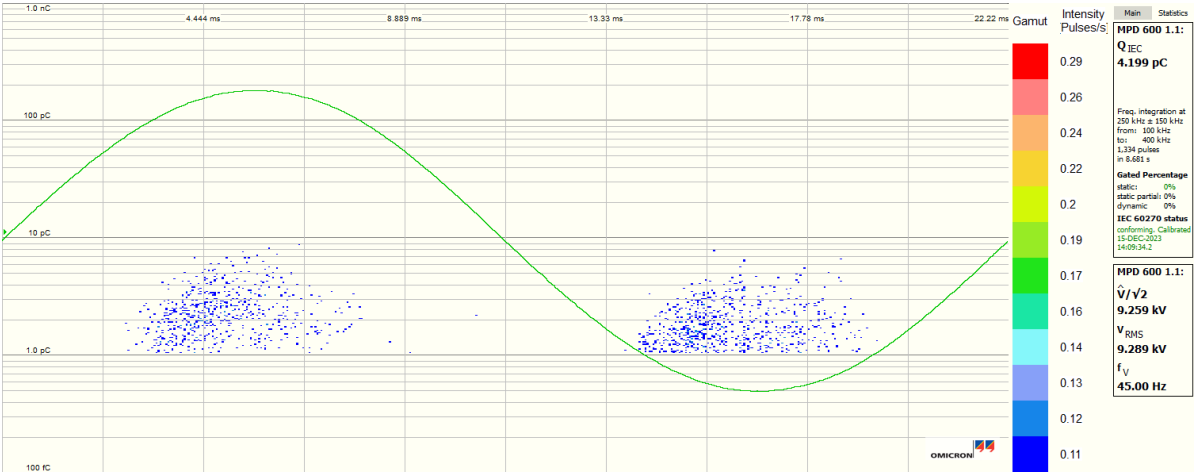


Figure 4.27: Histogram of testing Nytro 10XN at 45Hz for needle tip inspection

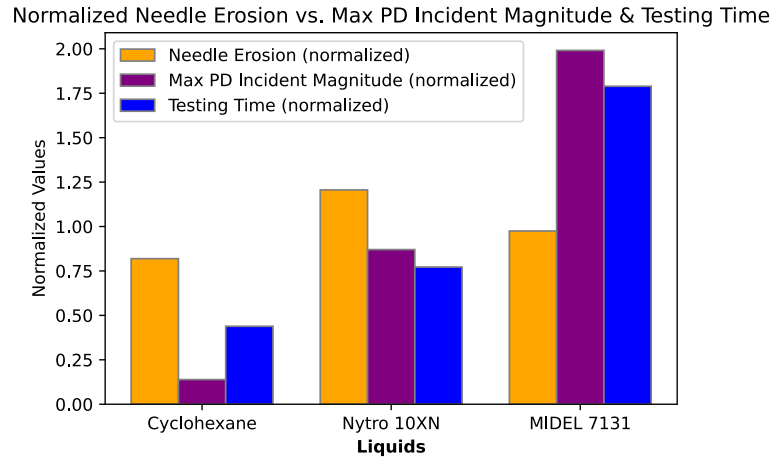


Figure 4.28: Normalized needle tip erosion, testing time and maximum PD incident magnitudes

Subsequently, the test proceeded beyond PDIV, to 9.3 kV and was halted after the first PD burst. The needle tip radius was then measured, as depicted in Figure 4.26c.

The pattern of the first PD burst has been shown in Figure 4.27. There were 1334 pulses measured in 8.6s until the voltage has been cut off. The needle tip inspection result has been shown in Figure 4.26c. The results of these two tests indicate that needle tip erosion occurs during conductive current flow in the liquid, but it predominantly happens during PD incidents. Significant changes in the needle tip radius are observed over longer periods in the first test under HV.

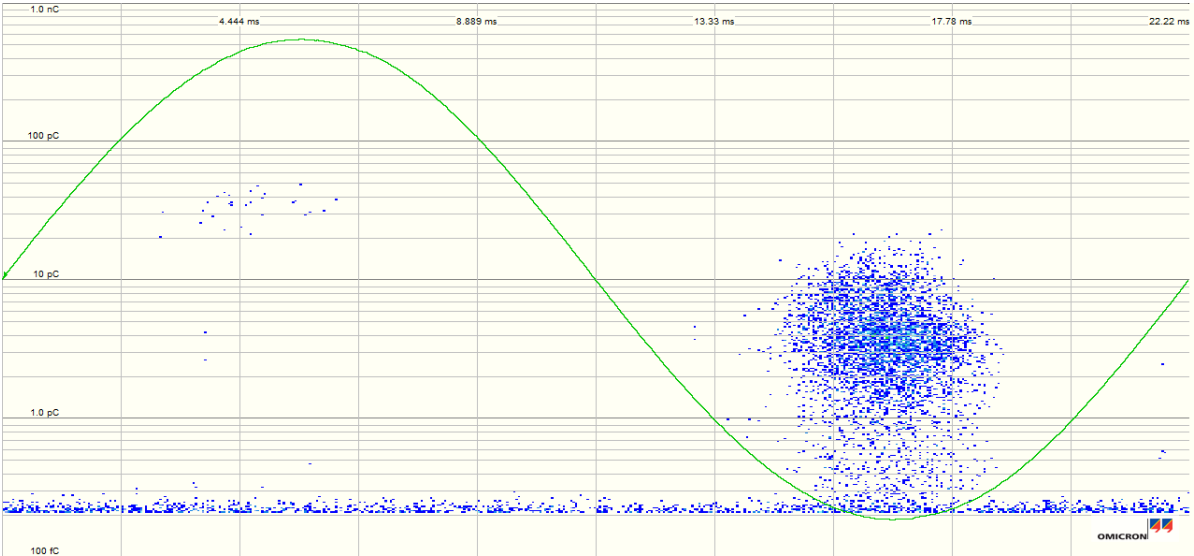
In previous results, considering the PD patterns in Cyclohexane in Figure 4.1, the largest PD incidents are around 30 – 70 pC, the PD patterns in Nytro 10XN in Figure 4.2, the largest PD incidents are over 100 pC and for MIDEL 7131 in Figure 4.3, the largest incidents are around 500 pC. Comparing these PD incident magnitudes with needle tip erosion has been shown in a bar graph in Figure 4.28. The overall result shows that the needle tip erosion is different in different liquids and is less related to the testing time and Max PD incident value.

4.5. PDIV of the Liquids Under Test

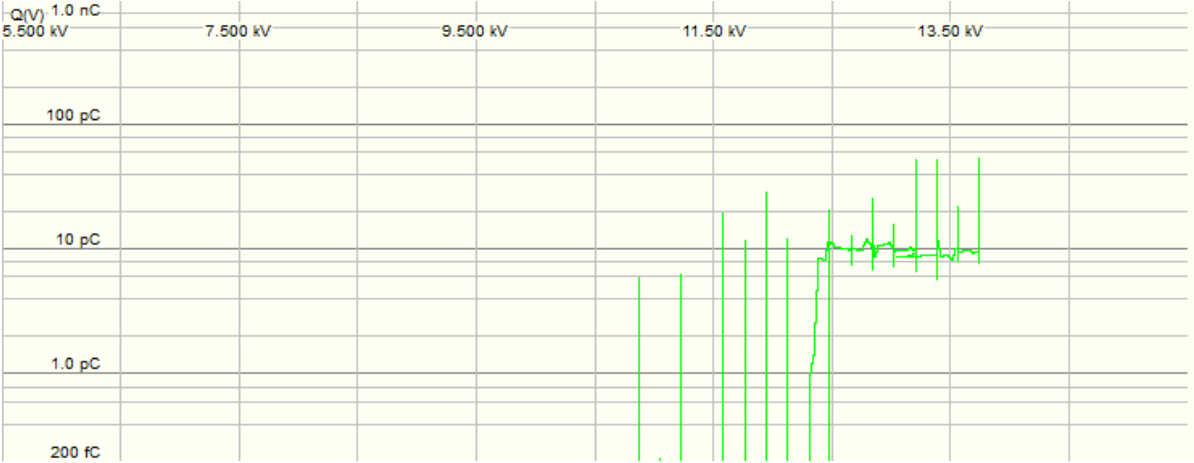
According to explanations in section 3.7, PDIV level is not a fixed threshold. Therefore, a specific method should be defined for measuring correct PDIV. According to the defined procedure in section 3.7, The voltage step of the first partial discharge incident larger than 2 pC is considered as the PDIV level. The voltage will increase gradually in steps. Every step magnitude is 250 V and is applied for 1 minute.

The histogram of the PDIV test for Cyclohexane in 45 Hz has been shown in Figure 4.29a. It shows the pattern of PD incidents happening in the earliest burst after passing the PDIV voltage. The voltage has been increased to 13.8 kV_{RMS} for Cyclohexane, and the PD pattern is recorded for a period of time after PDIV. The PD charge versus voltage has been shown in Figure 4.29b Nytro 10XN and MIDEL 7131 have also been tested after Cyclohexane. The charge versus voltage and histogram have been shown for each one in Figure 4.30 and Figure 4.32. The results of PDIV tests have been shown in Table 4.1. Table 4.1 shows the PDIV levels for each liquid in 45 Hz.

Figure 4.31 shows the bar graph of the PDIV levels in tests with Cyclohexane, Nytro 10XN and MIDEL 7131.



(a) PD burst pattern

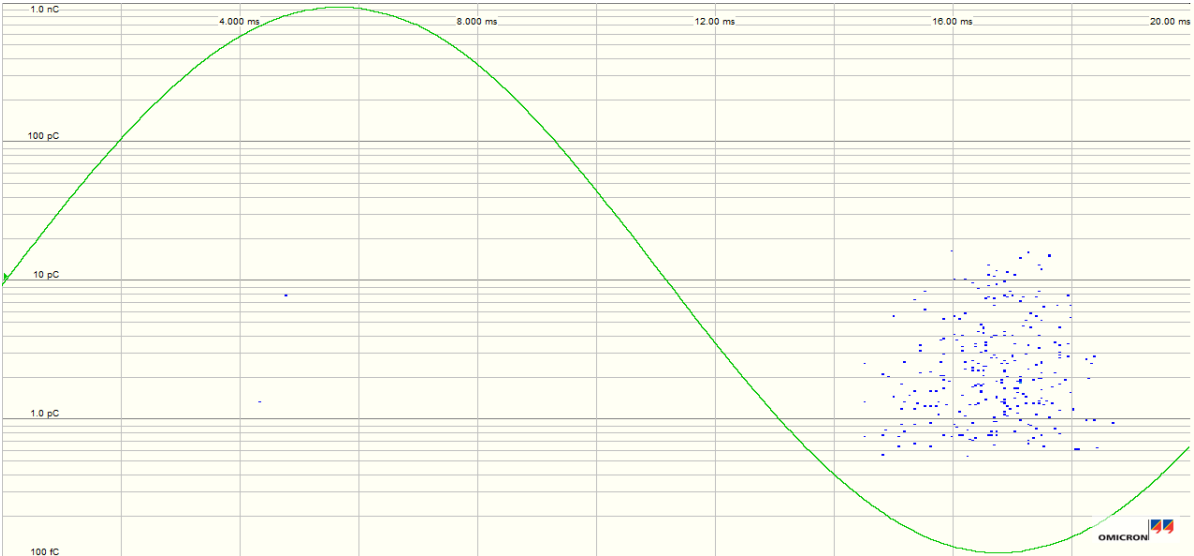


(b) PD charge versus Voltage

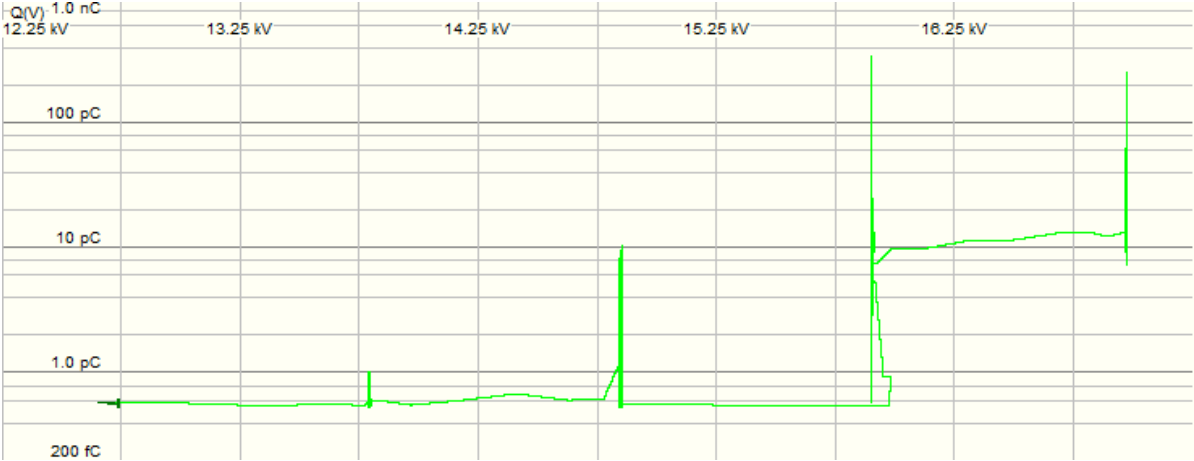
Figure 4.29: PDIV histogram for the Cyclohexane at 45 Hz

Table 4.1: PDIV Voltage Level of tested liquids

Liquid	Frequency	Needle tip radius	PDIV level
Cyclohexane	45 Hz	2 μ m	10.85 kV
Nytro 10XN	45 Hz	5.3 μ m	14.85 kV
MIDEL 7131	45 Hz	4.5 μ m	13.76 kV



(a) PD burst pattern



(b) PD charge versus voltage

Figure 4.30: PDIV histogram for Nytro 10XN at 45 Hz

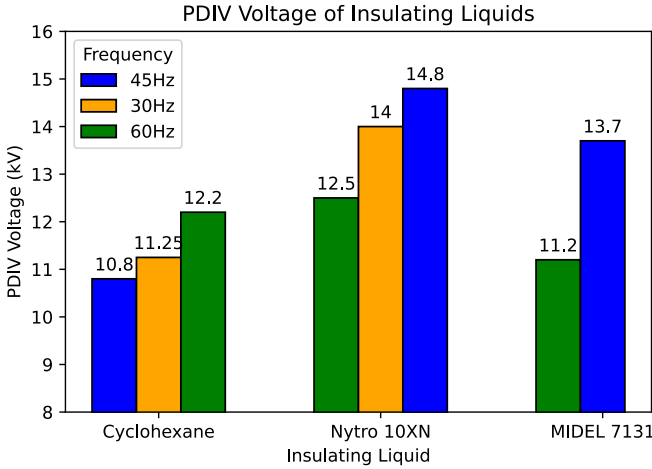
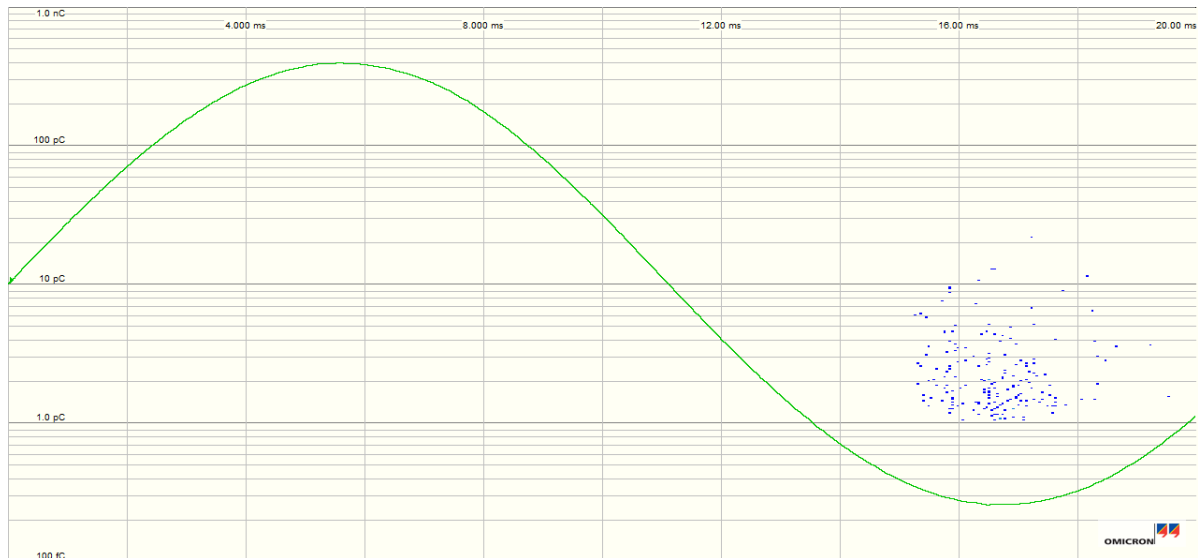


Figure 4.31: Bar graph of PDIV of tested liquids



(a) Histogram showing the PDs happened after PDIV level



(b) PD charge versus voltage

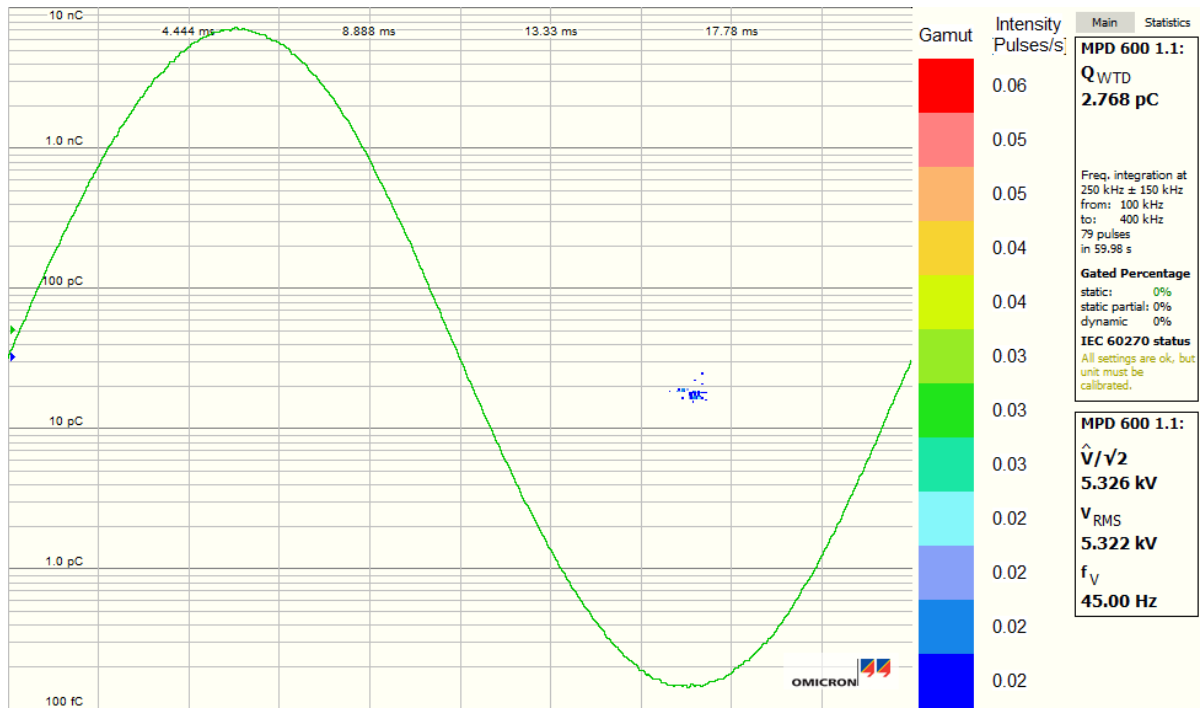
Figure 4.32: PDIV MIDEL 7131 at 45 Hz

4.6. PD Detection By Optical Method

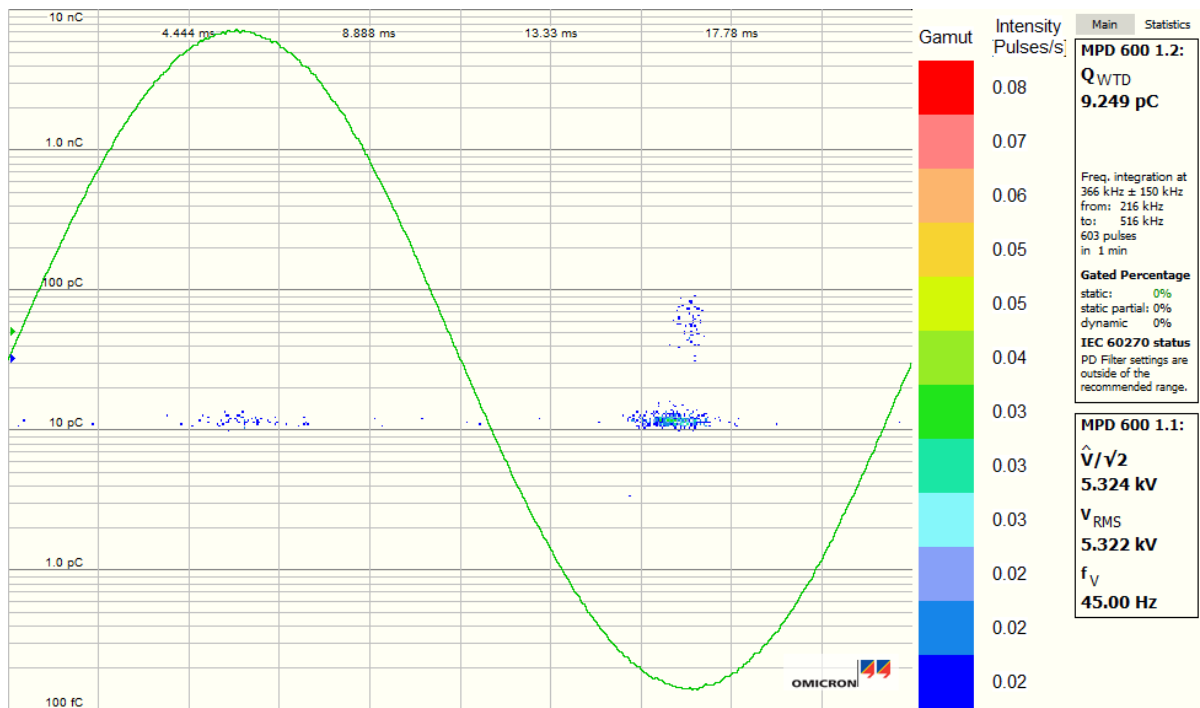
The optical PD measuring system has been explained in section 3.5. An experiment has been conducted to investigate the initiation stage of the PDs happening in the air from a wire which has been connected to the HV arm inside the test cell.

To facilitate this investigation, the test cell was removed from the test chamber, and a sample wire was positioned on the HV in the air. Subsequently, the voltage was gradually increased to observe PDs. Specifically, an RMS voltage of 5.26 kV at a frequency of 45 Hz the first PD incidents occurred. The voltage was kept constant for a duration of 60 s, and the PD pulses were recorded both electrically through an omicron connected to CPL542 and optically through a second MPD600 omicron module connected to the output of the gate module of the coincidence circuit (see section 3.5). The resulting histogram of the test has been shown in the Figure 4.33.

Furthermore, Figure 4.34 shows the same wire has been tested under RMS voltage of 6.32 kV at the frequency of 45 Hz for the same duration of time.

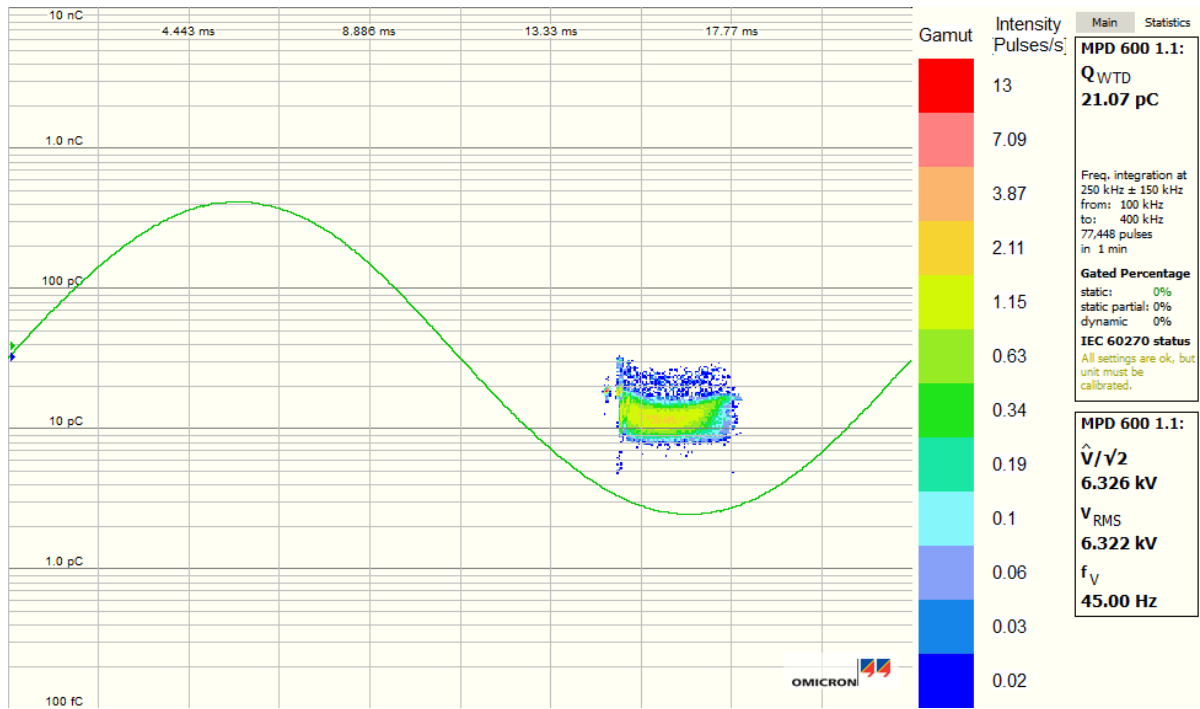


(a) Histogram of the electrical PD pulses detected at 5.26kV for 1minute

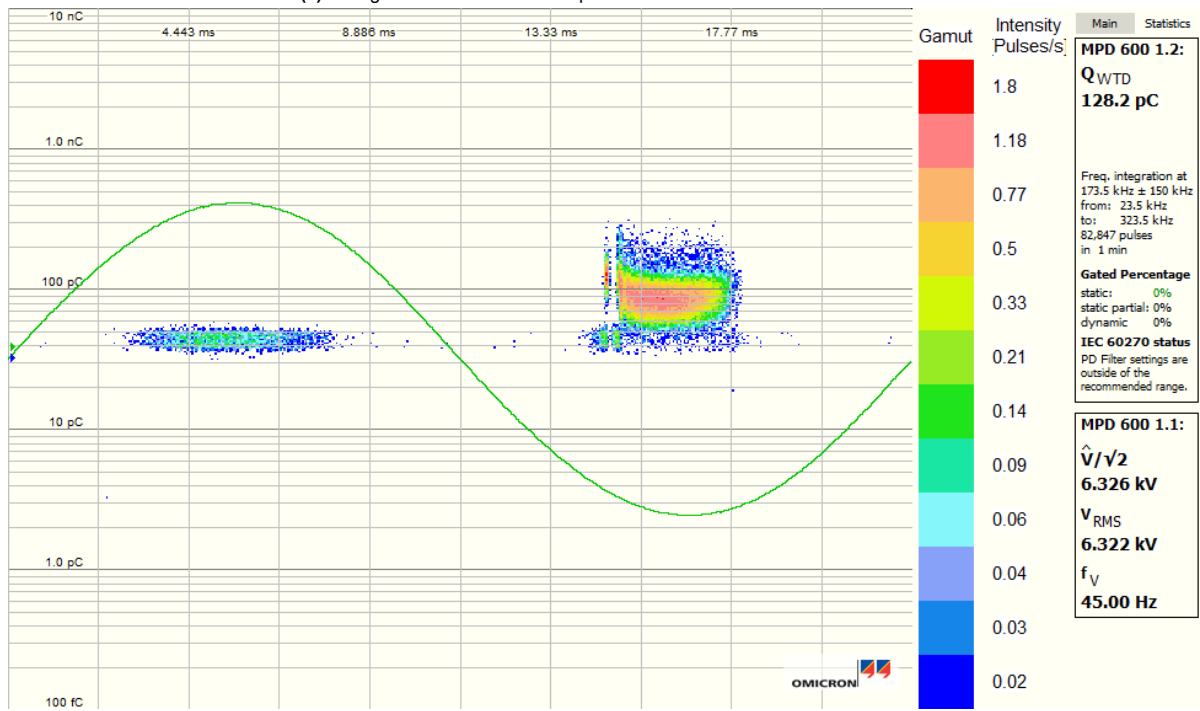


(b) Histogram of the optical PD pulses detected at 5.26kV for 1minute

Figure 4.33: PD measurements by electrical and optical methods around PDIV in air at 5.26 kV



(a) Histogram of the electrical PD pulses detected at 6.32kV for 1minute



(b) Histogram of the optical PD pulses detected at 6.32kV for 1minute

Figure 4.34: PD measurements by electrical and optical methods around PDIV in air at 6.32 kV

5

Discussion

The mechanism of creating a partial discharge in the liquid causes the formation of space charges in the insulating liquid, as mentioned in section 2.5. The effect of these space charges is discussed in section 2.6. The following part intends to analyse the results from the observations in chapter 4 with the given insight from the space charge mechanism mentioned in chapter 2.

5.1. Explaining Space Charge Effect

The concept of space charges as an important phenomenon that initialises partial discharges in insulating liquids is the most interesting part of this work. The quantity or the amount of charge of the space charges in the vicinity of the needle cannot be determined without studying the conductive current. Hence, they become detectable by observing the current flow in this region because of the effect of moving charges and the electric field change carried out in the needle tip. Note that a method to measure space charges directly in practice is missing. Still, the effect of these charges on the field and conductive current flowing in the insulating liquid identifies the presence of these charges.

The effects of space charges in the region were explored previously and investigated by obtaining conductive current from tests with different voltage waveforms as in [23]. However, it was difficult to effectively filter and extract conduction current at high frequency in that study due to the switching of the AC source. As described in section 2.5, various mechanisms effectively produce space charges in the region. The speed with which these space charges move is quite considerable, hence the impression that in previous experiments with low frequencies, it was not possible to catch effects because they are moving fast.

Successfully performing the tests at higher frequencies by changing data filtering methods and using a resonance AC source for the tests in this work made it possible to ponder the nature of space charges in the liquid deeply.

Some analytical results from experimental relations, like fitting data points, have been established based on the theories in section 2.6.

5.2. Space Charge Effect on Conductive Current Shape

Analysis of the resistive current extracted from specific sections revealed a distinct pattern of asymmetrical half-cycles in most tests. Notably, this asymmetry appears to correlate with the strength and frequency of the electric field.

To check space charges, it should be noted that the Equation 2.15 shows that the presence of space charges in the region causes current. This current depends on the mobility and strength of the electric field in Equation 2.16. Considering a half-cycle of the conduction current waveform in MIDEAL 7131 at a voltage range of about $V_{peak} = 10\text{kV}$ and with the same test conditions and at three different frequencies that are part of the results in the last row of Figure 4.32b and shown in Figure 5.1. The normalised time in this graph has been obtained by

$$\text{Normalized time} = \frac{\text{data time}}{\tau} = \text{data time} \cdot \text{frequency} \quad (5.1)$$

Since data has been shifted in time for 0.02, the peak of the voltage is at 0.27, and all current curves are shifted as well.

By comparing the voltage waveform, the two-quarter cycles are clearly different, and this difference becomes more apparent by applying a higher frequency. This difference is due to the difference in the effect of space charges formed in the region. If the space charges in the strong electric field region are considered the red circle in Figure 5.2. It creates in the region and moves towards the cathode because of the strong electric field. In Figure 5.2(b1), the needle has changed the polarity, and the direction of the electric field has also been changed, but the space charge is outside of the strong field region; therefore, the new electric field has less effect on the speed and direction of it. Therefore, it also causes smaller fluctuations in the current.

By increasing the Frequency, the formed space charges in the region have less time to move outside

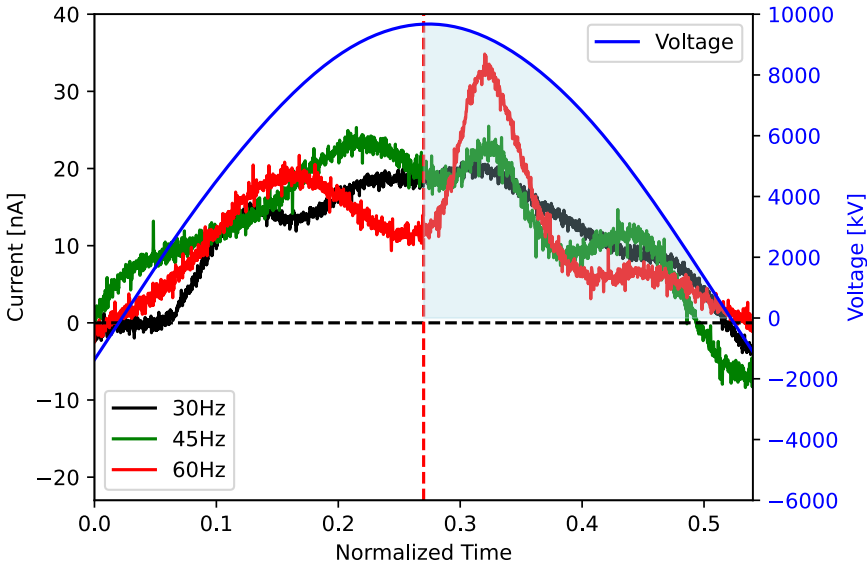


Figure 5.1: Asymmetry in the conductive current of MDEL 7131 at different frequencies

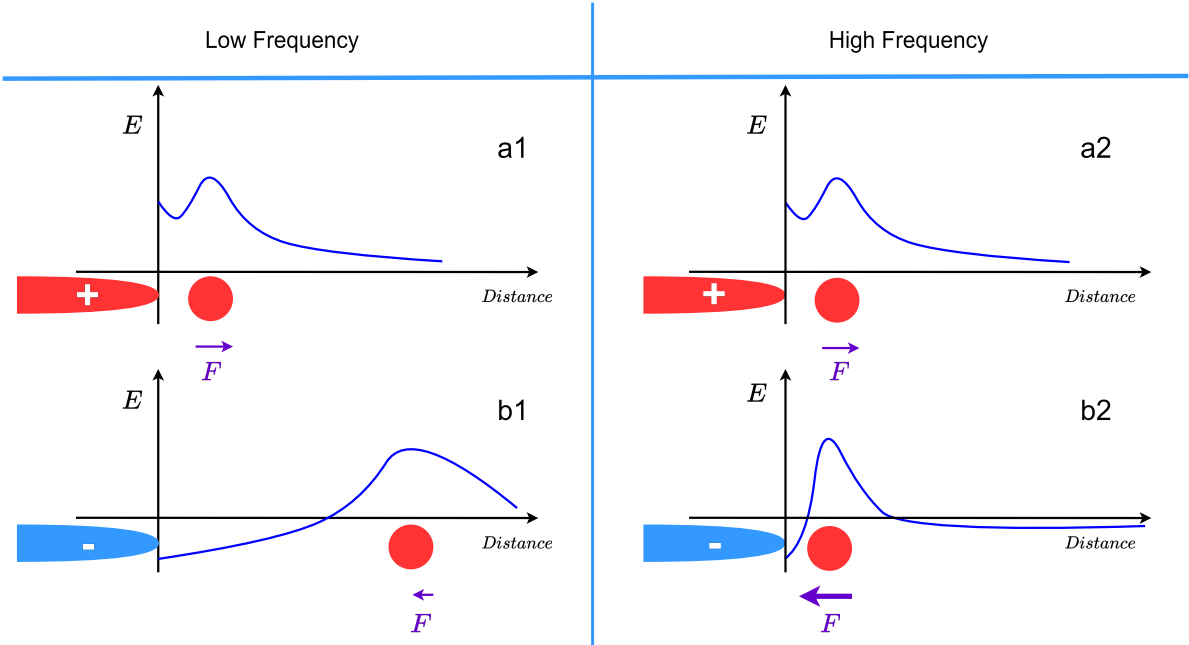


Figure 5.2: Drawing showing space charges remaining in the strong electric field region

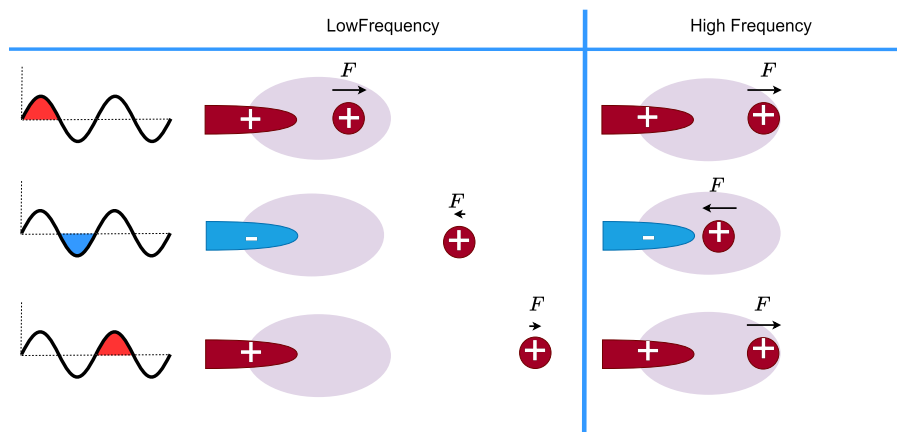


Figure 5.3: Drawing showing the behaviour of space charges moving in the strong electric field region

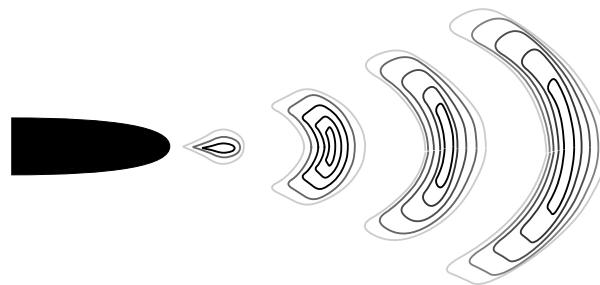


Figure 5.4: Drawing Showing different areas of space charge density around the needle tip

the strong field region; therefore, changing the polarity of the needle and electric field has a big effect on the direction and speed of the space charge. This leads to a huge local overshoot in the second quarter of the conductive current waveform, repeated every cycle and for the negative half-cycle. In the following, assuming that the frequency of the electric field change is large enough to be rival to the speed of movement of space charges in the region, at the same time, the forces applied to space charges and their displacement in the region can change constantly, be in the form of Figure 5.3. It can be seen that increasing the frequency to a certain extent can limit the mobility and range of movement of space charges within the region. This is the cause of different wave shapes in the conductive current waveform.

Considering the movement of space charges as a longitudinal wave propagating from the tip of the needle with drift velocity and the changing electric field in the form of another wave with a certain frequency and with a much higher propagation speed than that. The theory of interference of longitudinal waves may be used for analysing the placement of different areas around the needle with different densities or different polarities of space charges like in Figure 5.4.

5.3. Space Charge Effect on the Conductive Current Amplitude

Two different parts with different conductivity (slope) are detectable in all results, such as peak current at different voltages in different liquids. (Figure 4.7, Figure 4.8, and Figure 4.9)

Region 1 at low voltages, the current shows a linear behaviour with voltage similar to Equation 2.17, but at further increasing fields, which causes the formation of space charges in the region; therefore, these space charges change the electric field inside the region, and hence, the conductive current becomes more nonlinear according to Mott-Gurney's law that has given in Equation 2.15.

The geometry considered for this relation is plane-plane. Hence, tests have been taken on a needle-plane geometry; therefore, there is no way to compare the values from the test using this relation. However, comparing what is available from these relationships, the trend of the conductive current demonstrates a clear agreement with the theory.

This shows that the form of electrodes affects the resistance value between conductors, but the trend is the same.

The peak current waveform in all results changes earlier or in lower voltage by increasing the frequency. This agrees with the hypothesis that higher frequency causes more space charges to stay in the region around the needle tip. Therefore, the space charge effect can be seen earlier in the conductive current. On each of the three tests, the peak value of the main component of the conductive current has also been drawn side by side with the absolute peak current to be able to differentiate between the peak value of high pulses and the overall trend of the conductive current. For example, the negative part of the absolute peak current in Figure 4.8 and Figure 4.7 at 60 Hz are showing large deference due to PD incidents recording during the tests.

Focusing on the Figure 4.7, which shows the absolute value of the peak current for Cyclohexane in negative and positive peaks of the voltages. The reason for the large overshoot in the negative current peak at 60 Hz at a voltage above 7 kV is due to a PD incident that happened during the current measurement.

The only inconsistency that can be raised with the presented hypotheses is regarding the shape of the peak curve of Nytro 10XN in Figure 4.8. The reason affecting this test is the different tip radius of the needle between different tests for this liquid.

In the test at 45 Hz, a smaller needle tip radius ($1.8\mu\text{m}$) has been used other than the needle for 30 Hz and 60 Hz tests. ($5.3\mu\text{m}$)

This caused the difference in the electric field strength and the region in front of the needle, reflected in the graphs.

The results for the MIDEL 7131 in Figure 4.9 show the expected behaviour better since the test has been done with the same needle and it has been tried to keep the testing time as short as possible to lower the erosion of the needle tip.

The peak of the main component versus voltage for MIDEL 7131 shows the same trend in the Figure 4.9.

Based on the discussed results, the peak value of the conductive current is related to the electric fields and frequency of the test for all three liquids. Another factor that also plays a critical role in the peak current value is needle tip radius; therefore, in Figure 4.10, the peak current versus voltage steps have been drawn for Nytro 10XN with different needle tip radii.

It shows that the non-linear part of the conductive current related to region 2 at Figure 2.9 happens more or less in all tested insulating materials. Therefore, the amplitude of the conductive current is a function of the insulating material molecule, the strength of the applied electric field (tip radius and voltage), and frequency. But the field strength seems to be a more important factor since different liquids (Nytro 10XN and Cyclohexane) in Figure 4.10 and Figure 4.7 are showing quite close peak values at the same tip radius.

5.4. Residual Space Charges and Their Impact on Subsequent Current Waveforms

In further exploring the impacts of space charges on the measured current, it's noteworthy to mention previous findings indicating a link between the peak current of the conductive current in the cycles following a PD incident.

As depicted in Figures 4.11 to 4.15, the conductive current exhibits a notable alteration in peak current following a significant PD event. Specifically, the behaviour observed is characterised by an increase in peak values of the conductive current of the negative half-cycle of the following waveform. It follows an exponential decrease in most of them.

Obtaining the theory of the mobility of space charges from the region makes it possible to fit analytical curves to the current peaks in the subsequent PD cycles.

The exponential factors derived from the fitted exponential curves according to Equation 5.2 for each graph are elucidated in the Table 5.1.

Remarkably, the influence of space charges mostly impacts the negative half-cycles while leaving the

positive side unaffected. However, in Figure 4.14, The same trend has also been detected on the positive side.

$$y = a_{fit} \cdot e^{b_{fit}x}, \tau = \frac{1}{b} \quad (5.2)$$

Table 5.1: Exponential function Values and Time Constants

Liquid	Frequency	a	b	τ [ms]
Cyclohexane	30Hz	-107.54	-7.71	129.7
	45Hz	-27.96	-8.47	118
	60Hz	± 660.41	-18.68	53.5
Nytro 10XN	30Hz	-54.65	-7.89	126.7
	60Hz	-204154.8	-11.91	83.9
MIDEL 7131	60Hz	-103.51	-4.95	202

5.4.1. Needle Tip Radius Change Over Time

The experiments demonstrate that the test results are susceptible to variations in the tip radius of the needle. Because the tip radius of the needle plays a decisive role in producing the highest electric field in the needle-plane region. From section 2.2 resulted that electric field strength depends more on the curvature of the needle tip radius than its distance from the plane. The maximum field strength is of the most importance for the PD test in needle-plane geometry.

Following the study done by T.G.Aakre, the tip radius of the testing needle changes over time during the test [23]. A test procedure has been developed to inspect the needle tip radii over time. It has been described in detail at subsection 2.2.2. Tests have been done on three different liquids, and results have been shown in Figure 4.25. In addition, a cubic root function was fitted to the data from each material separately. Cubic root function has been suggested in previous studies. Therefore, this form of function has also been used in this study. The mathematical form of this function is at Equation 2.28, and different constant parameters obtained for each curve have been gathered in Table 5.2

Table 5.2: Constant Value of the Cubic root function fitted

Liquid test Data	Constant value
Previous study MIDEL 7131	400
Cyclohexane	183.63
Nytro 10XN	255.8
MIDEL 7131	231

The final result from the needle tip radius inspection in section 4.4 defines that the needle tip radii are unaffected by the testing liquid according to the results of the three voltage levels before the PD incident in Figure 4.26a, Figure 4.26d, Figure 4.26b and Figure 4.26c.

When the conductive current has been passed through the liquid. In the voltages above PDIV, the undiluted facts appear to be that PD events with amplitudes smaller than 10 pC do not substantially impact the alteration of the radius of the needle tip. Therefore, when compared, 1334 PD incidents under 10 pC that happened in 8 s with the fact that testing Cyclohexane with around 150 minutes of high voltage with several PD events of more than 10 pC significantly alters the needle tip. In conclusion, The needle tip is unaffected by such a test as long as the PD events are not so large.

5.5. Space charges Effect in Rectified voltage Regime

The half-cycle test aimed to examine the effects of space charges in the case of the field only in one direction. This observation is important in several ways

1. With this observation, the polarity of the formed space charges in the region can be found
2. It can also give a good insight into the contribution of positive and negative polarity of the needle to forming space charges
3. By comparing the results of this test, which only includes the field in one direction, with the previous results in two directions, space charge movement can be investigated more

The measured and filtered voltage waveform has been presented in the Figure 4.16. The filtered voltage can follow the measured voltage by detecting higher-order frequencies. Still, unfortunately, it does not follow the cracked point in the voltage waveform very well when turning off the diode D1. The reason for this can be the limited ability of FFT filtering in nonsinusoidal cracked points of the waveform. The data from the digitiser before being processed looks like Figure 4.17. It clearly shows some high-frequency oscillations due to parasitic elements in the circuit, which can also be detected in the histogram at Figure 4.19. Patterns 1,2,5 and 6 are considered to be because of the high-frequency noises due to the switching of the diode D1. There are two more parasitic PD patterns (3 and 4) due to the poor structure of the diode. They start at voltages higher than $V_{peak} = 10.43\text{kV}$, which could not be eliminated on positive or negative half-cycle tests.

It can be seen in the Figure 4.20 that the partial discharge at the negative peak of the voltage (which occurs at the positive peak in this histogram) with a size greater than 100 pC in the liquid, this partial discharge occurred at a higher voltage level than the corresponding partial discharge in the Full cycle mode. Correspondingly, a positive half cycle has been tested in the Figure 4.21 histograms, and in this test, partial discharges with a magnitude greater than 100 pC can be seen at the positive peak of the voltage (in the histogram it is displayed as a negative peak). Unfortunately, due to the high disturbance in the results of the histograms in both figures due to the structure and switching of the diode on the sinusoidal voltage waveform of the source, which can be seen in the green curve in Figure 4.19. It is impossible to draw a complete and reliable conclusion from this test, and it can only be briefly stated that the application of a half-cycle of voltage in both positive and negative states has increased the inception voltage of partial discharges in the insulating liquid.

5.6. Optical Detection of PDs

In the explanation of the experiment conducted in section 4.6, it was mentioned that two separate measurement methods were used simultaneously to measure partial discharge with electrical and optical methods. An electrical method based on ABC-circuit, which has been explained in section 2.9 and an optical method devised using PMTs and coincidence circuit described in section 3.5. But as stated in the description of the test in section 4.6, the voltage of the wire was raised step by step to reach the first PD pulse, and then the voltage was kept constant, and PD measurement continued for a minute. However, according to the existing assumptions, it is expected that the PD will start from the negative peak, and then it will be seen at much higher voltages in the positive peak of the voltage. Considering Figure 4.33 and Figure 4.34, the PD pulses happened approximately in the same phase angle (the delay is negligible), but the pulse amplitude in optical measurement is relatively higher than electrical. Considering that the optical pulse amplitudes are dependent on many aspects, such as the gain of the PMT, the shape of the pulse, and the equivalent resistance of the circuit from the PMT side. The MPD600 has not been calibrated to the PMT pulses; therefore, the amplitude of the optically measured PD incidents can not be trusted.

The results of the electrical measurement are in accordance with the assumptions, but the results of the optical measurement are contrary to the existing assumptions and show that the PD starts simultaneously at the negative and positive peaks of the voltage. The correctness of the output of the optical measurement circuit has been researched with the tests mentioned in section 3.5, and it seems that

the PD starts at the positive and negative peaks of the voltage simultaneously, but for some reason, in the electrical PD measurement, the first electric PD pulses appear only in negative voltage peaks. This test has also been performed at a higher voltage level, the results of which are shown in Figure 4.34. These results also show the presence of light pulses at the needle tip in the positive peak of the voltage, which was not recorded in the electrical measurement using the ABC-circuit method. This is a phenomenon that should be further investigated in future studies.

5.6.1. PDIV

The PDIV voltage of insulating liquids in different frequencies in Figure 4.31 shows a higher PDIV level for the Nytro 10XN than MIDEL 7131 as it is obvious in both 45 Hz and 60 Hz. The PDIV level for Cyclohexane is relatively lower than other liquids. One of the most possible reasons might be that the needle tip radius in Cyclohexane was smaller than Nytro 10XN and MIDEL 7131 during the tests. According to the results presented in section 4.4, the needle tip erosion causes considerable changes in the needle tip radius. The sorting of the bar graph in Figure 4.31 is according to the testing order, and it is obvious that the PDIV level is affected more by the needle tip erosion between tests other than frequency or molecule.

The patterns of PDs in figures 4.29a, 4.30a and 4.32a to compare with PD pattern of each liquid in Figure 4.1, Figure 4.2 and Figure 4.3. Comparing these two groups shows the same pattern for each liquid, thus validating the PDIV test in liquids.

An important result obtained from the half-cycle tests is that PDIV peak voltage and the magnitude of PD incidents have been increased in the half-cycle mode. This increase in the magnitude of PD pulses is more in the negative half-cycle, and it shows the effect of space charges in the region on the PDIV level.

5.6.2. Comparing Results With IEC 61294 Process

IEC 61294 is a standard that has been defined to find out the PDIV of insulating liquids. The procedure measures the inception voltage for a large PD with a voltage increment rate of 1kV/s.

Due to uncertainty in the PD activity, the PDIV test might not correctly reflect certain insulation rates for different liquids.

An example of this uncertainty in PD behaviour is evident in the experiments conducted in this research. According to the results of maximum conductive current in Figure 4.7, Figure 4.8 and Figure 4.9, it seems that the change in conductive current from region 2 to region 3 according to the conduction diagram in insulating liquids in Figure 2.9 happens at relatively same field strength at the field corresponding to 7kV. The PDIV level in liquids is considered to be proportional to the level of space charge formation voltage. Therefore, PDIV determination in the IEC standard is uncertain due to the different PD patterns of different liquids. Multiple testing is recommended in the IEC 61294 test procedure, but the PD pattern of the liquid is in favour of or against passing a dielectric liquid. The results of PD patterns for these liquids are shown in Figure 4.1, Figure 4.2 and Figure 4.3. The probability of 100 pC pulse occurrence is different in time. For example, for MIDEL 7131, in measuring the maximum PD in all three frequencies, the threshold level of the IEC standard has been exceeded, but for Cyclohexane, the PD occurrences are more often but all below the threshold determined in the IEC standard. In comparison, it is also important that the conductive current in the voltage range of 10 kV and frequency of 60 Hz in Cyclohexane can be much larger than the conductive current in MIDEL 7131 or Nytro 10XN, which of course is expected in this test with the radius of the needle tip being smaller in the Cyclohexane test. But contrary to the expectations, PD pulses are larger in MIDEL 7131, which seems to have caused inconsistent results regarding the determination of PDIV.

6

Conclusion and Future Work

6.1. Conclusion

This study has investigated the PD characteristics of different insulating liquids and the connection of space charges with the initialization of PDs. The main purpose was to experiment with the behaviour of space charges and investigate the effect of their presence on the initiation of PDs in both regular and newly developed insulating liquids.

Evidence of the existence of space charges during the formation of PDs in insulating liquids has been presented. It has shown, with the help of the half-cycle rectified HV source, that space charges include different polarities. It has also shown that the speed of movement of space charges does not depend on the frequency of the applied voltage and is more a function of the type of molecule.

The effect of space charges on conductive currents has been well studied and compared, and new paths have been created to investigate their nature. A more efficient method, using an FFT filter, is used to extract conductive current in insulating liquids. It has shown that the changes in conductive current caused by space charges are non-linear, and the presence of space charges caused by the PD event in the domain causes changes in the conductive current for some time. The effect of space charges on current waveforms has been successfully demonstrated in multiple cycles after a PD incident. This change decreases exponentially with the displacement of space charges.

Also, observations of the asymmetry between the first and second quarters of the cycle in higher frequency have been made. The theory of the impact of space charges on the asymmetry has been developed.

In addition, the results of practical tests on the PDIV level in three different dielectric liquids under three different frequencies have been reported.

The effects of applying voltage and PDs on the erosion of the needle tip in a needle-plane geometry have been investigated in two different aspects. The results have been presented and compared with previous research. It was shown that the erosion of the needle depends more on the type of material, on the rate of PD events, and on their magnitude than on the conductive current flow.

The optical partial discharge detection system has been set up and successfully tested using PMTs and a coincidence circuit. One of the goals achieved in this work is to prove the difference between the results obtained from the optical setup and the results from the charge acquisition detection. (ABC-circuit)

The IEC 61294 standard was criticized on the PDIV detection methodology by the results obtained from PD patterns of different insulating liquids.

6.2. Future Work

The research conducted has been an important step in the study of space charges in insulating liquids. To continue the research on space charges, more tests should be done around the causes of asymmetry of the conductive current waveform in insulating liquids. Further research on the PDIV surface. Another suggestion is to use a PDIV level test according to the IEC61294 setup and compare the results with this study.

There is a large research opportunity related to the optical measurement of PDIV. The question has now been raised as to what causes PD pulses to start at positive and negative half-cycles simultaneously and why the regular PD measuring system based on charge acquisition can not detect them.

Following the design of the rectified HV source in this study one can use this setup to further investigate the effects of different polarity space charges in dielectric liquids. Another suggestion for future work can be to use diodes without PDs to improve the rectified source to have less noise or use a different method to filter the data received from the digitiser in the presence of switching devices so that the conduction current can be calculated more accurately.

The use of optical PD detection, which has been demonstrated in detail in this work, can be taken as another option for testing with a half-cycle to neutralize the switching effect.

Another point for future work can be the use of an HV source with a higher capacity to maintain the perfect sinusoidal waveform during the switching procedure in half-cycle tests.

Many efforts were taken to filter the Nytro 10XN and MIDEL 7131 to $2\ \mu\text{m}$. despite using adequate filters and a vacuum system to force the liquid through the filters multiple times; unfortunately, due to the high viscosity and adhesion of the used liquids, there were some particles remaining in the test cell. This has not happened with testing Cyclohexane because of the comparably lower adhesion and viscosity, but for Nytro 10XN and MIDEL 7131, the filtering was ineffective in some tests since there were particles in the test cell. Perhaps a better method of washing the test cell and filling in a more isolated situation should be considered for future work. the points for future work are:

- A more investigation into PDIV definition of IEC 61294 and testing more dielectric liquids.
- studying the causes of PD pulses starting at the positive half-cycle simultaneously with the negative half-cycle.
- Quantifying the space charge effect based on Mott-Gurney's law and developing the proper relations based on needle-plane geometry.
- further tests with rectified HV source to compare PD patterns with full-cycle tests.
- testing optical PD detection method for PDIV level detection.

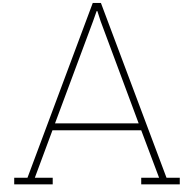
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Appendix 1: Previous studies

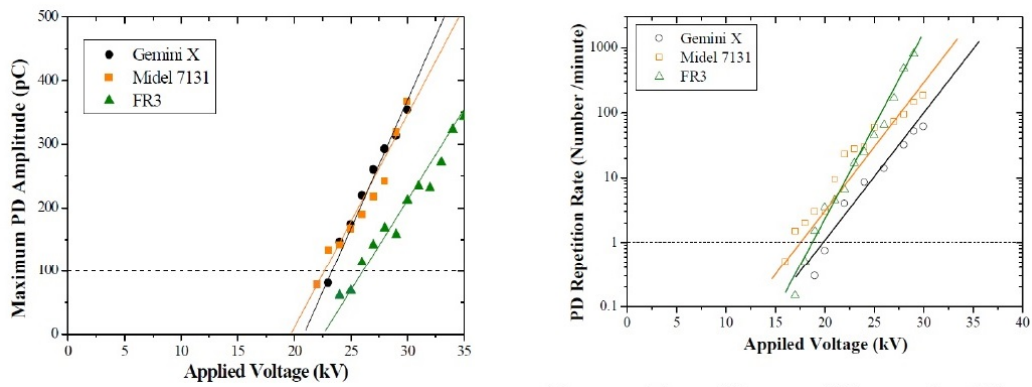


Figure A.1: maximum PD amplitude and repetition rate for three different materials adopted from [33]

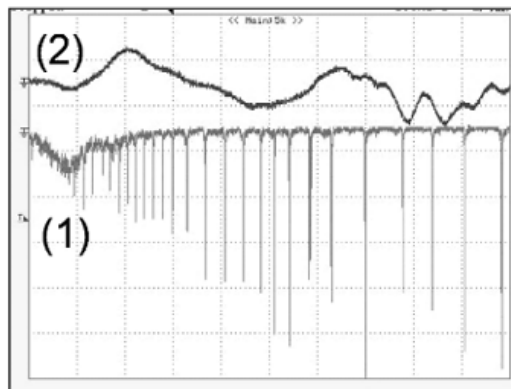


Figure A.2: Comparison of conventional detected PD signal (2) and optical detected pulses (1) in oil adopted from [65]

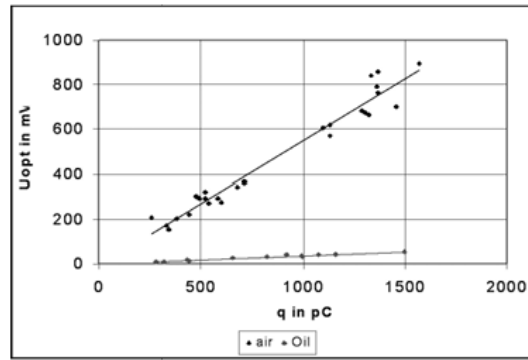


Figure A.3: The relationship of the optical signal and the discharge level in the air and in oil at the same arrangements adopted from [65]

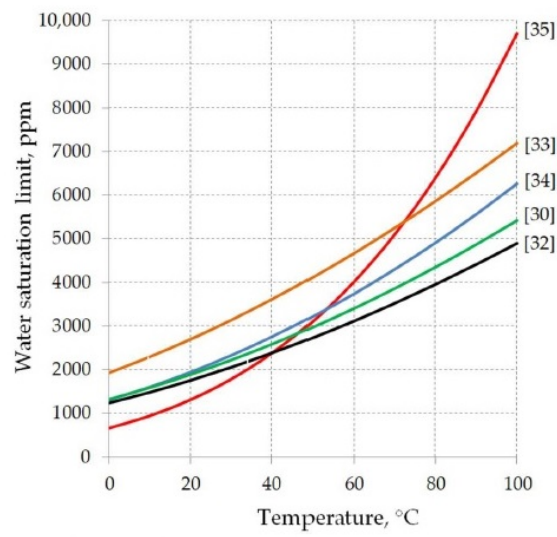


Figure A.4: details of MDEL 7131 water contamination effect [74]

B

Appendix 2: Datasheets

MIDEL 7131 Fluid Properties

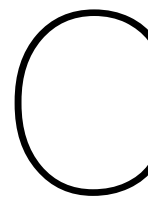
Property	Test Method	IEC 61099	MIDEL 7131
		Un-used new fluid property requirements	Typical Values
Physical			
Colour	ISO 2211	Max. 200 Hazen	125
Appearance		Clear, free from water and suspended matter and sediment	Clear, free from water and suspended matter and sediment
Density at 20°C (kg/dm ³)	ISO 3675 or ISO 12185	Max. 1	0.97
Kinematic Viscosity (mm ² /sec)	ISO 3104		
at 40°C		Max. 35	29
at -20°C		Max. 3000	1440
Flash Point PMCC (°C)	ISO 2719	Min. 250	260
Fire Point (°C)	ISO 2592	Min. 300	316
Pour Point (°C)	ISO 3016	Max. -45	-56
Crystallization	IEC 61099 (2010 Annex A)	No crystals	No crystals
Biodegradation	Readily Biodegradable		Fully/Readily Biodegradable
Electrical			
Dielectric Breakdown (kV)	IEC 60156	Min. 45	>75
Power Factor at 90°C	IEC 60247	Max. 0.03	<0.008
DC Resistivity at 90°C (GΩ.m)	IEC 60247	Min. 2	>20
Chemical			
Water Content (mg/kg)	IEC 60814	Max. 200	50
Acidity (mg KOH/g)	IEC 62021-1 or IEC 62021-2	Max. 0.03	<0.03
Oxidation Stability (164hr)	IEC 61125C		
Total Acidity (mg KOH/g)		Max. 0.3	0.02
Total Sludge (% mass)		Max. 0.01	<0.01

MIDEL 7131 is supplied in accordance with IEC industry standard acceptance values and test methods. The displayed typical values do not form part of this specification.

Figure B.1: MIDEL 7131 details from Midel company' website [31]

Table B.1: Nytro 10XN Transformer Oil product datasheet [29]

Property	Unit	Guaranteed Data	Typical Data
1. Physical			
Appearance	–	Clear, free from sediment	–
Density, 20°C	kg/dm ³	0.895	0.876
Viscosity, 40°C	mm ² /s	8.0	7.6
Viscosity, -30°C	mm ² /s	800	730
Flash Point, PM	°C	140	144
Pour Point	°C	-45	-60
2. Chemical			
Neutralization value	mg KOH/g	0.03	<0.01
Corrosive sulphur	–	Non-corrosive	–
Aromatic content	%	–	7
Antioxidant, phenols	Wt%	–	0.3
Water content	mg/kg	30	<20
3. Electrical			
Dielectric dissipation factor at 90°C	–	0.005	<0.001
Interfacial tension	mN/m	40	45
Breakdown voltage	kV	30	40-60
4. Oxidation Stability			
At 100°C 164h	–	–	–
Neutralization value	mg KOH/g	–	<0.01
Sludge	Wt%	–	<0.005
At 120°C	–	–	–
Induction period	h	–	>236
At 110°C 28 days	–	–	–
Baader test	–	–	–
Saponification number	mg KOH/g	0.15	–
Sludge	Wt%	–	<0.005
Dielectric dissipation factor at 90°C	–	0.003	–



Appendix 3: Python Script

Python script used for analysing the digitizer data for cyclohexane at 30 Hz and 14.52 kV with sample rate 200 K

```
1  """
2  Created on Sat Feb 17 12:02:58 2024
3
4  @author: Mohammad Fazlalizadeh
5
6  Norges teknisk-naturvitenskapelige universitet
7
8  """
9  import pandas as pd
10 import matplotlib.pyplot as plt
11 import os
12 from scipy.signal import find_peaks
13 import numpy as np
14 from scipy.optimize import minimize
15 import math
16
17 # Function to load data from a folder path and store it in the dictionary
18 def load_data(folder_path):
19     file_list = os.listdir(folder_path)
20     data = {}
21     for file_name in file_list:
22         if file_name.endswith('.asc'):
23             file_path = os.path.join(folder_path, file_name)
24             data[file_name] = pd.read_csv(file_path, delim_whitespace=True,
25             ↪ header=None, skiprows=13)
26     return data
27
28 # Cyclohexane
29 # Define folder paths for each frequency and sampling rate
30 folder_paths = {
31     '30Hz': {
32         'long_33630': 'file path should be added here',
33         'PDtriggered_200KSs': 'file path should be added here'
34     },
35     '45Hz': {
36         '50KSs_33630': 'file path should be added here',
```

```

36     '200KSs_33630': 'file path should be added here'
37 },
38 '60Hz': {
39     '50KSs_17071': 'file path should be added here',
40     '100KSs_17071': 'file path should be added here',
41     '100KSs_33630': 'file path should be added here',
42 }
43 }
44 # Specify the folder path to save the graphs
45 save_folder_path = 'save folder path should be added here'
46
47 # Create the save folder if it doesn't exist
48 os.makedirs(save_folder_path, exist_ok=True)
49
50 # Create an empty dictionary to store data
51 data_dict = {
52     '30Hz': {'long_33630': {}, 'PDtriggered_200KSs': {}},
53     '45Hz': {'50KSs_33630': {}, '200KSs_33630': {}},
54     '60Hz': {'50KSs_17071': {}, '100KSs_17071': {}, '100KSs_33630': {}}
55 }
56
57 # Load data from folder paths
58 for freq, freq_data in folder_paths.items():
59     for sampling_rate, folder_path in freq_data.items():
60         data_dict[freq][sampling_rate] = load_data(folder_path)
61
62 ### Plotting data in different graphs and saving them
63 for freq, freq_data in data_dict.items():
64     for sampling_rate, sampling_data in freq_data.items():
65         graph_number = 0 # Reset graph number for each combination of frequency and
66         ↪ sampling rate
67         for file_name, data in sampling_data.items():
68             plt.plot(data.iloc[:, 0], data.iloc[:, 1], label=file_name)
69             plt.title(f'{file_name} (Graph {graph_number})')
70             plt.xlabel('Time')
71             plt.ylabel('Voltage')
72             save_file_path_svg = os.path.join(save_folder_path,
73             ↪ f'{file_name.replace(".asc", f"_plot_{graph_number}.svg")}')
74             save_file_path_jpeg = os.path.join(save_folder_path,
75             ↪ f'{file_name.replace(".asc", f"_plot_{graph_number}.jpg")}')
76             plt.savefig(save_file_path_svg, format='svg')
77             plt.savefig(save_file_path_jpeg, format='jpeg')
78             plt.show()
79             graph_number += 1 # Increment graph number
80
81 ### Extracting data from dictionary and plotting on the same graph
82 # Define the location in the nested dictionary
83 dictionary_loc = ('30Hz', 'long_33630')
84 # The name of the graphs being saved
85 plot_title = "Cyclohexane_30Hz_14.52kVpeak_1uAV_33360Vrv.asc"
86 peakvoltage = 14.52
87 frequency = 30
88 voltage_data_set = 39
89 current_data_set = 38
90 sensor_ratio = 1000 # nA per V
91 FFT_cycle_num = 24

```

```

89 frequency_limit = 300
90 sample_rate = 200000
91 plot_show_time = (-0.0440, -0.008) #cyclo 30
92
93 # Access the nested dictionary based on the location specified
94 if dictionary_loc[0] in data_dict and dictionary_loc[1] in
    ↪ data_dict[dictionary_loc[0]]:
95     nested_dict = data_dict[dictionary_loc[0]][dictionary_loc[1]]
96     file_names_list = list(nested_dict.keys())
97     # Check for the file_names_list to contain at least one name
98     if file_names_list:
99         # Check if the indices are within the range
100        if 0 <= voltage_data_set < len(file_names_list) and 0 <= current_data_set <
            ↪ len(file_names_list):
101            # Access the specified data frames in the nested dictionary
102            voltage_data = list(nested_dict.values())[voltage_data_set]
103            current_data = list(nested_dict.values())[current_data_set]
104            # Plotting Voltage and Current data on the same graph
105            plt.figure(figsize=(20, 12)) # Set the figure size
106            plt.plot(voltage_data.iloc[:, 0], voltage_data.iloc[:, 1] / 50,
                ↪ label=f"Voltage[V]/50 ({file_names_list[voltage_data_set]})")
107            plt.plot(current_data.iloc[:, 0], current_data.iloc[:, 1] * sensor_ratio,
                ↪ label=f"Current[nA] ({file_names_list[current_data_set]})")
108            plt.title("Voltage and Current raw Data")
109            plt.xlabel('Time')
110            plt.ylabel('Value')
111            plt.legend()
112            plt.grid(True) # Add grid
113            plt.show()
114        else:
115            print("Invalid indices provided.")
116    else:
117        print("No data available in the nested dictionary.")
118 else:
119     print("Location not found in the main dictionary.")
120 # Write data frames into new separate arrays
121 num_voltage = voltage_data.iloc[:, 1].to_numpy(dtype=np.float64)
122 num_current = current_data.iloc[:, 1].to_numpy(dtype=np.float64) * sensor_ratio
123 time = voltage_data.iloc[:, 0].to_numpy(dtype=np.float64)
124 # Combine the arrays horizontally
125 array = np.column_stack((time, num_voltage, num_current))
126
127 #%% Finding zero crossings and complete cycles
128 # Function to find the index of the maximum peak in the voltage data
129 def find_max_peak_index(voltage_data_values):
130     max_peak_index = np.argmax(voltage_data_values)
131     return max_peak_index
132
133 # Function to find the index corresponding to pi/2 before the maximum voltage
134 def find_index_pi_over_2_before_max(data_values, max_peak_index, frequency):
135     # Calculate the period of the sinusoid (assuming a sinusoidal waveform)
136     period = 1 / frequency
137     # Calculate the time difference corresponding to pi/2 phase difference
138     time_difference_pi_over_2 = period / 4
139     # Find the time value corresponding to pi/2 before the maximum voltage

```

```

140     time_pi_over_2_before_max = data_values[max_peak_index, 0] -
        ↪ time_difference_pi_over_2
141     # Find the index corresponding to this time value
142     index_pi_over_2_before_max = np.argmin(np.abs(data_values[:, 0] -
        ↪ time_pi_over_2_before_max))
143     return index_pi_over_2_before_max
144 # Function to trim the data at the zero crossing just before the maximum voltage
145 def trim_to_zero_crossing_before_max(data_values, zero_crossing_index ,
        ↪ FFT_cycle_num, samples_per_period):
146     cycles_points = samples_per_period * FFT_cycle_num
147     trimmed_array_data = data_values[0: cycles_points]
148     return trimmed_array_data, cycles_points
149 samples_per_period = int(sample_rate / frequency)
150 voltage_max_peak_index = find_max_peak_index(array[:,1])
151 index_pi_over_2_before_max = find_index_pi_over_2_before_max(array,
        ↪ voltage_max_peak_index, frequency)
152 trimmed_array, cycles_points =
        ↪ trim_to_zero_crossing_before_max(array, index_pi_over_2_before_max, FFT_cycle_num,
        ↪ samples_per_period)
153
154 %% Perform FFT on each signal
155 fft_voltage = np.fft.fft(trimmed_array[:, 1])
156 fft_current = np.fft.fft(trimmed_array[:, 2])
157 # Frequency values corresponding to the FFT result
158 freq1 = np.fft.fftfreq(len(fft_voltage), d=(trimmed_array[1, 0] - trimmed_array[0,
        ↪ 0]))
159 # d is the distance between two data points for example 1399-1400
160 freq2 = np.fft.fftfreq(len(fft_current), d=(trimmed_array[1, 0] - trimmed_array[0,
        ↪ 0]))
161 # d is the distance between two data points for example 1399-1400
162 # Find the index corresponding to the main frequency
163 main_f_voltage_index = np.argmax(np.abs(fft_voltage))
164 main_f_current_index = np.argmax(np.abs(fft_current))
165 # Extract the phase angle for the main frequency
166 phase_angle_voltage = np.angle(fft_voltage[main_f_voltage_index])
167 phase_angle_current = np.angle(fft_current[main_f_current_index])
168 # Convert the phase angle from radians to degrees
169 phase_angle_voltage_degrees = np.degrees(phase_angle_voltage)
170 phase_angle_current_degrees = np.degrees(phase_angle_current)
171 # Extract the amplitude of the main component in the current signal
172 #amplitude_main =
        ↪ np.abs(fft_current[main_f_current_index])/(0.5*len(trimmed_current)) # amplitude
        ↪ is multiplied by half of the number of data points by definition of FFT
173 # Set the threshold for FFT component detection
174 threshold = 1.00036 # Adjust this threshold as needed
175 threshold_voltage = threshold * np.max(np.abs(fft_voltage))
176 # Find FFT components above the threshold for voltage
177 significant_indices_voltage = np.where(np.abs(fft_voltage) > threshold_voltage)[0]
178 significant_freqs_voltage = freq1[significant_indices_voltage]
179 # Find the main frequency index for voltage
180 main_f_voltage_index = np.argmax(np.abs(fft_voltage))
181 # Exclude the main frequency from the list of significant frequencies
182 significant_freqs_voltage = significant_freqs_voltage[significant_freqs_voltage !=
        ↪ freq1[main_f_voltage_index]]
183 significant_freqs_voltage = significant_freqs_voltage[significant_freqs_voltage !=
        ↪ -freq1[main_f_voltage_index]]

```

```

184 # Find corresponding indices in current FFT and eliminate those components
185 significant_indices_current = [np.abs(freq2 - freq).argmin() for freq in
    ↪ significant_freqs_voltage]
186 fft_current_filtered = fft_current.copy()
187 fft_current_filtered[significant_indices_current] = 0
188 fft_current_filtered[0] = 0
189
190 %% Plot the FFT output of the voltage data
191 plt.figure(figsize=(20, 12))
192 # fft voltage and current amplitude per length that is necessary
193 FFT_voltage = fft_voltage / (0.5 * len(trimmed_array))
194 FFT_current = fft_current / (0.5 * len(trimmed_array))
195 fft_current_filtered = fft_current_filtered / (0.5 * len(trimmed_array))
196 plt.plot(freq1, np.abs(FFT_voltage), label='FFT of the voltage signal')
197 plt.title(plot_title.replace('.asc', '_FFT'))
198 plt.axvline(x= frequency, color='red', linestyle='--', label= dictionary_loc[0])
199 plt.axvline(x= frequency_limit, color='orange', linestyle='--', label= 'cut-off
    ↪ frequency')
200 plt.xlim(0, 1000)
201 #plt.ylim(0,100)
202 plt.xlabel('Frequency')
203 plt.ylabel('Amplitude [v]')
204 plt.legend()
205 plt.grid()
206 # Set y-axis to be logarithmic
207 plt.yscale('log')
208 save_file_path_svg = os.path.join(save_folder_path, plot_title.replace('.asc',
    ↪ '_FFT_voltage.svg'))
209 save_file_path_jpeg = os.path.join(save_folder_path, plot_title.replace('.asc',
    ↪ '_FFT_voltage.jpeg'))
210 plt.axhline(y= threshold_voltage / (0.5 * len(trimmed_array)), color='red',
    ↪ linestyle='--', label= dictionary_loc[0])
211 plt.savefig(save_file_path_svg, format='svg')
212 plt.savefig(save_file_path_jpeg, format='jpeg')
213 plt.show()
214
215 # Save the voltage FFT to an ASC file
216 np.savetxt(os.path.join(save_folder_path, plot_title.replace('.asc',
    ↪ '_FFT_voltage_data.asc')), np.column_stack((freq1, FFT_voltage)),
    ↪ header="Frequency\tAmplitude", delimiter='\t', comments='')
217 %% Plot the FFT output of the Current data
218 plt.figure(figsize=(20, 12))
219 plt.plot(freq2, np.abs(FFT_current)* sensor_ratio , color='green', label='FFT of the
    ↪ current signal')
220 plt.plot(freq2, np.abs(fft_current_filtered)* sensor_ratio , color='red',
    ↪ label='filtered current signal')
221 plt.title(plot_title.replace('.asc', '_FFT'))
222 plt.axvline(x= frequency, color='red', linestyle='--', label= dictionary_loc[0])
223 plt.axvline(x= frequency_limit, color='orange', linestyle='--', label= 'cut-off
    ↪ frequency')
224 plt.xlim(0, 1000)
225 #plt.ylim(0,100)
226 plt.xlabel('Frequency')
227 plt.ylabel('Amplitude [nA]')
228 plt.legend()
229 plt.grid()

```



```

230 # Set y-axis to be logarithmic
231 plt.yscale('log')
232 save_file_path_svg = os.path.join(save_folder_path, plot_title.replace('.asc',
↳ '_FFT_current.svg'))
233 save_file_path_jpeg = os.path.join(save_folder_path, plot_title.replace('.asc',
↳ '_FFT_current.jpeg'))
234 plt.savefig(save_file_path_svg, format='svg')
235 plt.savefig(save_file_path_jpeg, format='jpeg')
236 plt.show()
237
238 # Save the voltage FFT to an ASC file
239 np.savetxt(os.path.join(save_folder_path, plot_title.replace('.asc',
↳ '_FFT_current_data.asc')), np.column_stack((freq1,FFT_voltage)),
↳ header="Frequency\tAmplitude", delimiter='\t', comments='')
240 %% Calculating the Capacitance
241 # Find the indices where freq1 is above the frequency limit
242 indices_above_limit = np.where(np.abs(freq1) > frequency_limit)[0]
243 # Set the adjacent elements in fft_voltage to zero for the indices above the limit
244 FFT_voltage[indices_above_limit] = 0
245 FFT_voltage[indices_above_limit + 1] = 0
246 # eliminate the offset within the FFT signal of the voltage
247 FFT_voltage[0] = 0
248 # Set the adjacent elements in fft_current to zero for the indices above the limit
249 FFT_current[indices_above_limit] = 0
250 FFT_current[indices_above_limit + 1] = 0
251 FFT_current[0] = 0
252 # Find the mean value of voltage and current
253 mean_voltage = np.mean(trimmed_array[:, 1])
254 mean_current = np.mean(trimmed_array[:, 2])
255 # Subtract the mean value from voltage and current to remove offset
256 trimmed_voltage_no_offset = trimmed_array[:, 1] - mean_voltage
257 trimmed_current_no_offset = trimmed_array[:, 2] - mean_current
258 # find the max value of the voltage with no offset
259 trimmed_voltage_no_offset_max_index = np.argmax(trimmed_voltage_no_offset)
260 # Calculate the derivative of the voltage in the frequency domain
261 derivative_voltage_freq_domain = 2j * np.pi * freq1 * FFT_voltage * (0.5 *
↳ len(trimmed_array))
262 # Perform the inverse FFT to obtain the derivative in the time domain
263 derivative_voltage = np.fft.ifft(derivative_voltage_freq_domain)
264 # reproducing the voltage from the FFT of the voltage after filtering
265 voltage_redo = np.fft.ifft(FFT_voltage) * (0.5 * len(trimmed_array))
266 # reproducing the current from the FFT of the current after filtering
267 current_redo = np.fft.ifft(fft_current_filtered) * (0.5 * len(trimmed_array))
268 # capacitance of each harmonic by  $i(s)/(v(S)*2\pi*f)$  formula
269 capacitance_array = FFT_current * 1e-9 / (FFT_voltage * 2j * np.pi * freq1+ 1e-30)
270 # Calculate capacitive current using the optimal capacitance
271 capacitive_current = derivative_voltage *
↳ np.real(capacitance_array[main_f_voltage_index]) * 1e9 #113.5e-15 223.5e-15
272 # amplitude of the voltage has been multiplied because derivative of the voltage has
↳ been calculated based on trimmed voltage
273 resistive_current = (current_redo - capacitive_current)
274 fft_conductive_current = np.fft.fft(resistive_current)
275 # Frequency values corresponding to the FFT result
276 freq_cond_current = np.fft.fftfreq(len(fft_conductive_current),
↳ d=(resistive_current[1] - resistive_current[0]))
277 # d is the distance between two data points for example 1399-1400

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278 # Find the index corresponding to the main frequency
279 main_f_cc_index = np.argmax(np.abs(fft_conductive_current))
280 main_freq_cc = freq_cond_current[main_f_cc_index]
281 # Find the indices where freq1 is above the frequency limit
282 indices_other_than_main = np.where(np.abs(freq_cond_current) !=
  → np.abs(main_freq_cc))[0]
283 # Set the adjacent elements in fft_conductive_current to zero for the indices above
  → the limit
284 fft_conductive_current[indices_other_than_main] = 0
285 # Perform inverse FFT to get the time-domain signal
286 conductive_current_redo = np.fft.ifft(fft_conductive_current)
287 Max_negative_resistive_current = -1 *
  → resistive_current[np.argmax(resistive_current)]
288 Max_positive_resistive_current = -1 * resistive_current[np.argmax(-1 *
  → resistive_current)]
289 Max_first_harmonic_resistive_current =
  → conductive_current_redo[np.argmax(conductive_current_redo)]
290 plt.plot(trimmed_array[:, 1], color='red', label = 'measured')
291 plt.plot(voltage_redo, color='blue', label = 'filtered')
292 plt.xlim(0, 20000)
293 plt.legend()
294
295 %% Create the plot
296 bandwidth_title = f"bandwidth_0-{frequency_limit}Hz"
297 combo_plot_title = f"{plot_title.replace('.asc', '')}_{bandwidth_title}.asc"
298 fig, ax1 = plt.subplots(figsize=(14, 6))
299 # Find the indices corresponding to plot_show_time
300 start_time, end_time = plot_show_time
301 start_index = np.argmax(trimmed_array[:, 0] >= start_time)
302 end_index = np.argmax(trimmed_array[:, 0] >= end_time)
303 # Plot the real part of capacitive current on the left y-axis
304 ax1.plot(trimmed_array[start_index:end_index, 0], -1*
  → trimmed_voltage_no_offset[start_index:end_index] / 1000, 'b-',
  → label='measured_Voltage')
305 ax1.plot(trimmed_array[start_index:end_index, 0], -1*
  → voltage_redo[start_index:end_index] / 1000, color='violet',
  → label='filtered_Voltage')
306 ax1.set_xlabel('Time')
307 ax1.set_ylabel('kV', color='b')
308 ax1.tick_params('y', colors='b')
309 # Create the right y-axis and plot the real part of derivative voltage
310 ax2 = ax1.twinx()
311 ax2.plot(trimmed_array[start_index:end_index,0], -1*
  → capacitive_current[start_index:end_index], 'red', label='Capacitive_current')
312 ax2.plot(trimmed_array[start_index:end_index,0], -1*
  → trimmed_current_no_offset[start_index:end_index], 'orange',
  → label='Measured_current')
313 ax2.plot(trimmed_array[start_index:end_index,0], -1*
  → resistive_current[start_index:end_index], 'green', label='Conductive_current ')
314 #ax2.set_ylim(-1800, 1800)
315 #plt.axvline(x=0, color='red', linestyle='--', label='0V')
316 ax2.set_ylabel('nA', color='r')
317 ax2.tick_params('y', colors='r')
318 ax2.grid(True)
319 # Display the legend
320 lines, labels = ax1.get_legend_handles_labels()

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321 lines2, labels2 = ax2.get_legend_handles_labels()
322 ax2.legend(lines + lines2, labels + labels2, loc='lower left')
323 plt.title(combo_plot_title)
324 save_file_path_svg = os.path.join(save_folder_path, combo_plot_title.replace('.asc' ,
    ↪ '.svg'))
325 save_file_path_jpeg = os.path.join(save_folder_path, combo_plot_title.replace('.asc'
    ↪ , '.jpeg'))
326 plt.savefig(save_file_path_svg, format='svg')
327 plt.savefig(save_file_path_jpeg, format='jpeg')
328 plt.show()
329 # Save the resistive_current data to an ASC file
330 #np.savetxt(os.path.join(save_folder_path, plot_title.replace('.asc',
    ↪ '_resistive_current_data.asc')), np.column_stack((trimmed_array[:,0],
    ↪ resistive_current)), header="Time (s)\tResistive Current (nA)", delimiter='\t',
    ↪ comments='')
331
332 #%%
333 fig, ax1 = plt.subplots(figsize=(6, 4))
334 full_period_points = int((1 / frequency) * sample_rate) + 5
335 cc_start = full_period_points * 0 + 5650
336 ax1.plot(trimmed_array[cc_start:full_period_points*1 + cc_start, 0],
    ↪ -1*resistive_current[cc_start:full_period_points*1 + cc_start], 'green',
    ↪ label='Resistive current')
337 ax1.set_ylim(-50, 50)
338 # Add a horizontal line at y=0 on ax1
339 ax1.axhline(y=0, color='black', linestyle='--')
340 # Create the right y-axis and plot the real part of derivative_voltage
341 ax2 = ax1.twinx()
342 peakvoltage_label = f"Vpeak:{peakvoltage}[kV]"
343 ax2.plot(trimmed_array[cc_start:full_period_points*1 + cc_start, 0],
    ↪ -1*voltage_redo[cc_start:full_period_points*1 + cc_start]/1000, 'blue', label=
    ↪ peakvoltage_label)
344 # Set labels for y-axes
345 ax1.set_ylabel('Current [nA]')
346 ax2.set_ylabel('Voltage [kV]')
347 ax1.set_xlabel('Time [s]')
348 # Add legends
349 #ax1.legend(loc='lower left')
350 ax2.legend(loc='lower left')
351 # Adjust plot layout to prevent cropping in SVG
352 plt.tight_layout()
353 sec_combo_plot_title = f"{plot_title.replace('.asc',
    ↪ 'C&V_vs._t_filterbw')}_frequency_limit.asc"
354 save_file_path_svg = os.path.join(save_folder_path,
    ↪ sec_combo_plot_title.replace('.asc', '.svg'))
355 save_file_path_jpeg = os.path.join(save_folder_path,
    ↪ sec_combo_plot_title.replace('.asc', '.jpeg'))
356 plt.savefig(save_file_path_svg, format='svg')
357 plt.savefig(save_file_path_jpeg, format='jpeg')
358 plt.show()

```